

DIGITAL COMMUNICATION TECHNIQUES

- **Introduction and general issues**
- **Parallel bus communication (IEEE 488)**
- **Serial bus communications (RS232 ... and USART driving)**
- **Field bus (Fip, CanBus)**

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- **Encoding = bijective correspondence between transmitted info and symbols**
- **The encoding can be arbitrary but must be followed by both the communicating devices otherwise non comprehension**
- **Symbols are sent through signals associated to values ranges (within the calculator) or subject to modulation (in the used medium)**
- **Transmission media: wired twists and/or shielded, Wi-Fi (pay attention to the environmental disturbs), coaxial cables, optical fibers (pay attention to costs, maintenance, installation and required competences).**
- **Characteristics:**
 - ***bandwidth* (100 KHz - 1 GHz) ? max. transmission velocity**
 - ***attenuation* (0.4-1.5 dB/Km) ? max. distance without repeaters ($e^{-\alpha l}$)**
 - ***propagation velocity* (15000-150000 Km/s)**
 - ***characteristic impedance* (50-100 Ohm)**
 - ***noise immunity* (S/N, 100 dB)**

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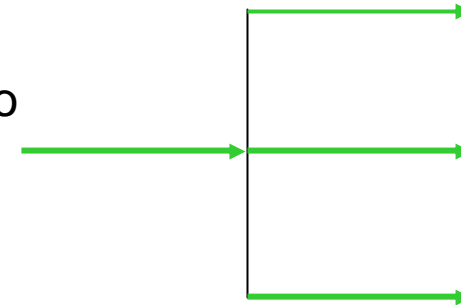
•THE GENERAL PROBLEMS OF THE DIGITAL COMMUNICATIONS

- Transfer direction (simplex, half and full duplex)
- Organize the transmission content (bits, characters and messages)
- Data synchronization and recognition (auxiliary signals)
- Error avoidance
- Error recognition and correction
- Spatial and temporal coordination of interlocutors through protocols (who can transmit and how) and routing (shortest vs most efficient path policies)

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- **TRANSFER DIRECTION**

- **simplex**, only one single-direction channel that allows to transfer from a sender to many (multicast) or all (broadcast) destinations.



- **half duplex**: 1 bi-directional shared channel or 2 single-direction opposed channels of which only one can send or receive (for instance the walkie talkie). Simple approach typical in local networks.



- **full duplex**: transmission and reception can occur simultaneously on the same channel (for instance telephone). The communication can consist of 2 simplex opposed channels or of 1 bi-directional channel.



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- **TRANSMISSION CONTENT**

- Hierarchical structure of information (bits, characters and messages)
- Characters are made up by bits (7, 8 or 9)
- ASCII or different format encoding to make the transmission content more compact and improve the robustness (immunity) to the noise
- Characters are organized in messages to
 - » Recover in a easier way the information content, by simplifying error detection and recognition
 - » Synchronize interlocutors
 - » Allow concurrent communications on the same transmission mean
- Characters length influenced by the error management techniques (duplications and redundancy)

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- **SYNCHRONIZATION AND RECOGNITION**

- Synchronization = to establish what time windows can be used to exchange information and what symbols are used to represent them
- Auxiliary signals
- If parallel communications ? one or more additive lines
- If serial communications ? data and synchronization info are brought by the same line (for instance in synchronous communication)
- Bit, characters and message structure ? synchronization signals must be organized in a similar structure to recognize those bits that constitute a character and the bits that constitute a message (for instance in the asynchronous communication start and stop bits, SOM and EOM, ...)

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- **ERROR PREVENTION**

- Disturbing phenomena restraint (filtering, differential amplification with high CMRR)
- Preventing from disturbs (shielding, sheaths, twisted pairs)
- “Robust” signals (high S/N)

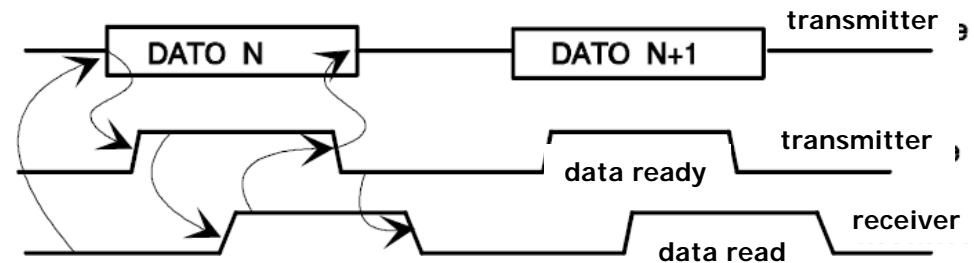
- **ERROR RECOGNITION AND CORRECTION**

- Errors can never completely eliminated ? error rate $\neq 0$ that must be accepted and foreseen in the specs
- Errors rate can be even high but it is important that errors are detected and when possible original data recreated
 - »Not detected errors (if possible with a low rate)
 - »Detected errors with relative information discarding
 - »Recognized errors with information recovery

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- **PARALLEL COMMUNICATIONS**

- Groups of bits simultaneously transmitted
 - Short distances (mechanical use and disturbs)
 - System bus (with separated data and addresses) vs communication bus (data and addresses together)
 - Auxiliary signals to synchronize the transfers (asynchronous bus). For instance:
 - » ReadReq (memory read requirement ? the address is put on the bus)
 - » DataRdy (the data are ready and on the bus: the signal is managed by the sender)
 - » Ack (ok reception of ReadReq and DataRdy and of the data)



- The slowest node defines the velocity
- Existent standards for interfacing: IEEE 488 and Centronics

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- OBSERVATION:** why do not use synchronous buses?

In a synchronous bus the clock signal is included in the control lines

The use of a synchronous bus requires a fixed protocol for the communication that is a sequence of steps executed at a certain clock cycle (CC) sequence

For instance: in the first CC a “read” command and the relative address are sent by a device to the memory. The memory after few cycles (for example 5) provides the requested data on the bus, then the device reads in three CC and elaborates...

This can be implemented with a small finite state machine ? very few digital logic required, very fast response and small interfacing, but ...

... all the devices must work at the same frequency (this is not said)

... the buses *must* be kept physically short to avoid the misalignment with respect other devices (*clock skew*)

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•PARALLEL COMMUNICATIONS: IEEE 488 STANDARD BUS

GPIB = General Purpose Interface Bus

At the beginning ('75) proposed by HP to carry out a network between calculation systems and instrumentations by different manufacturers (named GPIB)

Successively it was adopted as a standard by IEEE and redefined

The standard allows to connect up to 15 devices on the same bus

16 lines of which 8 of data, 3 for the *handshake* and 5 for interfacing management

The devices can play 3 possible roles: “*talker*”, “*listener*” and “*controller*”: thus it is possible to specify type and functions of the connected instrumentation

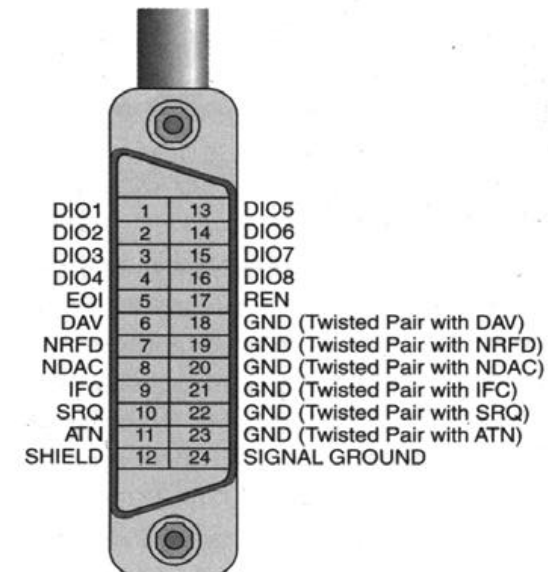
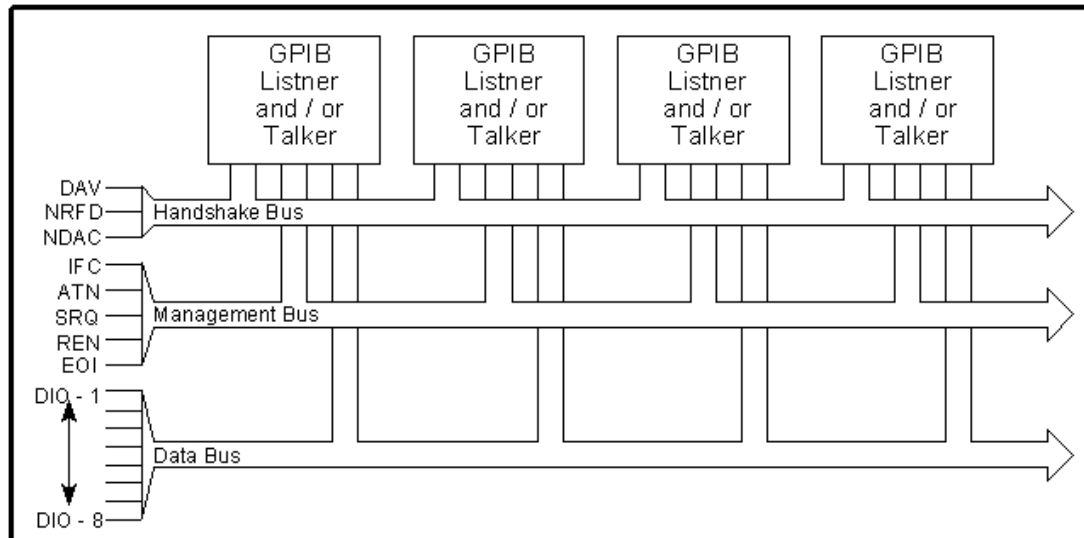
A device is programmed to be the System Controller at the beginning or the Active Controller during the normal working cycle and among its main functions can send a reset (IFC - interface clear) to all the devices or enable them to transmit (REN – remote enable)

Transmission velocity = MB/sec

Max. length of the bus 20 m; max. distance between devices 4 m.

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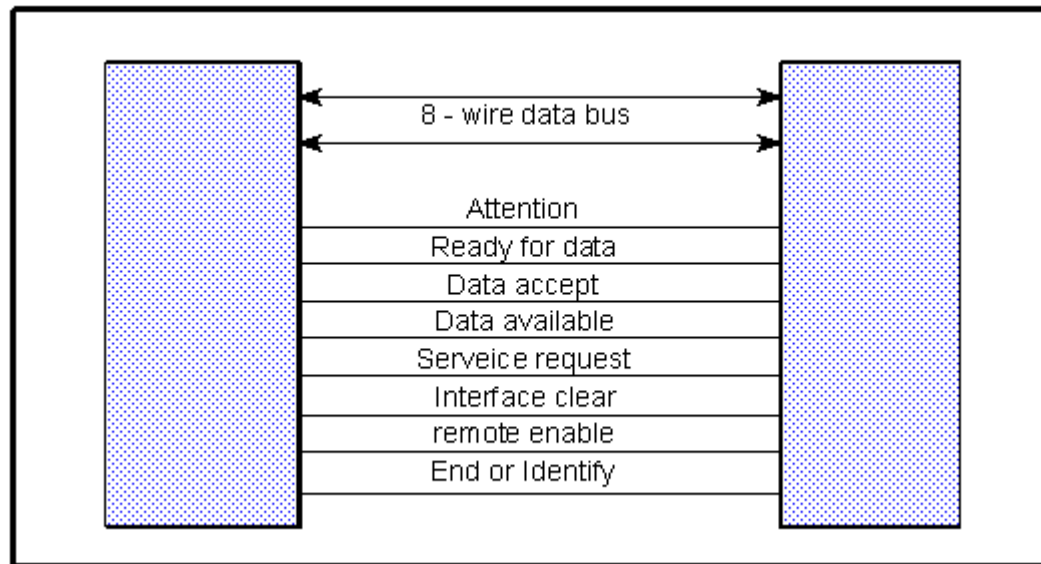
- **PARALLEL COMMUNICATIONS: IEEE 488 STANDARD BUS**



- Several "controllers" can be present but only 1 at a time will be active (In Charge)
- The control is passed among the "active controllers" through a token mechanism (only the System Controller is established on the basis of the given system configuration)
- A "listener" device receives data from the bus only when the "controller" enables it
- Similarly a "talker" device transmits only when enabled
- The "controller" can set a "multicast" transmission mode between a "talker" device and a "listeners" group

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- IEEE 488: lines to interface management



- ATN (attention) is asserted by the “Controller” to indicate the placement of an address/control on the bus. The “Controller” disables it to allow the “talker” to transmit then re-enables it to continue in the transmission management
- EOI (end or identify) two possible uses: if enabled by the “talker” it means end of transmission; if enabled by the “controller”, together with it ATN, it stands for a (parallel poll) to all the devices

- IFC (Interface Clear) enabled only by the “System Controller” to bring all the devices in a known initial state
- REN (Remote Enable) enabled only by the “System Controller” to arrange the devices in a remote control mode (command reception) that disables the local control from front panel (if present). The control will become complete only when those devices will be addressed as “listener”
- SRQ (Service Request) = a kind of interrupt by a whatever device that asks the “Controller” for a certain action. The “Controller” identifies the emitting device through a “serial poll”. When polled, the device releases the SRQ signal

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•IEEE 488: handshake signals

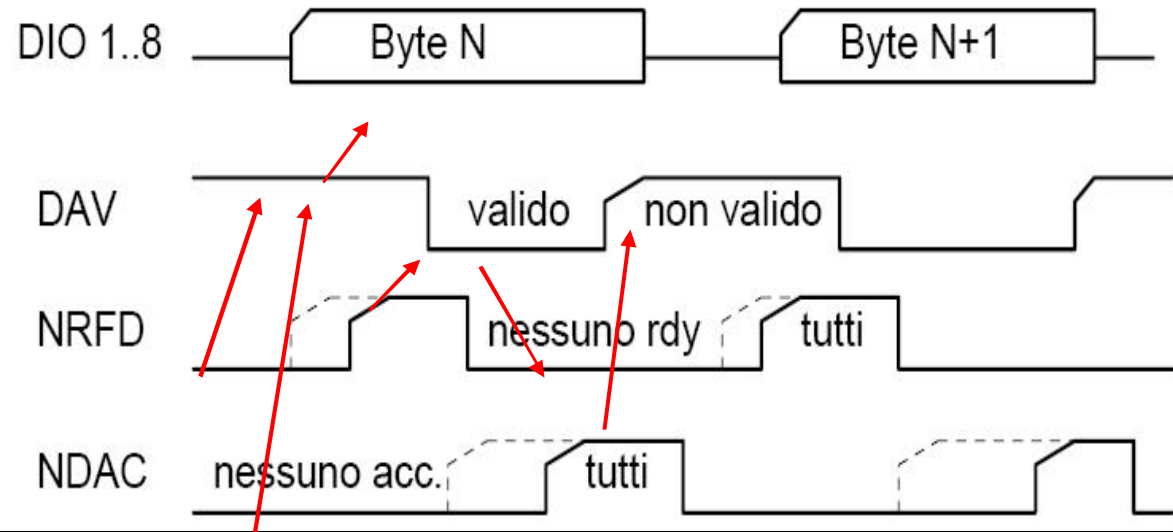
TTL inverse logic signals

NRFD (Not Ready For Data): set high from receivers (low active) if ready to accept data. Open collector

DIO: data lines

DAV (Data Valid): active low only after the data have been put on the bus and the NRFD signal has been found high

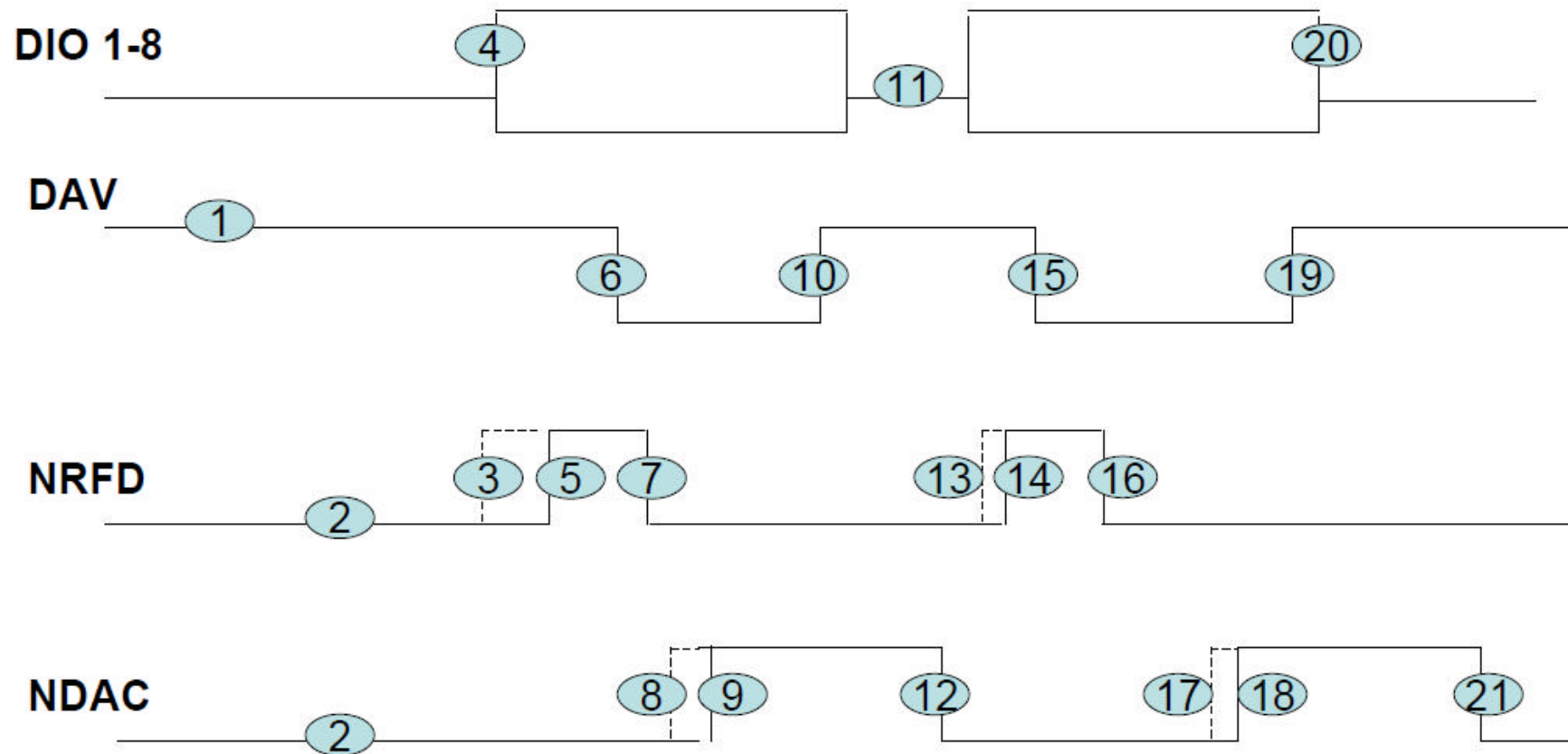
NDAC (Not Data Accepted): set low by receivers until a new byte has not been acquired. If high, DAV will be disabled and DIO lines changed with a new value. Open collector



Handshake

1. DAV high if the sender ("controller"/"talker") wants to transmit. If NRFD and NDAC low, data are written in the bus DIO. Only when NRFD high, DAV will be set low by the sender
2. Every receiver keeps NRFD low to signal to be "busy"; when the last one will have complete the reception, NDAC will be set high and the sender will set DAV high to transmit the next byte.
3. If, when set high DAV, the sender finds NDAC and NRFD high an error happens. Moreover if a device cannot perform correctly the handshake and releases the NDAC or the NRFD a error happens.
4. The error is managed by the operative system as a time-out error

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•DEVICE ADDRESSING (when ATN true)

The 5 low bits constitute a primary address (MSb always 0, parity check)

Bit 5 indicates Listener address

0	T	L	4	3	2	1	0
---	---	---	---	---	---	---	---

Bit 6 indicates Talker address

UNT = Disable talkers = 3F; UNL = Disable listeners = BF

*A secondary address can be conceived to address sub-units of a generic device.
Normal addressing followed by a secondary address with bit 5-6 set to 11*

Commands

The following commands control the bus operations and can be grouped in 4 classes:

Universal (Multiline, to all the peripherals with handshake): DCL, LLO, SPE, SPD, PPU, UNT, UNL

Uniline (to all periph. without handshake): IFC (Interf. Clear), REN (Remote En.), ATN (Attent.)

Addressed: GET, SDC, GTL, PPL, TCT

Secondary: PPE, PPD

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Multiline Command	Mnemonic	Action
Device Clear	DCL	All the instruments are brought in a predefined state by the manufacturer, that depend on the instrument and specified in the datasheet; the device must be in remote mode or previously addressed.
Local Lockout	LLO	Disable the control “return-to-local” for the device (in practice disable the instrument front panel). The device can be addressed in local or remote mode. To enable again the local “return-to-local” control the physical line REN must be brought to the logic value 0: in such a way the device is set for local controlling (enabling from the front panel).
Serial Poll Enable	SPE	A Serial Polling of all the devices that can be addressed as Talker is done; each one provides a status-byte, that represents with a suitable encoding the situation (service request, available data, errors, ...). Used to debug.
Serial Poll Disable	SPD	Terminates the Serial Poll mode while resets the devices in Talker mode, so they provide device-dependent information or other status generic information.
Parallel Poll Unconfigure	PPU	Terminates the peripheral parallel poll for diagnostic that is similar to the serial one but faster (each peripheral returns 1 bit) although with less information
Untalk	UNT	Disable the current Talker device. A similar result can be obtained by selecting another device as Talker.
Unlisten	UNL	Disable all the Listener devices since it is not possible to disable a single Listener; this command can be conveniently employed to disable all the listener devices, before enabling all those devices that we want to set as listener.

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<i>Addressed Command</i>	<i>Mnemonic</i>	<i>Action</i>
Group Execute Trigger	GET	All the Listener devices receive the command to start the action previously defined for each of them. The GET command allows to simultaneously synchronize many devices.
Selected Device Clear	SDC	All the devices addressed as Listener are brought back to a device-dependent pre-defined state. It is equivalent to the DCL command after the destination has been addressed
Go To Local	GTL	All the devices addressed as Listener are brought back in local control mode (exit from remote mode). The device will come back to the remote function if addressed as Listener with REN line 1.
Parallel Poll Configure	PPC	All the devices addressed as Listener are configured as the command following PPC, waiting for a parallel poll.
Take Control Talker	TCT	The “controller in charge” enable the current Talker device so as it assumes the role of Bus (Active) controller.

<i>Secondary Command</i>	<i>Mnemonic</i>	<i>Action</i>
Parallel Poll Enable	PPE	Configures all the devices that received the PPC command so as they can answer to the Parallel Poll on a particular data line (DIO), with a certain pre-defined value.
Parallel Poll Disable	PPD	Disable the devices that received the PPC command so as they do not answer to the Parallel Poll (end of Parallel Poll).

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Serial Polling

A service request (SRQ) allows to the controller to receive info on the device state, but it allows also to the controller to understand if one or more devices asks for service.

Procedure: the ATN signal is activated (bus in command mode), the command SPE is sent, all the devices are sequentially checked, each of them returns its status byte, then the command SPD is sent eventually together with UNT.

Parallel Polling

The same as before concerning the status of a device. The Controller activates ATN, then sends the PPC command to each device, assigns to each one an I/O line for answering through the PPE command and sends a EOI (data request). Then it reads lines so identifying the device that requests the service and asks for a status byte to understand the requirement to fulfill.

Sometimes shared lines to provide logic AND/OR of the status of the sharing devices

The answer from a device must arrive within 200 nsec. The controller must read this answer within 2 μ sec after the 200 nsec.

Eventual disable with con PPD and end with PPU

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SERIAL POLL

Advantages:

The controller achieves more information on the interrupt generation cause.

Drawbacks:

The time for a poll cycle augments with the number of connected devices.

PARALLEL POLL

Advantages :

The controller can poll up to 8 devices at the same time.

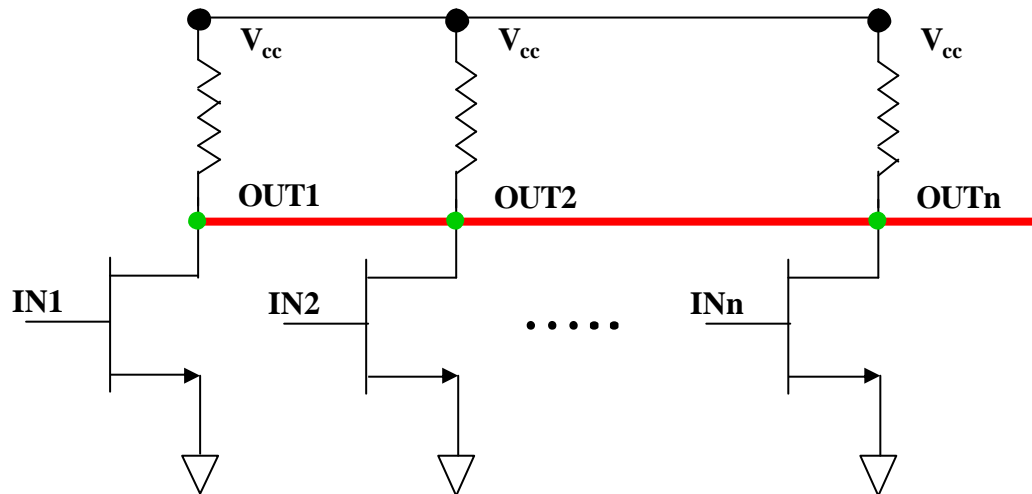
Drawbacks:

The status byte must be read to identify the interrupt cause.

Up to 8 devices can be parallel polled.

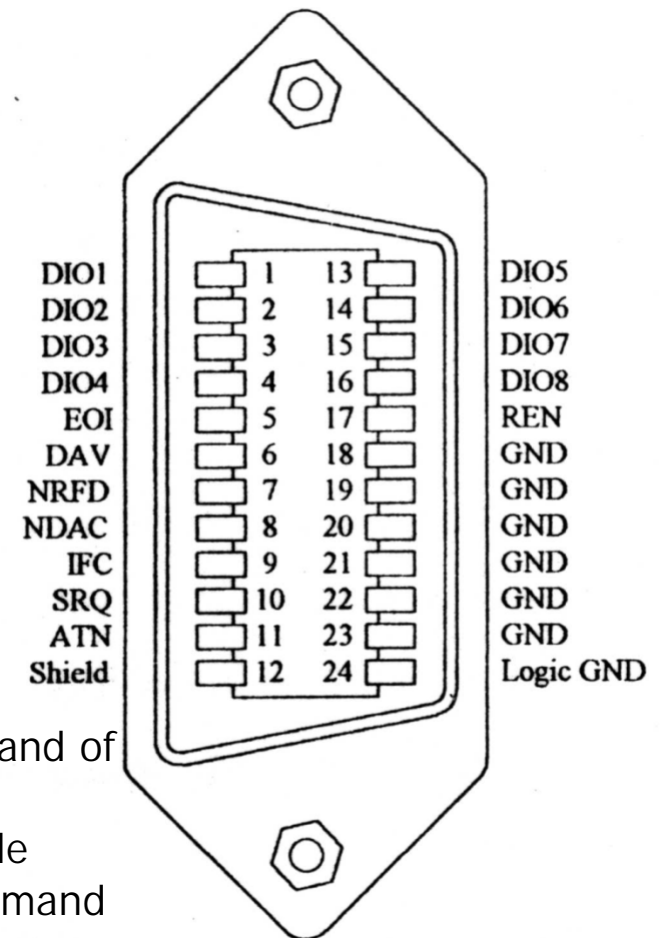
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- **PARALLEL COMMUNICATIONS:** collector wired or connection



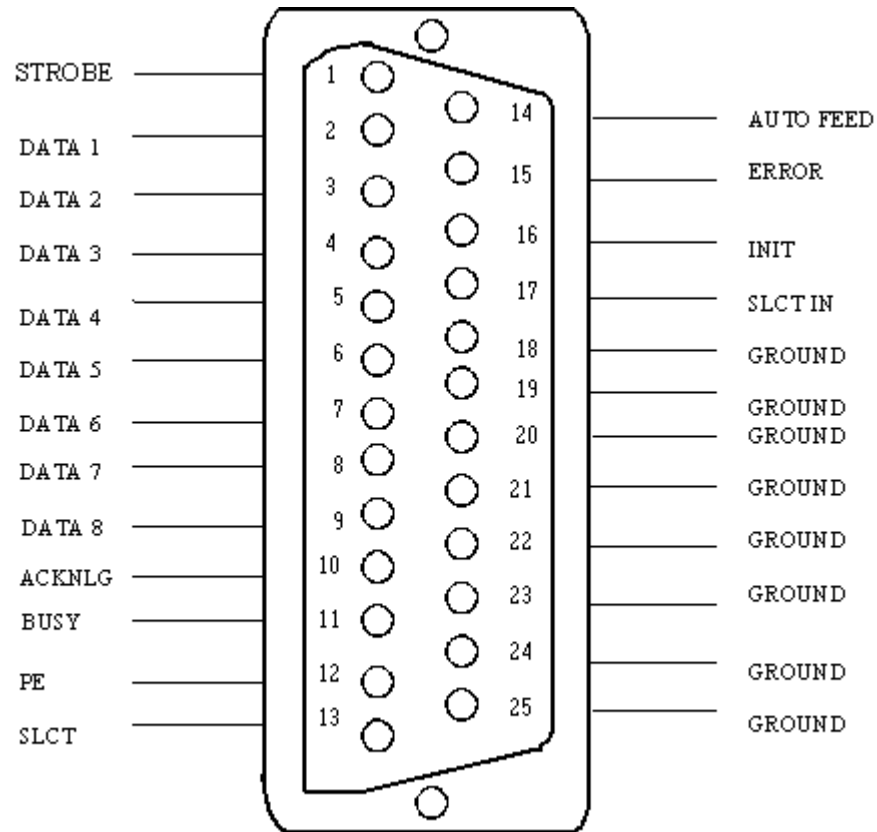
Automatic adjustment to the slowest device

IEEE 488.1 defines the standard from the electric, mechanical and of protocol point of view. The 488.2 version further enriches the standard by specifying data format, error management, possible controller functions, type of status information and typical command to be assigned to instruments



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- **PARALLEL COMMUNICATIONS** : Centronics interface



- Standard interfacing between PC and old printers
- 8 data lines and 9 control ones
- Output towards PC on a Cannon 25 poles jack
- Bi-directional half-duplex connection

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•SERIAL COMMUNICATIONS

A single signal bringing a sequence of bits one at a time

Short distances but more than parallel communications (up to hundreds of meters)

Transmission velocity expressed in symbol or bit per second (baud rate):

- 75?600 low standard velocities no more used

- 1200?9600 low standard velocities still used (PLC, ?P)

- 19200?56400 average standard velocities

- 100?1000 Kbit/sec average standard velocities for field bus

- 10?1000 Mbit/sec high velocities for local LAN

Characterizing issues:

- bit, character, message synchronization techniques

- signal values, codification and eventual modulation

- error management

Existent standards: RS232 standard and successive versions and USB

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•SERIAL COMMUNICATIONS: bit synchronization

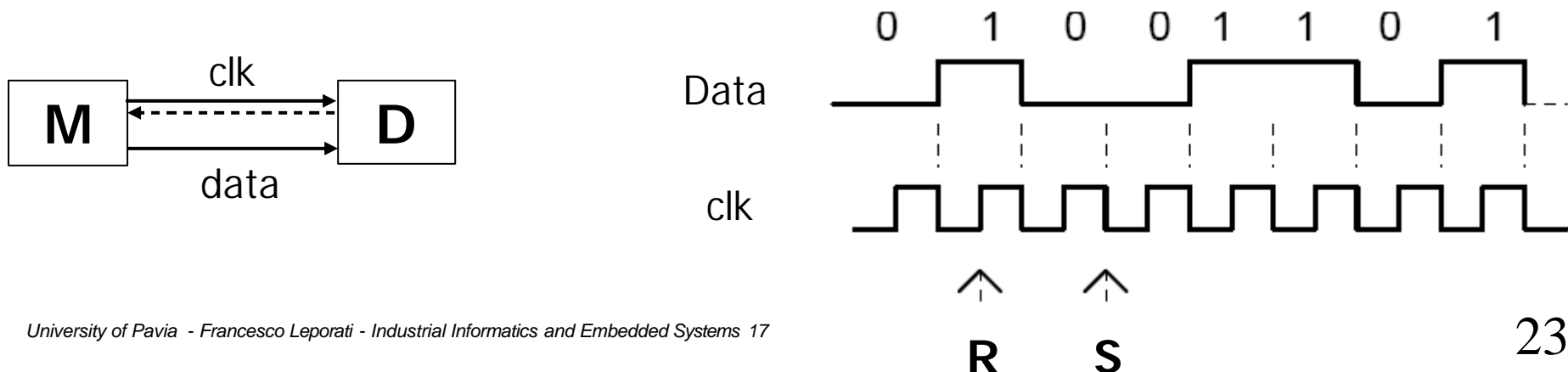
In what "time window" the receiver can sample and thus can read the voltage levels of the received signal

3 techniques: isochronous, asynchronous (mainly used with different manufacturers equipments) and synchronous (especially in field buses and local networks)

THE ISOCHRONOUS TECHNIQUE

A single clock signal shared between sender and recipients

The commutation edge on which the sender writes data is established conventionally (for instance falling edge) of course on the other edge the recipient will sample the line



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•SERIAL COMMUNICATIONS: THE ISOCHRONOUS TECHNIQUE

Synchronization characters are used

Precision in the pulses duration not required (edges determine the sample time)

If clock is generated by sender clock and data are in phase since they travel together

If clock is generated by the recipient it is not in phase with data. The recipient will receive data with a delay twice the time needed to propagate the clock toward the sender and then back ? the clock should be sized at least = to this time.

Limitations in velocity ? short connections (also to avoid to collect noise)

RS232 mode a typical example

In another version a unique conditioned clock is used:

- ✍ a open collector signal between sender and recipient
- ✍ the recipient keeps the signal low until not ready, then it raises up
- ✍ when the sender detects the signal high starts to transmit and keeps low the signal
- ✍ Philips I²C bus

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•SERIAL COMMUNICATIONS: THE ASYNCHRONOUS TECHNIQUE

Sender and recipient with their own independent clock regardless one or the other

Similar stable clocks (tolerance 1...3%) with standard values (300, 600, 1200, 2400, 4800, 9600Hz)

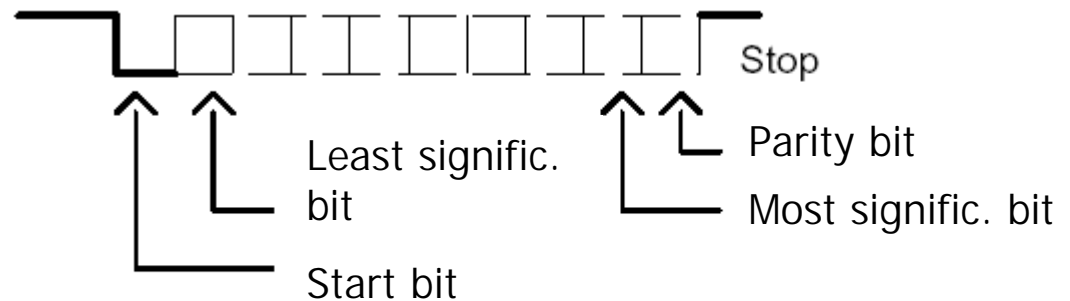
Byte synchronization through start (0) and stop (1) bits with duration 1, 1.5, 2 bit

Message synchronization through silence intervals (1) between a character and the next

Allowed phase shift ? half bit window time

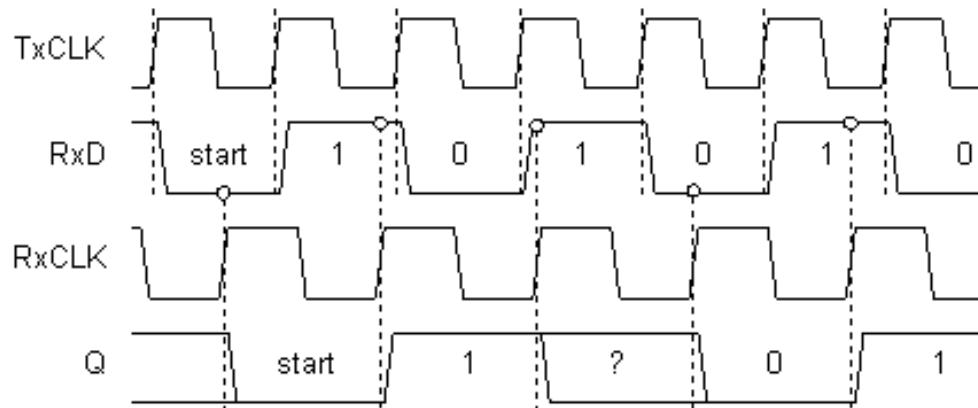
Each byte is synchronized when the first bit is received

Velocity in the order of Kbit/sec
(mouse, printers, PLC)



Asynchronous transmission with 8 bit datum and parity

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TxCLK = sender internal clock used to synchronize serial data (RxD)

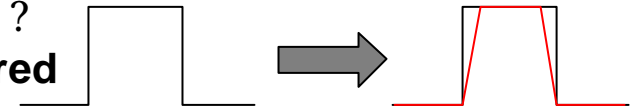
RxCLK = recipient internal clock, supposed slightly slower than TxCLK

Q = RxD bits as they arrive internally to the recipient

The two clocks feature a progressive phase shift that causes uncertainty on the value of the II bit (0) and the loss of the III bit (1). Conversely, if the sampling would be done very close to the beginning of each bit time window and if the sender clock frequency would be slightly minor than the recipient's one, the recipient could sample the same bit twice.

Therefore, the recipient clock must be re-synchronized at every start bit. The active edges (i. e. raising) will be located as much as possible in the center of the window time, so as to sample the RxD line as far as possible from those times into which the commutation can happen. This allows to tolerate slight differences between sender and recipient clocks.

The acceptable phase shift on 10 bit (8 data + start and stop) = ?
5% . This value diminishes when the edges cannot be considered as instantaneous (high velocities and limitation in the band.



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•SERIAL COMMUNICATIONS: THE SYNCHRONOUS TECHNIQUE

The sender features a free clock similar (but not equal) to the recipient: to cancel the phase shift a suitable PLL mechanism is provided PLL (phase lock loop)

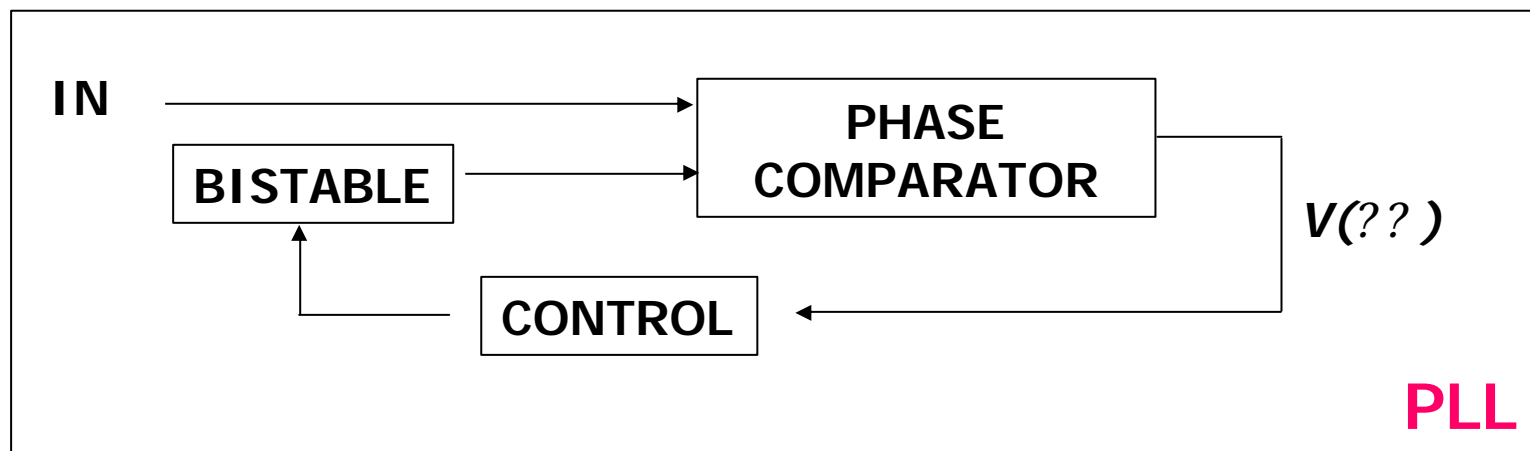
Synchronization bits are transmitted at first (from 3 to 5)

Suitable characters must be chosen to favor the phase coupling (AAAA ko, A6A6 ok)

Avoid long "1s" or "0s" sequences that may vanish the phase coupling

More complexity but more efficiency (no start and stop bits) and velocities

The PLL mechanism compensates eventual jitters on the clock



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BASE BAND ENCODING

NRZ (Not Reduced Zero): The signal level is kept constant throughout the bit time. It may cause synchronization problems if long sequences of equal bits should be sent.

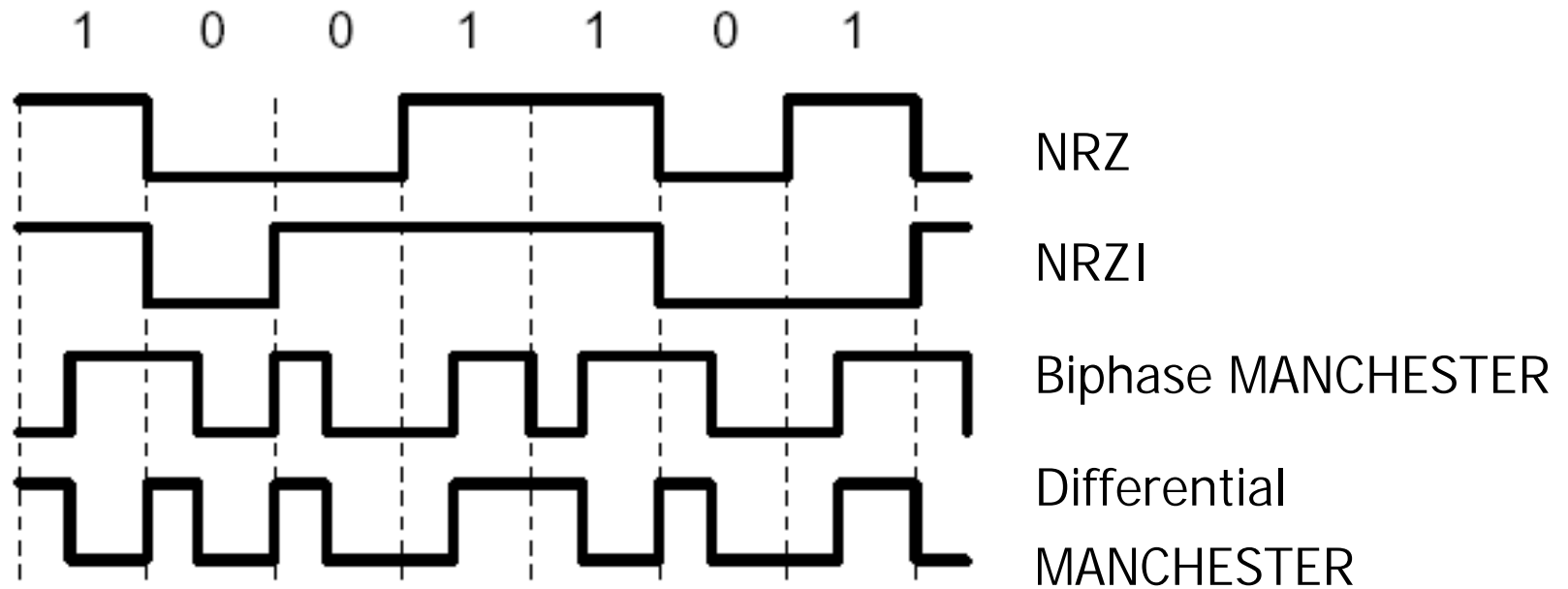
?RZ (Reduced Zero): The signal level is kept high only for a bit time fraction. Thus, the synchronization problems in case of long sequences of “1s” (high) is solved (but not in case of “0s” (low)).

?MANCHESTER: The basic idea is to divide the bit time in two portions equally long. Each bit will be sent as the sequence of two NRZ coded bits each in half of the time window. I. e. the bit 1 is transmitted as a 0 bit and 1 bit while the bit 0 is transmitted as a 1 bit + a 0 bit. This halves the transmission efficiency that is per each bit to transmit, indeed two are sent. In practice a bandwidth is required twice as much as it would be necessary. It is pretty used in local Ethernet and Token-Ring networks.

?NRZI (Non Reduced Zero Inverted): It solves the long “1s” sequences problem. The value 0 is coded maintaining the signal level throughout the period. This level corresponds to the one of the last transmitted bit. On the other hand, the “1” causes a level transition (from low to high or vice-versa) compared to the value of the last transmitted bit. Conversely if the transition level is due to a “0” it solves the problem caused by the transmission of long sequences of “0s”. Used in FDDI networks.

?Differential MANCHESTER: It joins Manchester with NRZI encoding.

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BASE BAND BINARY ENCODING

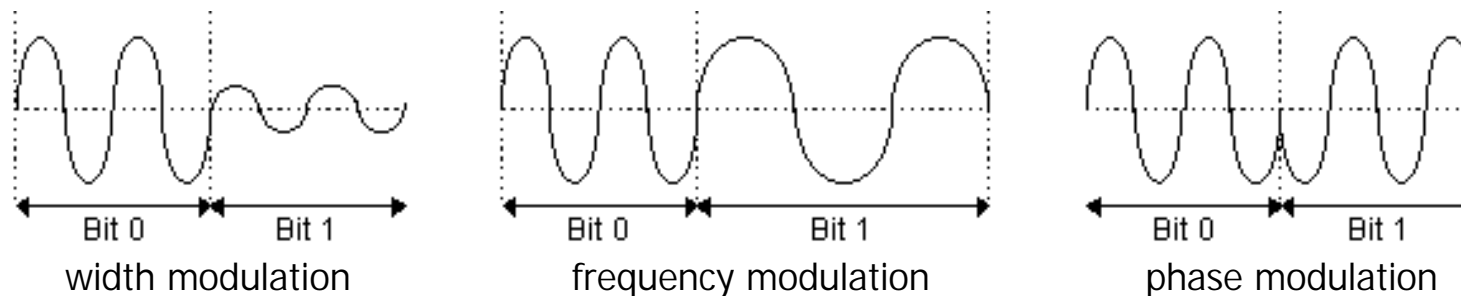
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•SINE SIGNALS ENCODING

Width, frequency, phase possible modulation

Used when communicating with modems on telephone lines, radio link or coaxial cables

The “base band” signal is provided to the modem which carries out the modulation



•POSSIBLE SIGNAL RANGES

TTL level (0V ? 5V/3.3V)

bipolar voltages (-6V ? 6V): RS 232

Differential voltages: RS 485

“current loop” 0-20 mA

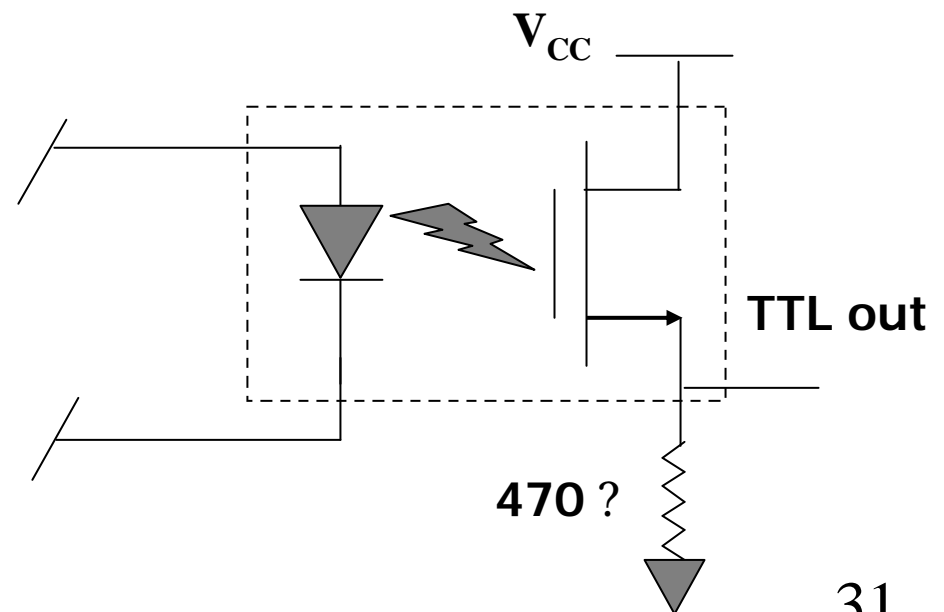
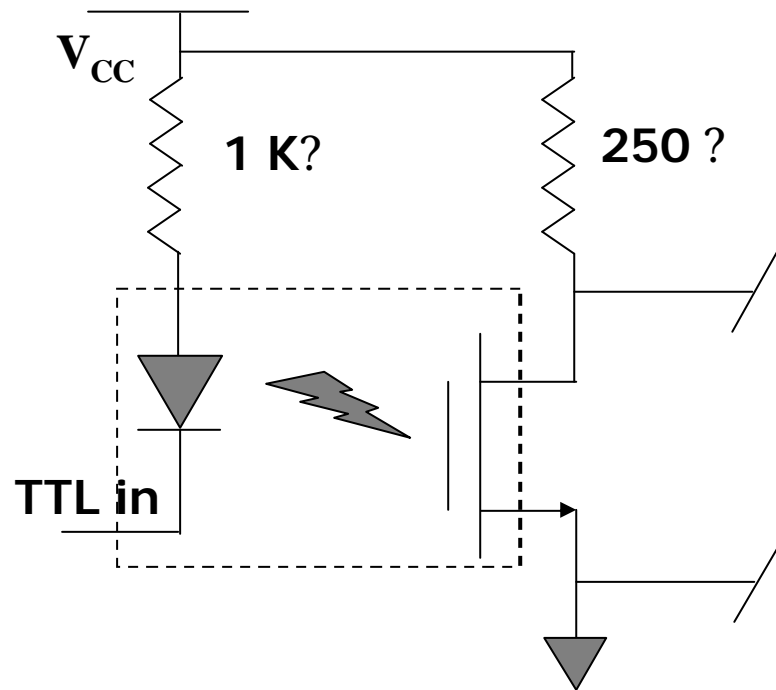
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•Current loop 0-20 mA

low baud rates (2400)

High noise immunity, miles allowed length (ok for industrial environment)

The transmitted signal is current (20 mA) not voltage



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- **Errors**

Recognition

- Error on a character: parity bits or autocorrecting Hamming code (corrects errors on M bits while recognizes them on N with $N < M$)
- Errors on messages: additional characters are present made up by Cyclic Redundancy Check evaluated on 16 bits and sent as 2 characters or checksum or longitudinal parity

Correction

- Hamming codes and CRC require hw complexity (i. e. devoted circuits) and redundancy. Used only in critical applications
- Re-transmission
- The message length depends on the foreseen error rate: if a re-transmission is required also this one could be eventually affected by the error thus the problem will be not solved and the communication cost will be very high

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The Cyclic Redundancy Check associates bit streams to polynomial with 0 and 1 coefficients. A K bit string is read as a polynomial with K coefficients with degree $\leq K-1$. For instance:

$$110001 \rightarrow x^5 + x^4 + x^0$$

The used arithmetic is *module 2*, without carries and borrows. Transmitter and receivers must agree on a $G(X)$ common polynomial, said *generating polynomial*. The idea is to integrate the sequence with a suitable *checksum* making the final polynomial divisible by $G(x)$. For instance:

- add r bit set to 0 to each bit frame so having a $m+r$ bit sequence that corresponds to the polynomial $x^r \times M(x)$, with r rank of $G(x)$, $M(x)$ is the polynomial corresponding to the original frame and m the bit number.
- $x^r \times M(x)$ must be divided (module 2) by $G(x)$: $x^r \times M(x) = G(x) \times Q(x) + R(x)$ the coefficients in $R(x)$ (remainder) will constitute the bit sequence of the checksum.
- Now, the remainder $R(x)$ is added (module 2) to $x^r \times M(x)$ achieving the new bit frame that must be transmitted provided also with the right checksum.

The generator of cyclic codes cannot be a whatever polynomial. Generally speaking it is possible to obtain the same $R(x)$ starting from different $M(x)$ dividend polynomials. All those polynomials that give the same $R(x)$ when divided by the same $G(x)$ generator are defined as *congruous module $G(x)$* . This approach does not detect those errors that transforms the transmitted bit sequence in a different one that however replicate a congruous polynomial. In any case it is a very efficient method for error recognition.

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•RS 232 C STANDARD INTERFACE (1969)

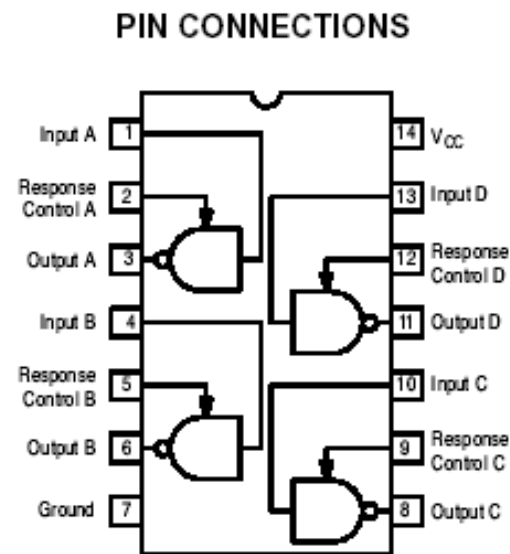
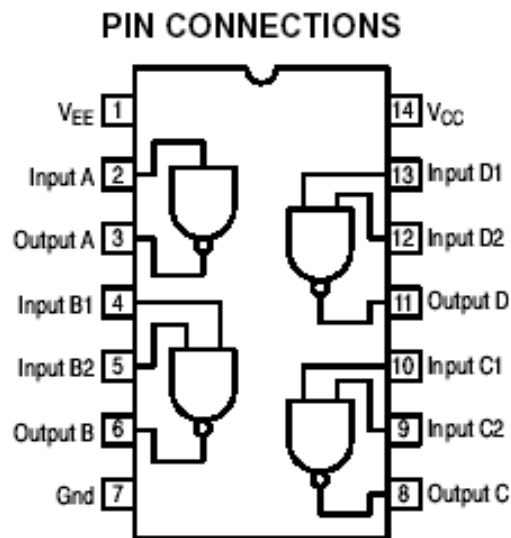
Cannon connector with 25 or 9 poles

Allowed length from few meters until tenths

Negative logic voltage levels: "0" (3V ? 25V) / "1" (-3V ? -25V). Usually +12 V e -12 V are used. Wide noise immunity range but not ground reference

TTL signals must be converted in voltage levels compliant with RS232 values: line drivers and receivers must be employed

**Line driver
1488**



**Line receiver
1489**

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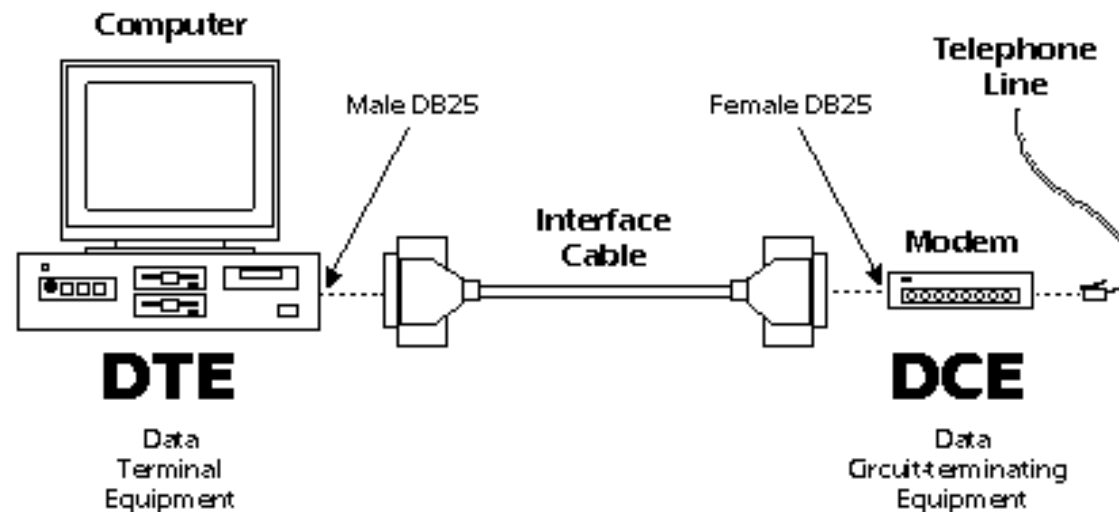
•RS 232 C STANDARD INTERFACE (1969)

Proposed for the first time in '60s by Electronic Industry Association (EIA) and then upgraded several times (last in '97) but still used in industrial manufacturing

The connection model is among so-called Data Terminal Equipments (DTE) (typically computers connected through modems) and DCE - Data Communication Equipments (peripherals connected through modems)

Half duplex and full duplex communications

Null modem version as direct connection between DTE devices



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•RS 232 C PINOUT

1-7: **Signal ground (AB)**

2: **Transmitted data (TxD)**: data output

3: **Received data (RxD)**: data input

4: **Request-to-send (RTS)**: DTE ready to transmit (out in DTE and in DCE)

5: **Clear-to-send (CTS)**: DCE ready to receive (out in DCE and in DTE)

6: **DCE ready (DSR)**: modem ready to start communication

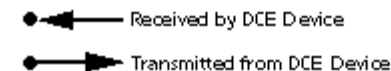
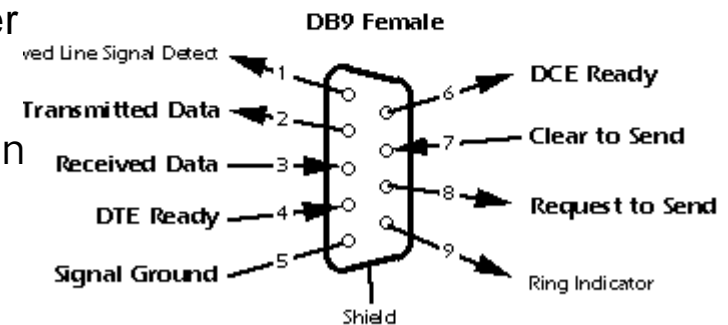
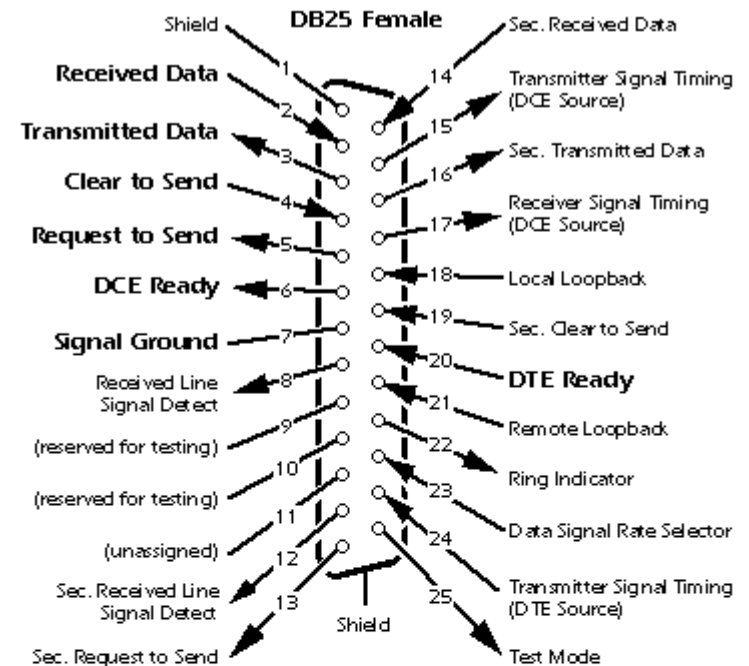
8: **Received line signal detector (CD)**: a remote transmitting modem is detected

15: **TXC** (Transmit Clock): clock generated by the transmitter (in case of isochronous synchronization)

17: **RXC** (Receive Clock): clock generated by the receiver (in case of isochronous synchronization)

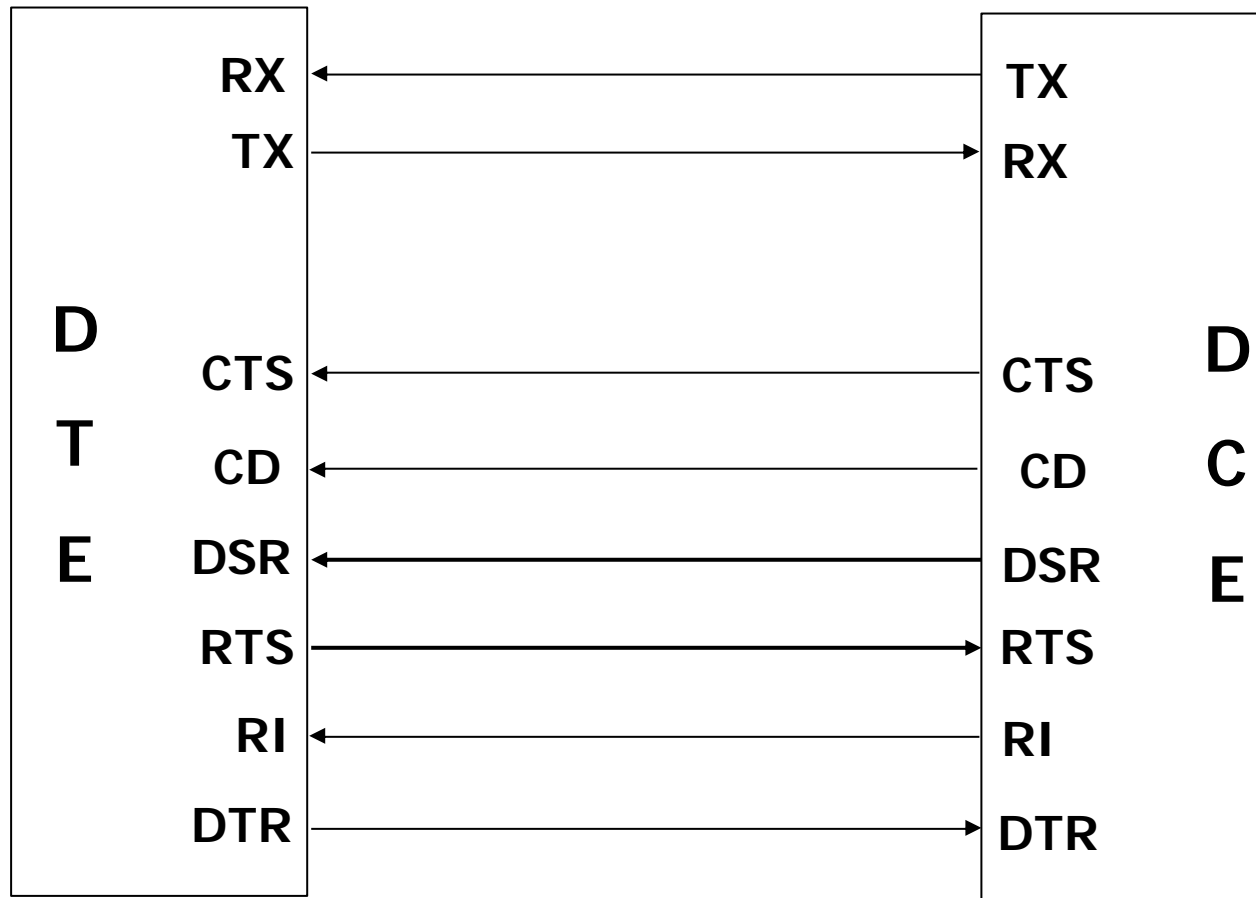
20: **DTE ready (DTR)**: DTE ready to receive

22: **Ring Indicator (RI)**: modem alerts DTE about a call



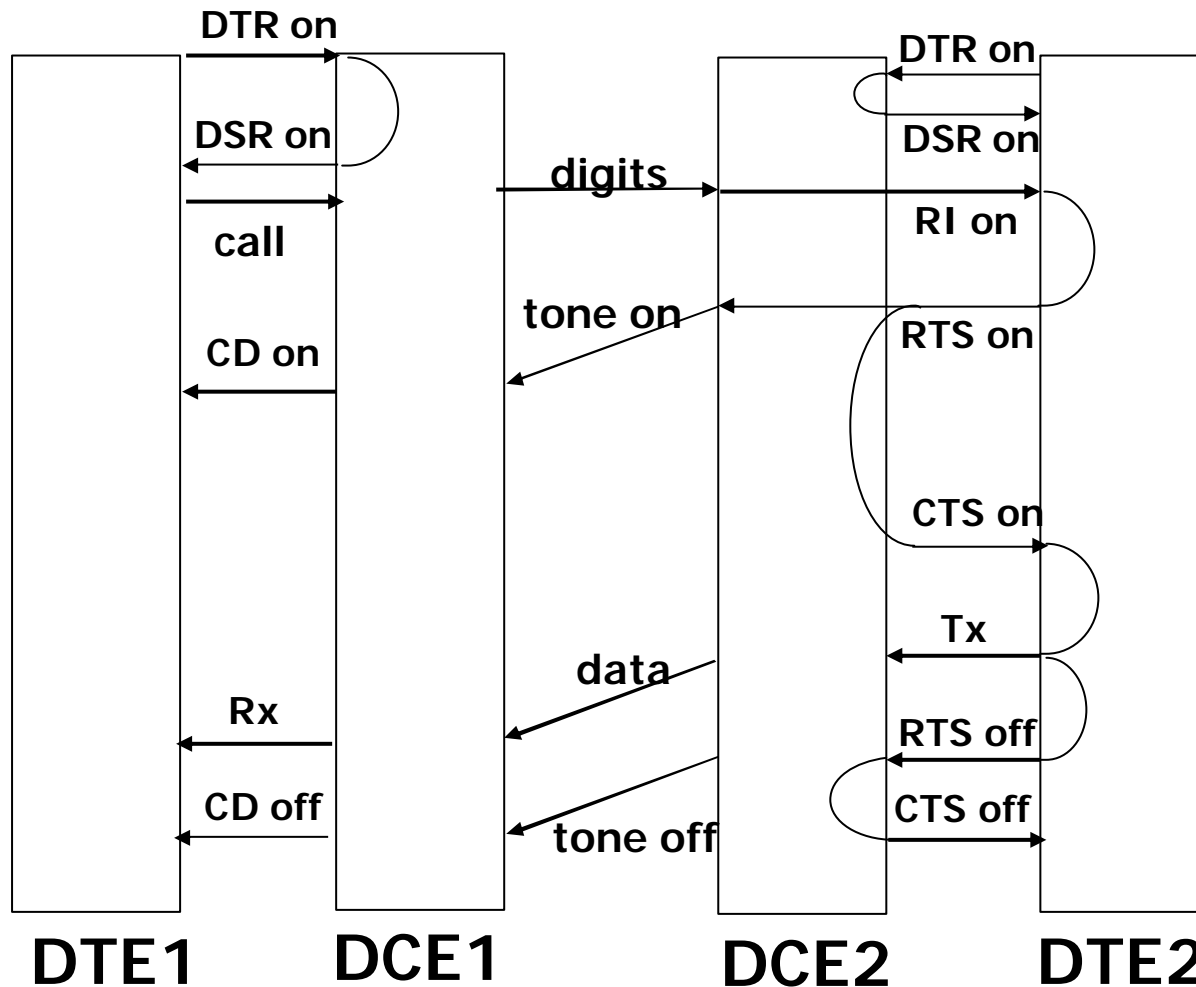
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- DTE-DCE connection



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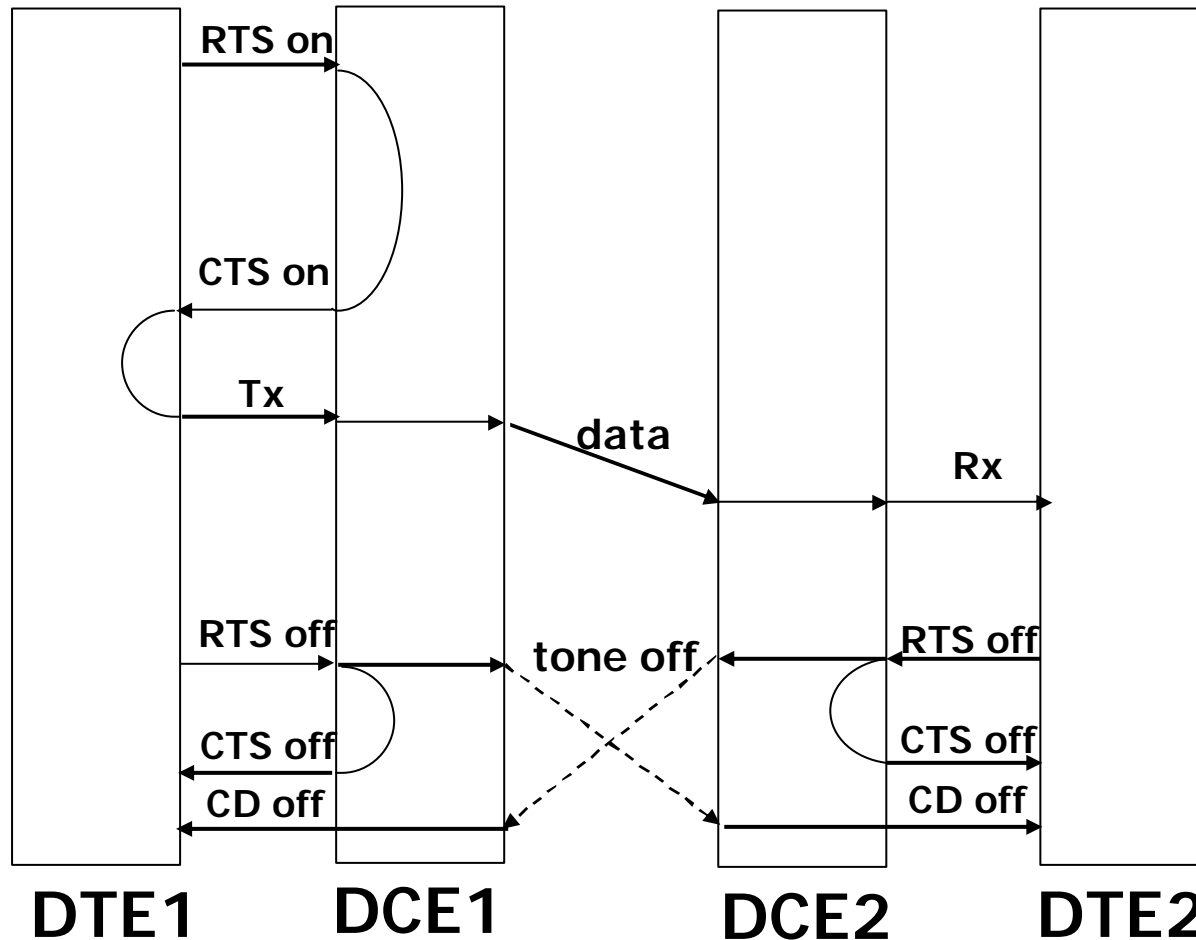
•Setup call



- Half duplex mode
- DTE2 available
- DTE1 calls
- DCE2 alerts DTE2
- DTE2 confirms (handshake RTS-CTS)

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•Data transfer and end of the call



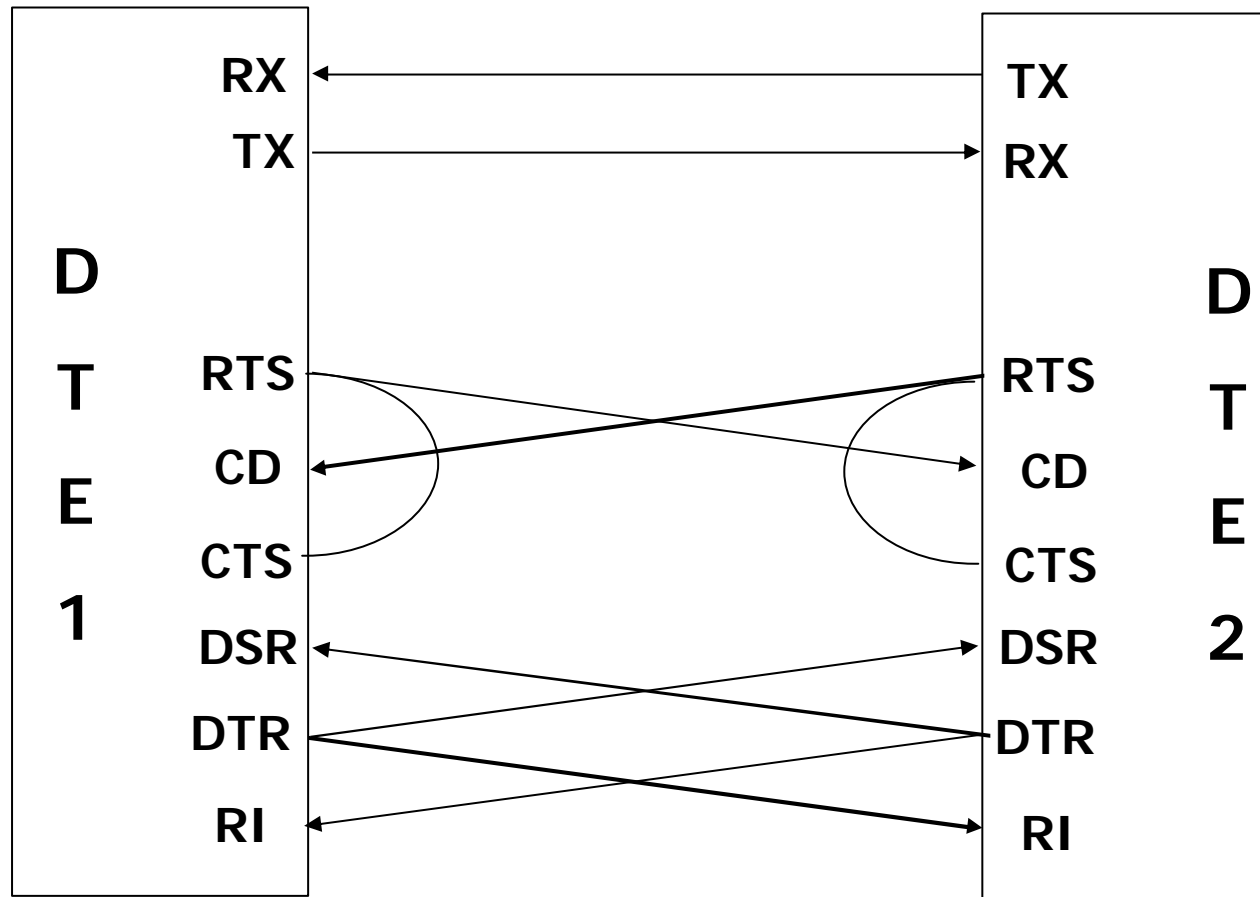
- DTE1 answers (handshake RTS-CTS) ...

- ... and sends those data of which the communication consists

- At the end, one of the DTEs closes the connection with a closing handshake so terminating the communication session

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• Null-modem half duplex connection (if DTE close)



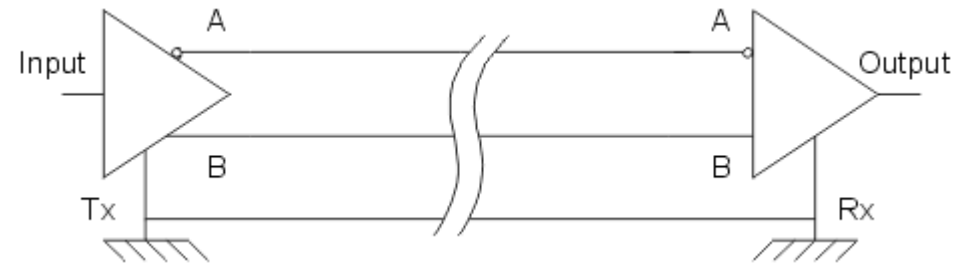
• On the contrary, if full duplex is selected only DTR/DSR/CD and not RTS/CTS are used. DTE1 verifies if DSR of DTE2 is active (ready to work). If yes, DTE1 samples the CD pin to verify the eventual presence of a carrier signal on the line (the partner is trying to communicate) otherwise it starts to communicate.

• If the isochronous temporization is selected pins 15 (TXC) and 17 (RXC) must be used. The transmission clock is used to generate data in isochronous mode, travels with them and will be brought on the RXC pin of the recipient.

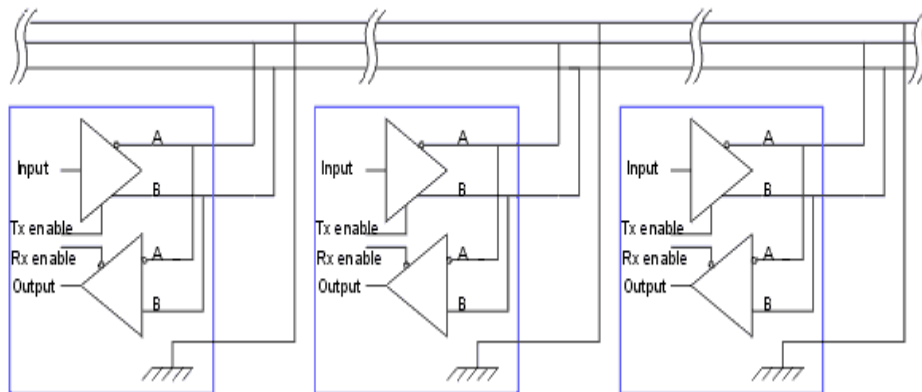
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•RS232 final considerations

- Popular variants of the RS232 are RS422 and RS485.
- In the first case velocities up to 10 Mbit/sec are achieved on distances close to the Km.
- The transmission is in "differential" mode that is using voltage levels slightly different from 232 ($A < B$ "1" e $A > B$ "0") therefore level converters must be introduced.
- Up to 10 recipients connected to a single transmission driver are allowed.



RS422 link schematization



RS485 link schematization

- The RS 485 is very similar to the 422 from the electric point of view, but allows the coexistence of several couples Tx/Rx on the same transmission line (multi-point), although only one transmitter at a time can be active.
- Up to 32 recipients are allowed.
- Velocities within 100 Kb/sec and 10 Mb/s.
- Distances within 1200 m and 15 m.
- Used when connecting PC, PLC and instrumentations in a bus topology.

DIGITAL COMMUNICATION TECHNIQUES

Characteristics of RS485 compared to RS232, RS422 and RS423

	RS232	RS423	RS422	RS485
Differential	no	no	yes	yes
Max number of drivers	1	1	1	32
Max number of receivers	1	10	10	32
Modes of operation	half duplex full duplex	half duplex	half duplex	half duplex
Network topology	point-to-point	multidrop	multidrop	multipoint
Max distance (acc. standard)	15 m	1200 m	1200 m	1200 m
Max speed at 12 m	20 kbs	100 kbs	10 Mbs	35 Mbs
Max speed at 1200 m	1 kbs	1 kbs	100 kbs	100 kbs
Max slew rate	30 V/ μ s	adjustable	n/a	n/a
Receiver input resistance	3..7 k?	4 k?	4 k?	12 k?
Driver load impedance	3..7 k?	450 ?	100 ?	54 ?
Receiver input sensitivity	± 3 V	± 200 mV	± 200 mV	± 200 mV
Receiver input range	± 15 V	± 12 V	± 10 V	-7..12 V
Max driver output voltage	± 25 V	± 6 V	± 6 V	-7..12 V
Min driver output voltage (with load)	± 5 V	± 3.6 V	± 2.0 V	± 1.5 V

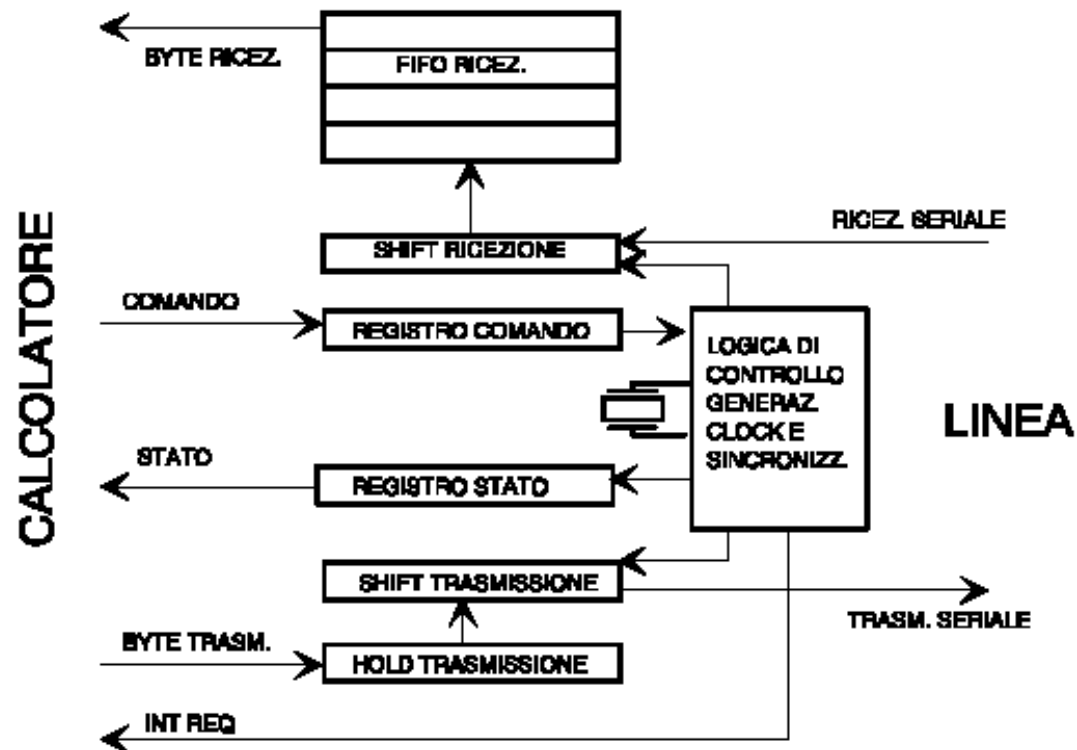
DIGITAL COMMUNICATION TECHNIQUES

USART (Universal Synchronous Asynchronous Receiver Transmitter)

- IC carrying out synchronous/asynchronous serial transmission
- Sometimes embedded within the microcontroller ? single chip computer
- UART if the transmission is only asynchronous
- The device implements all the transmission/reception task, temporization, parity check ..., while the voltage level adaptation is done through suitable driver
- Basic structure made up by:
 - transmission and reception buffers
 - transmission and reception shift register
 - mode, command, state registers
- COTS standard devices: NS8250 e NS16550

DIGITAL COMMUNICATION TECHNIQUES

USART : block scheme



A character put in the *transmission register* triggers the beginning of the transmission

The reception is performed by reading the *receiving register* where data are valid if suitable error flags are not active in the *status register*

The *state register* bits indicate:

- if synchronization has been done
- if the transmission buffer is empty
- if transmission shift reg. is empty
- if a character is in reception buffer
- errors (parity, framing, overrun)
- interrupt

Command reg.: reset state indicators, enable interrupt, set line signals (RTS ...

Mode reg.: set of the comm. velocity, parity, synchronization characters

DIGITAL COMMUNICATION TECHNIQUES

UART 8250: registers

8 registers in 8 consecutive address areas of the I/O memory space (base address = 3F8H for Com1 and 2E8H for Com2 ports)

some registers can be used in **sole reading** mode or in **sole writing** mode if the bit DLAB in LCR is zero or in **both the R/W modes** if the bit DLAB=1

Indirizzo	Read	Write	Read	Write
base	RBR (rec. buf.)	THR (transm. hold)	DLL (divisor latch LSB)	
base+1	IER (int. enab.)	IER	DLM (divisor latch MSB)	
base+2	IIR (int. identif.)	FCR (fifo control)	IIR	FCR
base+3	LCR (line control, R/W independently on DLAB)			
base +4	MCM (modem control, R/W independently on DLAB)			
base+5	LSR (line status)	/	LSR	/
base+6	MSR (modem status)	/	MSR	/
base+7	SCR (scratch, R/W independently on DLAB)			

DIGITAL COMMUNICATION TECHNIQUES

UART 8250: registers

DLR = Divisor Latch Register

16 bit registers used to set the baud rate on the basis of the values written in BA e BA+1

Baud Rate	High Byte	Low Byte
110	04H	17H
300	01H	80H
1200	00H	60H
2400	00H	30H
4800	00H	18H
9600	00H	0CH
19200	00H	06H
38400	00H	03H
57600	00H	02H
115200	00H	01H

- b7 . . b4 - Reserved always zero
- b3 - Modem Status Interrupt
- b2 - Receiver Line Status Interrupt
- b1 - Transmitter Holding Register Empty Interrupt
- b0 - Received Data Available Interrupt

IER = Interrupt Enable Register

enable UART to generate interrupt through setting of the correspondent bits

IIR = Interrupt Identification Register

the bit allows to identify the origin of the interrupt service request

b2	b1	What they say	How to reset
1	1	Receiver Line Status	Read LSR
1	0	Received Data Available	Read RBR
0	1	Transm. Hold. Register Empty	Read IIR or write THR
0	0	Modem Status	Read MSR

DIGITAL COMMUNICATION TECHNIQUES

FCR = Fifo Control Register

8 bit register used to setup the Fifo memories present in the 16550 device: in particular b0 enables the Fifo, b1 and b2 empty reception and transmission Fifos, b6 and b7 define the n° of bytes required to set the full Fifo flag (1, 4, 8, 14)

LCR = Line Control Register

8 bit register to set different functionalities

b7 (DLAB) = if 1 allows to access DLR, if 0 to access RBR, THR, IER

b6 = if 1, a constant voltage level corresponding to that of a start bit is put on the line to signal a break of transmission.

b4, b3 = set the parity

b2 = set the number of stop bits (1, 2)

b1, b0 = set the character length (5, 6, 7, 8 bit)

MCR = Modem Control Register

8 bit register used to set the RTS and DTR signal output from the UART:

b2 = global enable of the UART to interrupt

b1, b0 = if 1 each one activates RTS and DTR

DIGITAL COMMUNICATION TECHNIQUES

LSR = Line Status Register

8 bit register providing info about the transmission status

b5 = 1 if the THR register is empty. The corresponding interrupt is activated, if enabled

b4 = 1 if a low level is detected on the line (Break) longer than a character

b3 = 1 if after the character bits and the parity the expected stop bit is not detected (Framing Error)

b2 = 1 if after the character bits the parity is not correct (Parity Error)

b1 = 1 if a character in RBR has been overwritten before being read (Overrun Error)

b0 = 1 if a new, not read character is available in RBR (becomes 0 after reading)

MSR = Modem Status Register

8 bit register used to detect the value of the CD, RI, DSR e CTS (b7-b4) signals or if they were switched (b3-b0)

SCR = Scratch Register

Service register to read/write temporary values

DIGITAL COMMUNICATION TECHNIQUES

USART sw management

Polling

- The LSR is periodically polled: a new received character or a character to be transmitted or error will be signaled
- Low-medium communication velocities and busy waiting

Interrupt

- Each of the three previous situations generates an interrupt request (character reception, character transmission, error)
- IIR is read to adequately serve the interrupt
- Medium transmission velocities (9600 baud)

DMA

- controllers or coprocessors devoted to handshake between memory and USART
- velocities in the order of MB/sec
- Not always provided: only in expensive devices

DIGITAL COMMUNICATION TECHNIQUES

USART sw management

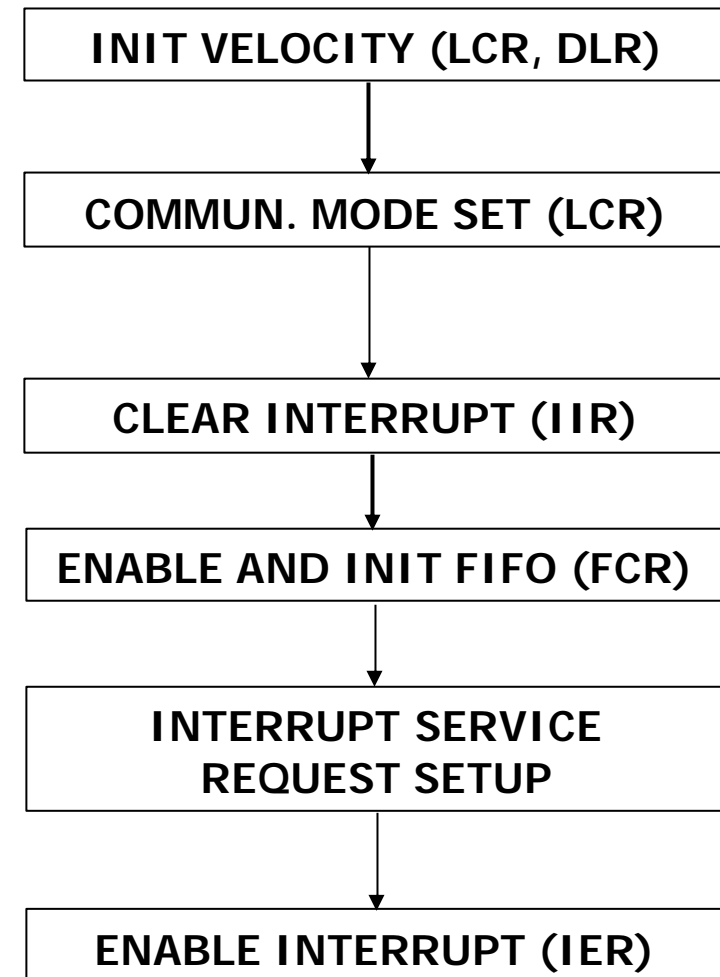
Init registers addresses and main masks for bits

BA	EQU	3F8H	;indirizzo base
RBR	EQU	BA+0	;dato ricevuto
THR	EQU	BA+0	;dato da trasmettere
DLRL	EQU	BA+0	;seleziona baud L
DLRH	EQU	BA+0	;seleziona baud H
B1200	EQU	0060H	;valore per 1200 baud
B9600	EQU	000CH	;valore per 9600 baud
IER	EQU	BA+1	;reg. abilitaz. interrupt
MKEIT	EQU	2	;maschera int. trasmissione
MKEIR	EQU	1	;maschera int. ricezione
IIR	EQU	BA+2	;reg. tipo interrupt
MKI	EQU	6	;maschera bit di tipo
FCR	EQU	BA+2	;reg. controllo FIFO
MKF1	EQU	0	;int. ogni byte
MKF4	EQU	04H	;int. ogni 4 byte
MKF8	EQU	08H	;int. ogni 8 byte
MKF14	EQU	0CH	;int. ogni 14 byte
MKFRT	EQU	04H	;maschera reset FIFO trasm.
MKFRR	EQU	02H	;maschera reset FIFO ricez.
MKFEN	EQU	01H	;maschera abilita FIFO
LCR	EQU	BA+3	;reg. controllo linea
MKDLAB	EQU	80H	
STDLINE	EQU	1BH	;impostaz. standard pari, 1 stop, 8 bit
MCR	EQU	BA+4	
LSR	EQU	BA+5	;reg. stato linea
MKTHRE	EQU	20H	;maschera THRE
MKBRK	EQU	10H	;maschera Break
MKFRE	EQU	08H	;maschera framing err.
MKPAE	EQU	04H	;maschera parity err.
MKOVE	EQU	02H	;maschera overrun err.
MKDRDY	EQU	01H	;maschera dato ricevuto ready
MSR	EQU	BA+6	

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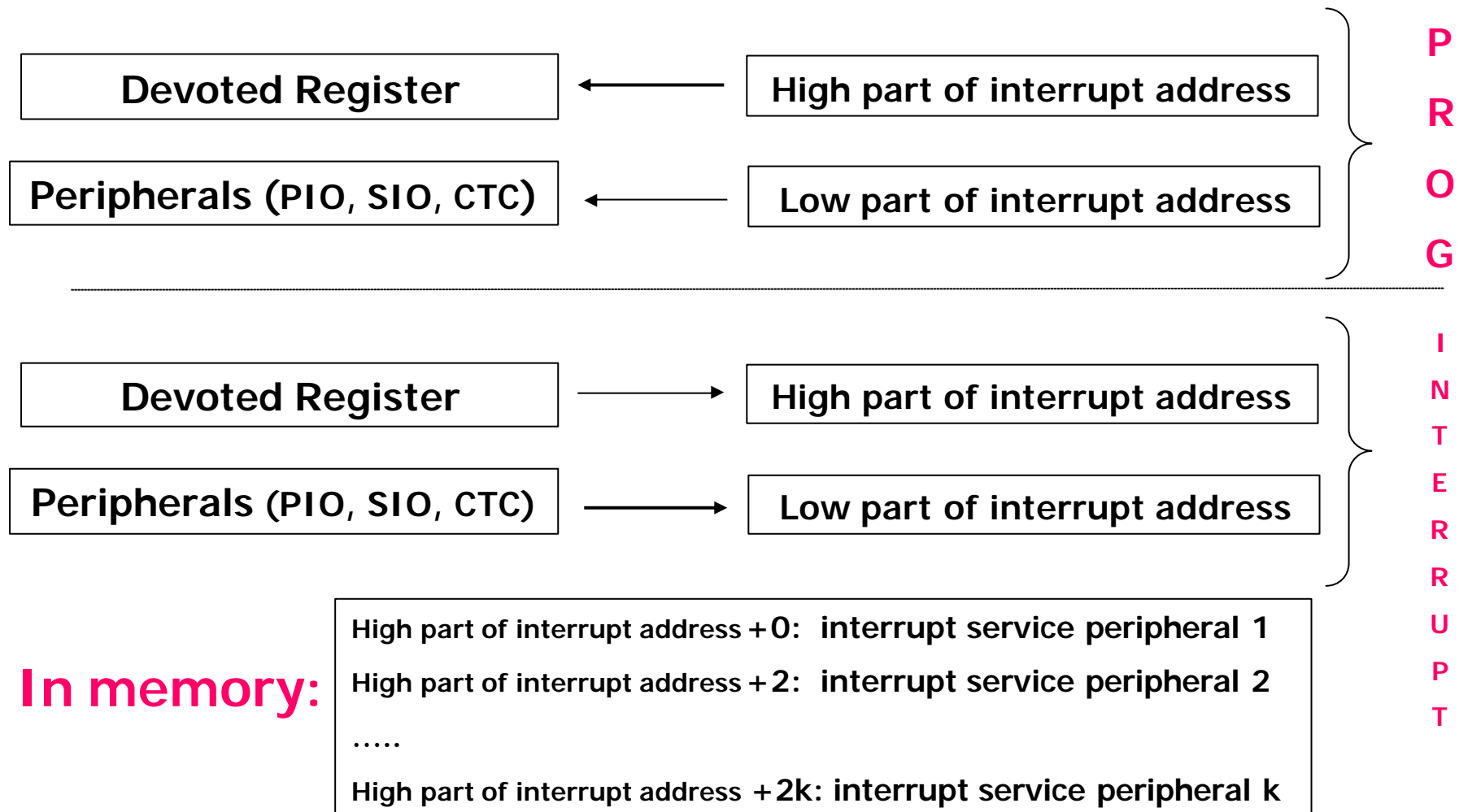
UART setup

```
;imposta velocita' di comunicazione
MOV    AL, MKDLAB
MOV    DX, LCR
OUT    [DX],AL
MOV    AX, B9600
MOV    DX, DLRL
OUT    [DX],AL
INC    DX
OUT    [DX],AH
;imposta modalita' di comunicazione
MOV    AL, STDLINE
MOV    DX, LCR
OUT    [DX],AL
;cancella eventuali interrupt pendenti
MOV    DX, IIR
IN     AL, [DX]
;inizializza la FIFO e la abilita
MOV    AL, MKFRT+MKFRR+MKFEN
MOV    DX, FCR
OUT    [DX],AL
;aggancia la routine di servizio interrupt
CALL   AGGANZIA_INT_UART
;abilita UART a generare richieste di interrupt
MOV    AL, MKEIT+MKEIR
MOV    DX, IER
OUT    [DX],AL
```



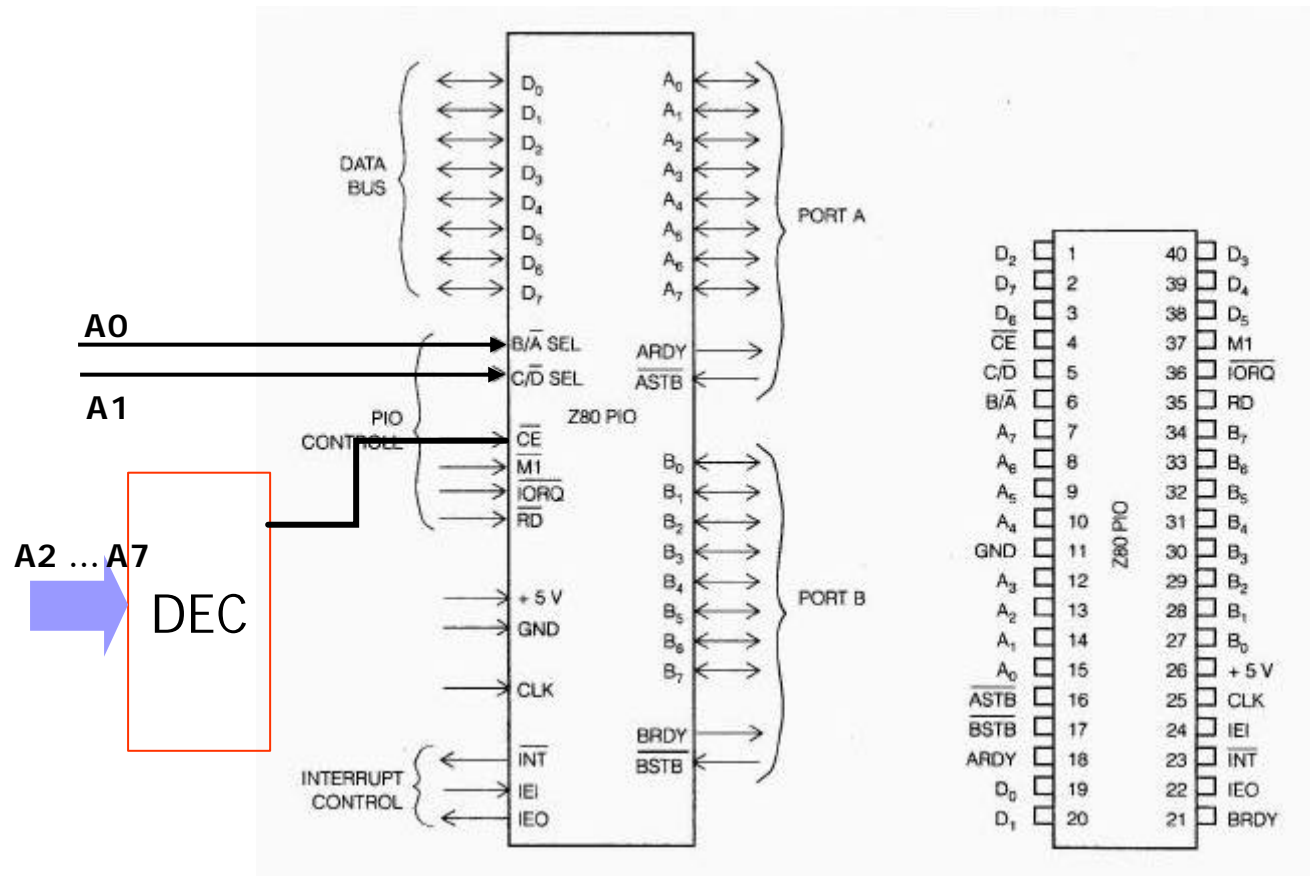
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Aggancia_INT_UART



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- The parallel port of a typical microprocessor



B/A: select port B or A

CE: chip enable

M1: fetch active

IEO/IEI: interrupt en.

A/B RDY: port ready to communicate

A/B STB: peripheral ready to communicate

INT: int. request

IORQ: mem. mapped I/O

C/D: command or data

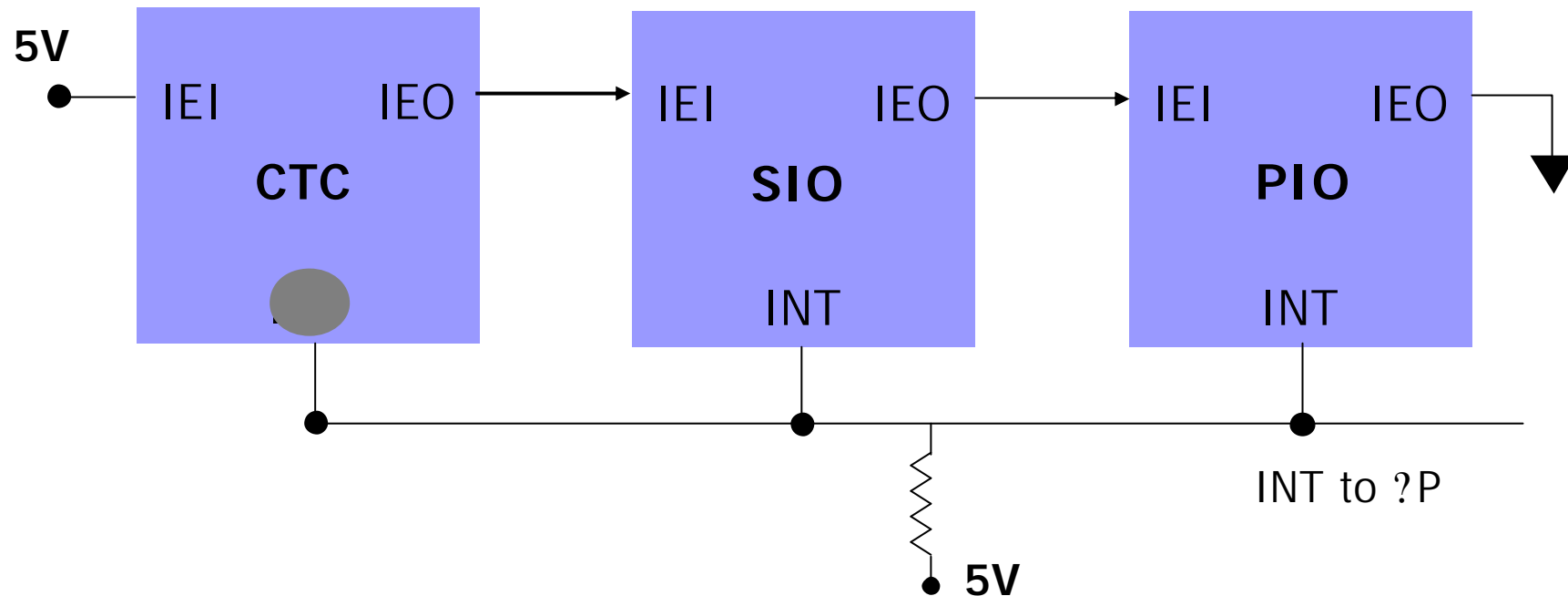
RD: read or write

A0-A7: out port A

B0-B7: out port B

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- A hardware priority chain: *daisy chain* (priority diminishes from left to right)



When a peripheral (parallel, serial, timer) generates an interrupt the ?P can recognize the request and sets M1 and IORQ lines simultaneously at zero (IntAck).

The peripherals, looking two lines at zero, will put on the bus the low part of the interrupt vector (specific for itself). The PIO must be connected with the M1 e IORQ lines to know that the microprocessor acknowledged the interrupt.

-

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•PIO port programming (modes 0, 1, 2)

A port can be configured to work in mode 0 (output only), 1 (input only), 2 (bidirectional) or 3 (bit by bit). This is possible by writing suitable bytes to the port that however must be set in “command” (configuration) mode (C/D=1).

The first three modes require a control word into which the configuration bits setting mode 0, 1 or 2 are coded in bits D7 and D6

M2	M1	X	X	1	1	1	1
----	----	---	---	---	---	---	---

Then the low part of the interrupt vector (address) will be sent to the port (that is the value given back by the port itself on the data bus when it will require an interrupt service).

Finally it will be possible to disable the interrupt for that port by sending the following word with the bit 7 set to 1 (enable) or to 0 (disable).

E/D	X	X	X	0	0	1	1
-----	---	---	---	---	---	---	---

DIGITAL COMMUNICATION TECHNIQUES

•PIO port programming (mode 3)

In mode 3 the same procedure is followed as before that sends a *mode control* word and the interrupt vector to the port.

Then a word is sent that specifies what bits are in input (set to 1) and what in output (set to 0); for instance below the bits 7-6-3-2-1-0 are set as input and 5-4 as output.

1		1		0		0		1		1		1		1
---	--	---	--	---	--	---	--	---	--	---	--	---	--	---

The *interrupt control word* features the following format into which the bit D4 specifies if inputs must be masked or not i. e. if all of them (0) or only few (1) will cause interrupt service request. In this case a mask will follow that indicate what bit will generate the interrupt (0) and what ones must be ignored.

1/0		1/0		1/0		1/0		0		1		1		1
-----	--	-----	--	-----	--	-----	--	---	--	---	--	---	--	---

The bit D5 indicated the interrupt activation edge (0 falling / 1 raising), D6 if the generating condition is the OR or the AND of more bits, and D7 if the interrupt request is enabled or not.

DIGITAL COMMUNICATION TECHNIQUES

NAME MAIN					
4 0000			8 000A	LENGA EQU 0AH ;length tab. prog. SIOA	
6	LIST E,O,S,T		8 000C	LENGB EQU 0CH ;lenght tab. prog. SIOB	
7 0000			8 0020	SIOA EQU 20H	
8	INCLUDE COST.H ; equates		8 0021	SIOB EQU 21H	
8 0030	TABINT EQU 30H ;most sign. byte in the address of interrupt tab.		8 0022	SIOAC EQU 22H	
8 0040	PIO1A EQU 40H		8 0023	SIOBC EQU 23H	
8 0042	PIO1AC EQU 42H		8 00F0	SIOVEC EQU 0F0H ;tab. interrupt SIO	
8 0041	PIO1B EQU 41H		8 00FF	LCD_C EQU 0FFH ;address command/state LCD	
8 0043	PIO1BC EQU 43H		8 00FE	LCD_D EQU 0FEH ;address dat LCD	
8 0060	PIO2A EQU 60H		30 0000		
8 0062	PIO2AC EQU 62H		31		CSEG
8 0061	PIO2B EQU 61H		33 0000 F3		DI
8 0063	PIO2BC EQU 63H		34 0001		
8 0000			35 0001 CD BE 06 C	CALL RITARDO	
8 000F	MOD00 EQU 0FH		36 0004 31 FF F7	LD SP,STAC	
8 004F	MOD01 EQU 4FH		47 0022 ED B0	LDIR	
8 008F	MOD02 EQU 8FH		48 0024 08	EX AF,AF'	
8 00CF	MOD03 EQU 0CFH		49 0025 D9	EXX	
8 0007	DISINT EQU 07H ;disable interrupt		50 0026		
8 0083	LIN2A EQU 83H ;ISOL, REEDB input, ADC -BSY input		51 0026 CD CB 03 C	CALL RETI	;reset hw all peripherals (4)
8 00A7	CNINTP EQU 0A7H ;int. abil., OR, high edge active		52 0029 CD CB 03 C	CALL RETI	
8 00FE	MASK EQU 0FEH ;the line D0 only activates interrupt		53 002C CD CB 03 C	CALL RETI	
8 0000	CTCCH0 EQU 00H ;channel 0 CTC		54 002F CD CB 03 C	CALL RETI	
8 0001	CTCCH1 EQU 01H ;channel 1 CTC		55 0032 ED 5E	IM 2	;vector interrupt mode set
8 0002	CTCCH2 EQU 02H ;channel 2 CTC		56 0034 3E 30	LD A,TABINT	
8 0003	CTCCH3 EQU 03H ;channel 3 CTC		57 0036 ED 47	LD I,A	
8 0007	CTCCN2 EQU 07H ;no int, timer mode, prescaler=16, whatever edge		58 0038		
8	;no trigger, time constant follows, reset		59	; Programming PIO, CTC, SIO	
8 00C7	CTCCN3 EQU 0C7H ;enable int., counter mode, no prescaler, neg. edge		60	; inzialization PIO	
8	;no trigger, time constant follows, reset		61	; *****	
8 0017	SIOCNT EQU 17H ;control for SIO temporization		62 0038 3E 07	LD A,DISINT	
8	;through CTC outputs, no int., timer mode		63 003A D3 42	OUT (PIO1AC),A	
8	;16, positive edge, time constant follows, reset		64 003C D3 43	OUT (PIO1BC),A	
8 00C8	TIME2 EQU 0C8H ;CTC time constant channel 2 (9600 baud)		65 003E D3 63	OUT (PIO2BC),A	;disab. interrupt on PIO1A, PIO2B, PIO1B
8 0028	TIME3 EQU 40 ;CTC time constant channel 3		66 0040		
8 0001	BAUD EQU 1 ;temp. channels 0 and 1 CTC (9600 baud)		67 0040 3E 0F	LD A,MOD00	
			68 0042 D3 63	OUT (PIO2BC),A	

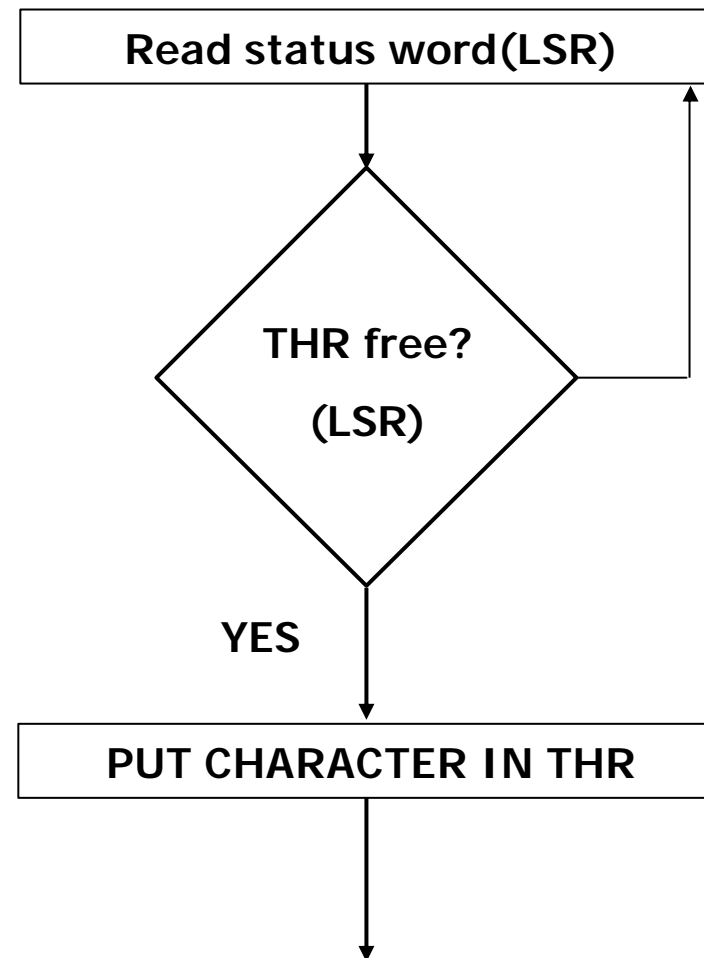
DIGITAL COMMUNICATION TECHNIQUES

69 0044 D3 43	OUT (PIO1BC),A	520 0398 EE 20	XOR 20H ; test bit D5 edge sensitive
70 0046 3E 30	LD A,30H ;READ e WRITE high to do not write spurious data	521 039A D3 62	OUT (PIO2AC),A
71 0048 D3 41	OUT (PIO1B),A	522 039C 32 B1 02 D	LD (PROG2A),A ; to successively program the port 2A
72 004A		523 039F	
73 004A 3E CF	LD A,MODO3	524 039F ED 4B 11 02 D	LD BC,(CIMP) ;load pulse counter REEDB
74 004C D3 42	OUT (PIO1AC),A	525 03A3 3A CE 02 D	LD A,(KEY)
75 004E 3E FF	LD A,0FFH	526 03A6 CB 7F	BIT PWRCLOSB,A
76 0050 D3 42	OUT (PIO1AC),A	527 03A8 C2 AF 03 C	JP NZ,TRY6
77 0052		528 03AB 0B	DEC BC
78 0052 3E E6	LD A,0E6H ; LSB interrupt routine	529 03AC C3 B5 03 C	JP CARICA
79 0054 D3 62	OUT (PIO2AC),A	530 03AF CB 77 TRY6	BIT PWROPENB,A
80 0056 3E CF	LD A,MODO3 ; PIO 3 mode	531 03B1 C2 B5 03 C	JP NZ,CARICA
81 0058 D3 62	OUT (PIO2AC),A	532 03B4 03	INC BC
82 005A 3E 83	LD A,LIN2A ; what input/output lines	533 03B5 ED 43 11 02 D CARICA	LD (CIMP),BC
83 005C D3 62	OUT (PIO2AC),A	534 03B9 D9	EXX
84 005E 3E A7	LD A,CNINTP ; interrupt control	535 03BA 08	EX AF,AF'
85 0060 32 B1 02 D	LD (PROG2A),A ; port 2A re-programmed in interrupt	536 03BB FB	EI
86 0063 C6 10	ADD A,10H ; a mask will follow	537 03BC ED 4D	RETI
87 0065 D3 62	OUT (PIO2AC),A	538 03BE	
88 0067 3E FE	LD A,MASK ; mask for those lines activating interrupt	
89 0069 D3 62	OUT (PIO2AC),A	1078	ORG 30E6H
90 006B		1079 30E6	
91 ; init CTC		1080 30E6 93 03 C	DEFW PPOINT
119 ; init SIO		1081 30E8 12 05 C	DEFW EIRETI
120 ; *****		1082 30EA 12 05 C	DEFW EIRETI
147		1083 30EC 12 05 C	DEFW EIRETI
148 00C6		1084 30EE 69 01 C	DEFW CTCINT
149 00C6 FB MAIN: EI		1085 30F0 CD 03 C	DEFW TRASB
257 0168 C9 RET		1086 30F2 D9 03 C	DEFW EXTB
513 0393		1087 30F4 E3 04 C	DEFW RICEB
514		1088 30F6 0A 05 C	DEFW SPECB
515		1089 30F8 BE 03 C	DEFW TRASA
516		1090 30FA E4 03 C	DEFW EXTA
*****		1091 30FC EF 03 C	DEFW RICEA
517 0393 08 PPOINT EX AF,AF'		1092 30FE EA 04 C	DEFW SPECB
518 0394 D9 EXX		1093 3100	
University of Pavia - Francesco Leporati - Industrial Informatics and Embedded Systems 17			
519 0395 3A B1 02 D LD A,(PROG2A) ;reprogramming the port on the negative edge			

DIGITAL COMMUNICATION TECHNIQUES

Transmission of a character through polling control

```
        mov     dx, LSR
wait_tr:  in      al, [dx]
        and     al, MKTHRE
        jz      wait_tr
        mov     dx, THR
        mov     al, [carattere]
        out     [dx], al
```

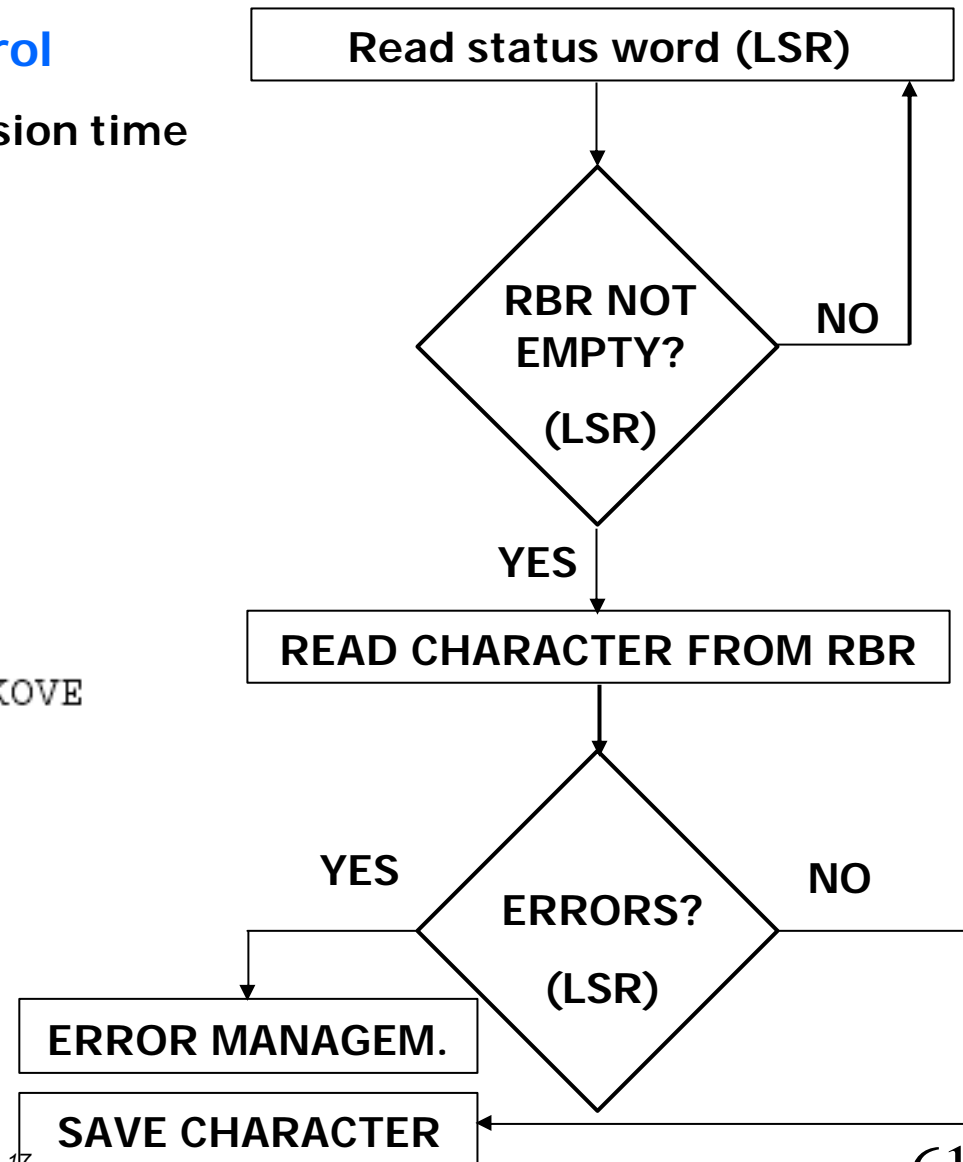


DIGITAL COMMUNICATION TECHNIQUES

Read character through polling control

polling with period < character transmission time

```
    mov     dx, LSR
wait_rc:
    in      al, [dx]
    mov     ah, al
    and     al, MKDRDY
    jz      wait_rc
    mov     dx, RBR
    in      al, [dx]
    and     ah, MKBRK+MKFRE+MKPAE+MKOVE
    jnz     err_ricez
    mov     [carattere], al
```



DIGITAL COMMUNICATION TECHNIQUES

Message format

Structured as "container" + "content"

Container = "header" + "trailer"

Initial characters = Start of message (1/2 bit)

Destination code for multipoint connections. Multicast and broadcast possible

Total n° of characters to establish the end of a binary message

Sender code (in particular for multipoint connections)

Progressive of transmission = order n° of the message (to verify the sequence correctness)

Progressive of reception = order n° of the last correct message received

Function code = the role played in the message by the current received fragment

Data (ASCII/binary)

Verification code = checksum, longitudinal parity or CRC

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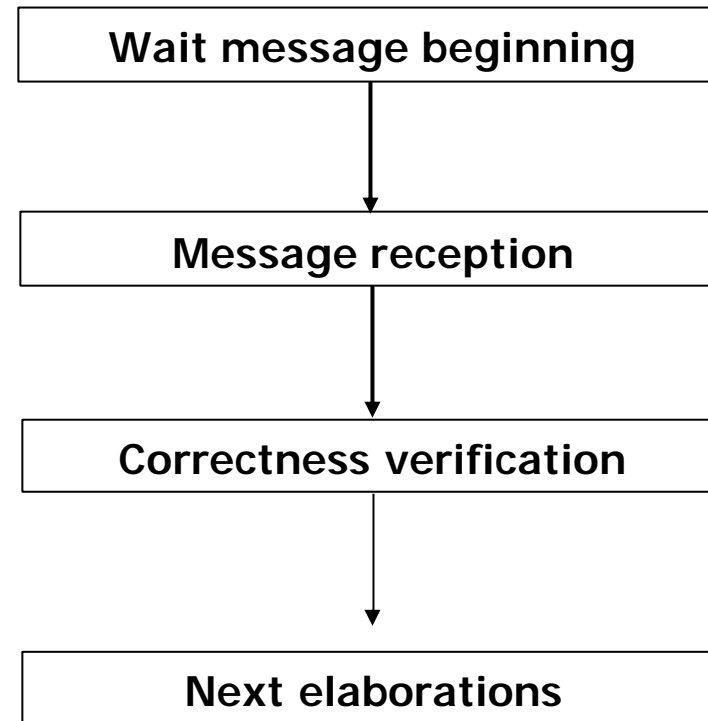
Message reception

“Relaxed” wait of the first character and fast reception of the next ones

Reception through polling/interrupt (often the 1° character with polling, the others with polling or interrupt), verification of the correctness and calculation of the verification code

Comparison of the two verification codes

Message elaboration and answer



DIGITAL COMMUNICATION TECHNIQUES

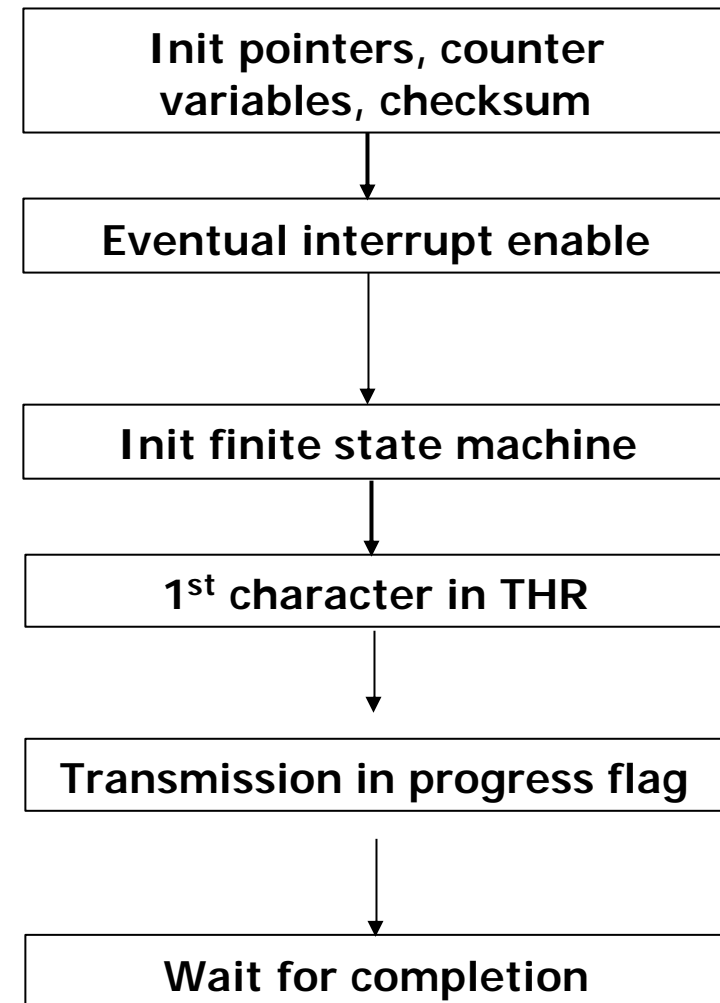
Transmission of messages - Constraints

if polling and if asynchronous transmission are implemented, too long break must be avoided to be no interpreted as time out (end of transm.)

if interrupt, the 1st character must be written in THR so triggering the interrupt request for next characters

The code is organized as a finite state machine

Before disabling the line, wait the end of the last character transmission



DIGITAL COMMUNICATION TECHNIQUES

An example: half duplex point-2-point protocol

Message format in application processes

tipo	byte di codice tipo messaggio
ncar	numero byte di testo ($0 < \text{ncar} < 256$)
testo	sequenza di byte di informazione utile

Available functions in application processes

<code>void ini_com</code>	(void)	inizializza la comunicazione
<code>void put_tra</code>	(mess)	inserisce messaggio in coda di trasmissione
<code>mess get_ric</code>	(void)	estrae messaggio da coda di ricezione. Se il sistema remoto non aveva nulla da trasmettere si riceve un messaggio di tipo = 0 e contenente un solo byte = 0 di testo.

Message format for communication processes

SOH	carattere iniziale <i>Start Of Header</i>
prog_t	codice progressivo del messaggio corrente
prog_r	codice progressivo dell'ultimo messaggio ricevuto correttamente
tipo	codice tipo messaggio
ncar	numero byte del testo
testo	contenuto del messaggio
cks	checksum di tutti i byte da SOH escluso

DIGITAL COMMUNICATION TECHNIQUES

An example: half duplex point-2-point protocol

Data structures

(communication level, 8 bit characters + parity)

<code>coda_ric</code>	coda di messaggi correttamente ricevuti in attesa di essere consumati dai processi applicativi.
<code>coda_tra</code>	coda di messaggi prodotti dai processi applicativi e da trasmettere.
<code>buf_ric</code>	array di caratteri per la ricezione del messaggio corrente.
<code>buf_tra</code>	array di caratteri per la trasmissione del messaggio corrente.

Communication procedures

<code>mess get_tra (void)</code>	estrae messaggio dalla coda di trasmissione. Se la coda è vuota si invia un messaggio di tipo vuoto (=0) con un solo byte (=0).
<code>void put_ric (mess)</code>	inserisce messaggio nella coda di ricezione
<code>interrupt int_ric ()</code>	funzione chiamata da interrupt di ricezione, cioè quando si è ricevuto un carattere o si ha un errore in ricezione. Organizzata come automa a stati che evolve con l'andamento del messaggio in corso di ricezione.
<code>interrupt int_tra ()</code>	funzione chiamata da interrupt di trasmissione, cioè quando si ha <i>Transmit Register Empty</i> . Organizzata come automa a stati che evolve con l'andamento del messaggio in corso di ricezione.
<code>int leggi_tempo (void)</code>	lettura tempo corrente, usata per il rilievo di time-out di fine messaggio

DIGITAL COMMUNICATION TECHNIQUES

An example: half duplex point-2-point protocol

Communication level variables

char	
stato_t	stato dell'automa di trasmissione
stato_r	stato dell'automa di ricezione
prog_t_rem	codice progressivo di trasmissione del sistema remoto
prog_t_loc	codice progressivo di trasmissione del sistema locale
prog_r_rem	codice progressivo di ricezione del sistema remoto
prog_r_loc	codice progressivo di ricezione del sistema locale
cks_t	checksum corrente del messaggio in trasmissione
cks_r	checksum corrente del messaggio in ricezione
count	numero caratteri del testo da ricevere
int	
it	indice corrente del buffer di trasmissione
ir	indice corrente del buffer di ricezione
long int	
tempo_prec	tempo di ricez. carattere precedente (per t-out)
tempo_corr	tempo di ricez. carattere attuale (per t-out)

In the following slides an example of code managing message reception and transmission interrupt after interrupt vectorization routines having been settled together with initialization of registers and identification through IIR of the interrupt origin

DIGITAL COMMUNICATION TECHNIQUES

An example: half duplex point-2-point protocol – Reception interrupt

```
void leggi_byte_uart (void)
{
    byte = leggi_rbr (); // registro ricezione UART
    errori = check_lsr (); // line status per verif
    if (errori)
        {tempo_prec = leggi_tempo (); // tempo arrivo by
        stato_r = S_ER;}
    switch (stato_r)
    {case S0:
        if (byte == SOH)
            stato_r = S1;
        else
            {tempo_prec = leggi_tempo (); // t arrivo
            stato_r = S_ER;}
        break;
    case S1: // aspetta progressivo trasmesso remoto
        cks_r = 0; // azzerà checksum
        ir = 0; // indice buffer a inizio
        prog_t_r = byte;
        cks_r += byte;
        stato_r = S2;
        break;

    case S2: // aspetta progressivo ricevuto remoto
        prog_r_r = byte;
        cks_r += byte;
        stato_r = S3;
        break;

    case S3: // aspetta codice tipo
        buf_ric [ir] = byte;
        ir++;
        cks_r += byte;
        stato_r = S4;
    }
```

```
        break;

    case S4: // aspetta numero byte testo (>=1)
        buf_ric [ir] = byte;
        ir++;
        count = byte;
        cks_r += byte;
        stato_r = S5;
        break;

    case S5: // aspetta i byte del testo
        buf_ric [ir] = byte;
        ir++;
        cks_r += byte;
        if (--count == 0)
            stato_r = S6;
        break;

    case S6: // aspetta byte di checksum
        if (byte == cks_r) // verifica checksum
            {put_ric (buf_ric); // nuovo messaggio corretto
            prog_r_l = prog_t_r;
            disab_ric ();
            enab_tra (); // abilita trasmissione
            stato_r = S0;}
        else // errore checksum
            {tempo_prec = leggi_tempo (); // t arrivo byte
            stato_r = S_ER;}

        break;

    /* ERRORE - Lascia passare, ignorandoli, tutti gli eventuali
    caratteri rimanenti e aspetta l'inizio del messaggio successivo
    riconosciuto dal carattere SOH ricevuto dopo lo scadere di time-o
    */
    case S_ER: // ERRORE -
        tempo_corr = leggi_tempo (); // tempo arrivo byte
        if ((tempo_corr-tempo_prec)>=TOUT) && (byte==SOH))
            stato_r = S1;
        else
            tempo_prec = tempo_corr;
        break;
    }
```

DIGITAL COMMUNICATION TECHNIQUES

An example: half duplex point-2-point protocol – Transmission interrupt

```
void scrivi_byte_uart (void)
{
    switch (stato_t)
    case S0:
        /* vede se l'ultimo messaggio trasmesso e' stato
           ricevuto correttamente da sistema remoto */
        if (prog_r_r == prog_t_l)
            prog_t_l++;
            buf_tra = get_tra (); // nuovo messaggio
            if (buf_tra == NULL) // nulla da trasmett
                {buf_tra[0] = 0; // tipo
                 buf_tra[1] = 1; // num. byte
                 buf_tra[2] = 0; } // byte testo
            byte = SOH;
            scrivi_thr (byte);
            stato_t = S1;
            break;

    case S1: // trasmette progressivo trasmesso locale
        cks_t = 0; // azzerà checksum
        it = 0; // indice buffer a inizio
        byte = prog_t_l;
        scrivi_thr (byte);
        cks_t += byte;
        stato_t = S2;
        break;

    case S2: // trasmette progressivo ricevuto locale
        byte = prog_r_l;
        scrivi_thr (byte);
        cks_t += byte;
        stato_t = S3;
        break;

    case S3: // scrive codice tipo
        byte = buf_tra [it];
        scrivi_thr (byte);
        it++;
        cks_t += byte;
        stato_t = S4;
        break;

    case S4: // trasmette numero byte testo (>=1)
        byte = buf_tra [it];
        scrivi_thr (byte);
        it++;
        count = byte;
        cks_t += byte;
        stato_t = S5;
        break;

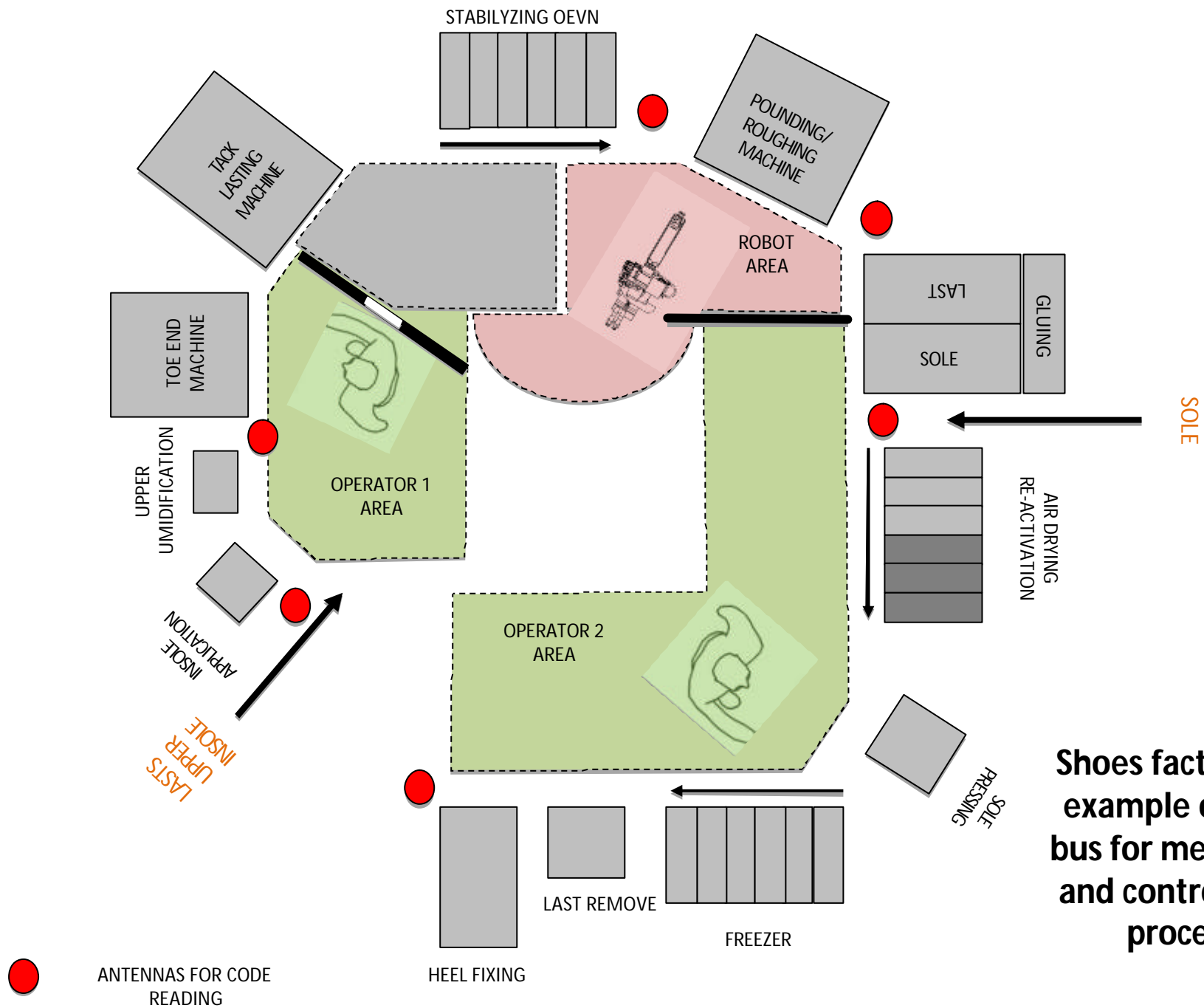
    case S5: // trasmette i byte del testo
        byte = buf_tra [it];
        scrivi_thr (byte);
        it++;
        cks_t += byte;
        if (--count == 0)
            stato_t = S6;
        break;

    case S6: // trasmette byte di checksum
        byte = cks_t;
        scrivi_thr (byte);
        stato_t = S7;
        break;

    case S7: // termine trasmissione
        disab_tra ();
        enab_ric ();
        stato_t = S0;
        break;
}
```

FIELD BUS – A typical application





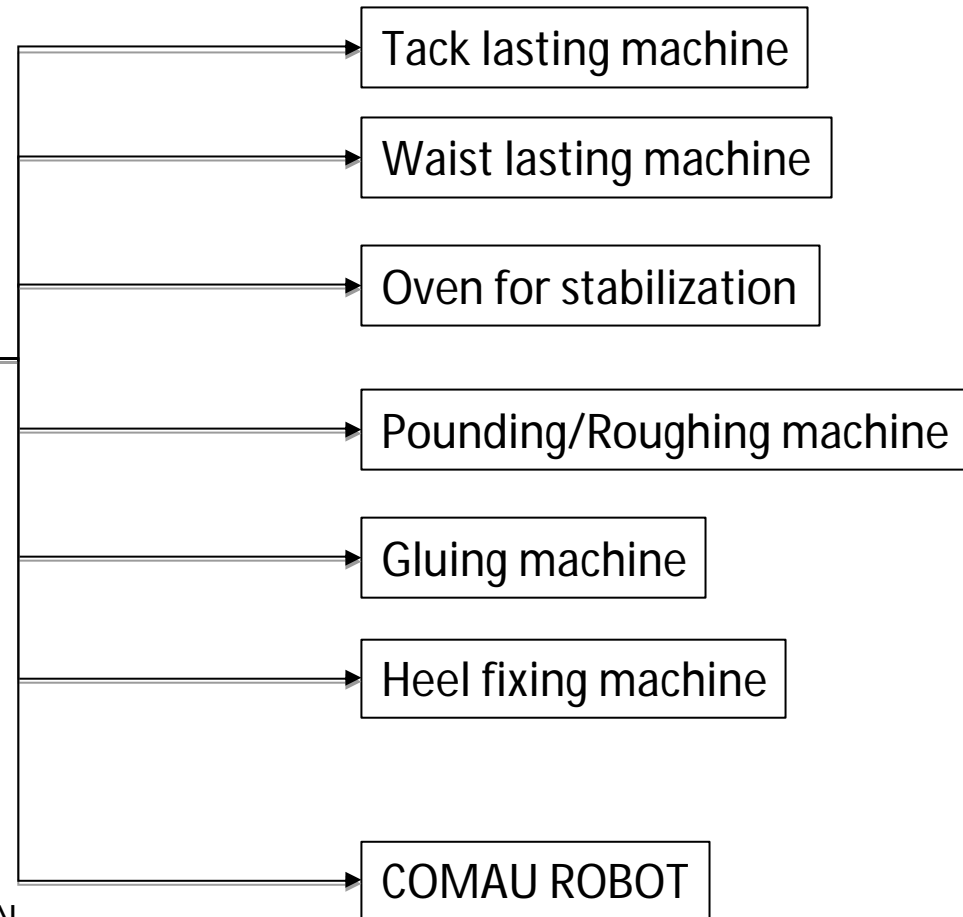
Shoes factory: an example of field bus for measuring and controlling a process

PC CONTROL and its connections



Serial line for communication with the machines

- COMPONENTS IDENTIFICATION
- MOUNTING STATION (island)
- POUNDING/ROUGHING INPUT
- SOLE IDENTIFICATION
- HEEL FIXING



Connection through antennas to read the semi-manufactured product

DIGITAL COMMUNICATION TECHNIQUES

Field Bus

Standardization of communications between devices used in data acquisition and industrial process control.

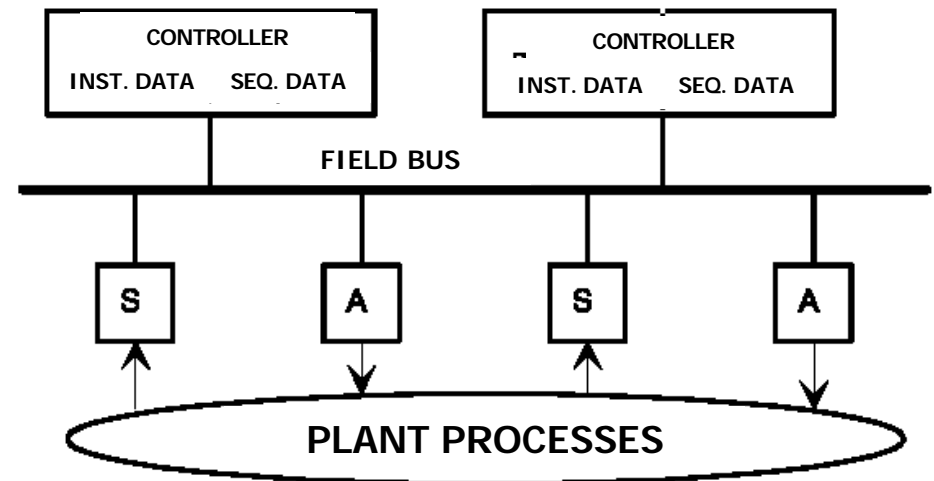
Two possibilities: industrial networks and field buses

- Networks (Ethernet) diffused and popular. Faster, wider, more expensive than field buses
- Field buses (Can) good compromise among performance, real time, robustness, costs

A field bus is a distributed control system made up by one or more controllers, sensors and actuators

Data are “logically divided” in controllers:

- “process data base”: status process variables, periodically refreshed through instantaneous sensor measurements
- “historical data base” with significant events organized in queues (sequential)



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Required services

periodic communication of sensor acquisitions and command towards the actuator

aperiodic communication with high priority of events (alarms, anomalies, ...)

aperiodic communication of low priority infos (configurations, on line diagnostics)

Constraints (specs)

Instantaneous data refresh frequency = sensor sampling frequency

Time validity of values = consider that the physical distance between producer and consumer implies a delay that does not compromise the system stability

Temporal coherence = data received from many different producers must refer to the same "time period"

Spatial coherence = if more controllers use the same variable they must use identical values

Interoperability = working although different vendors

Sporadic messages must be affected by the minimum possible delay (real time?)

Simple interfaces, cheap, reliable and safe

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Transmission media

copper cables: for short-medium distance connections (from m to Km) but the bandwidth diminishes with the distance due to dissipation in the medium. To achieve velocities greater than Mb/sec at great distances repeaters or bridges are necessary

twisted pairs: cheap, with good noise immunity and transmission velocities (Mb/sec). They can be shielded (but sometimes no) and they are also used to power supply the devices connected through the bus (modulation)

coaxial cables: definitively better (in terms of bit rate, electromagnetic compatibility and reachable distances) than twisted pairs; used where technical difficulties do not allow the use of optical fibers (i. e. vibrations)

cables for energy transport: used in automatic devices installation in buildings (domotic) or to bring digital communication exploiting wide distribution supports (water, electricity, gas); transmission is made through signal modulation superimposed on the AC network voltage thus using the same transmission medium. Sensible to electromagnetic disturbs.

optical fibers: at now the most performing technology in terms of velocity, distance and electromagnetic compatibility however with important costs. That can be partially reduced by the relatively new plastic optical fiber technology

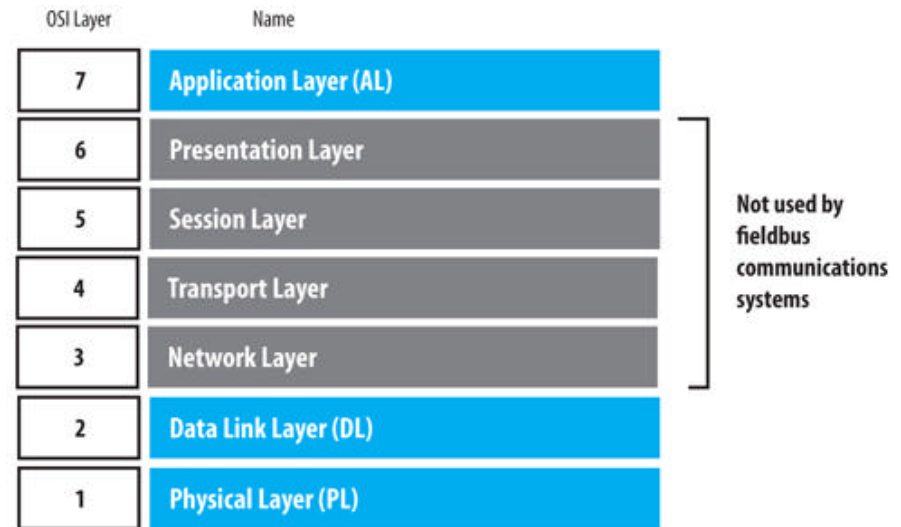
Wi fi: ever and ever increasing diffusion but still very sensible to disturbs

DIGITAL COMMUNICATION TECHNIQUES

Architecture of a typical field bus

For high performance communication systems (low latency, determinism, high throughput) or in case of limited available resources a slimmer architecture version can be implemented of the ISO/OSI model featuring the sole Application, Data Link and Physical (Enhanced Performance Architecture) levels

The 3, 4, 5, 6 levels are not explicitly implemented except some minimal functions are included in Application or Data Link or implemented in a new intermediate level



Efficiency

parameter that combines transmission velocity with the n° of data transferred in the unit of time

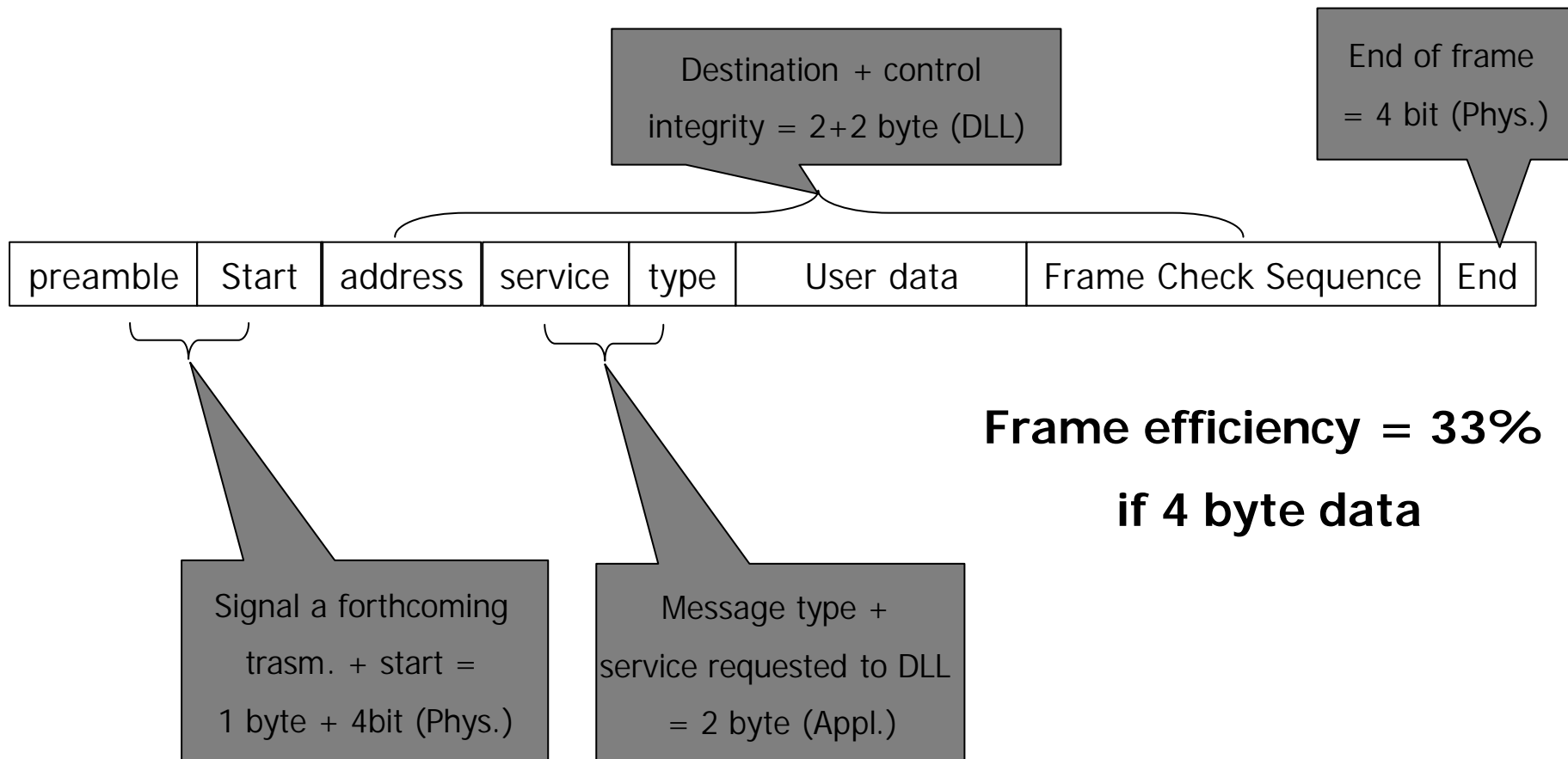
$$\text{Throughput} = \text{Data} * \text{Velocity}$$

while the velocity is generally known, the throughput is not since due to specific working conditions and depending on the frame overhead, the bus access mode, address technique and communication model, mode of use of the bus

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Frame overhead

The transmitted data frames are “packed” within predefined structure (Protocol Data Unit) that are composed in the different layers of the ISO/OSI model, for instance:



DIGITAL COMMUNICATION TECHNIQUES

Medium Access Control

Many devices access to the same transmission medium: this can create conflicts so compromising the data exchange correctness

MAC policy is fundamental to assure that data traffic respects the required time constraints i. e. *deterministic*

MAC is implemented in the Data Link Layer and differentiates the field buses on the basis of three access control modes:

- centralized: only 1 node assigns the control to the different connected stations
 - master/slave or bus-arbiter
- distributed: many or all the stations can assume the transmission mode control
 - Token passing or TDMA
- multiple: policies in which all the nodes can transmit at the same time; the difference is made on the basis of conflicts management
 - Carrier Sense Multiple Access with collision detection or avoidance or resolution or arbitration

DIGITAL COMMUNICATION TECHNIQUES

Master/slave

A unique station called master is authorized to spontaneously transmit

The other nodes can answer on the basis of a cyclic interrogation by the master following a schedule predefined (polling list) or modifiable (dynamic) traffic-sensitive

Every slave, when asked, provide data or asks for a service

Advantage: deterministic traffic control ? upgrade times assured

Bus arbiter (Link Active Scheduler)

A unique station is the communication arbiter and controls a network of nodes configured as information (variables) producers or consumers

Every variable is identified by its proper name and is “called” by the arbiter on the bus following a scheduling table dynamically managed

The producer node will put its own variable on the bus at every call

The consumer node will read that variable at every transmission

The arbiter can temporarily distribute the control to other nodes to support different communication modes (client server)

Advantage: deterministic traffic control ? upgrade times assured

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Token passing

The access to the bus is reserved to the station (master) that holds a suitable token

A token is a particular frame format exchanged among stations

The token is held until a suitable maximum time or until the communication ends

We have a “ring” among token exchanging nodes (with a fixed Token Rotation Time)

If the token acquisition time is constant within the TRT a “Time Locked Token Rotation” is performed otherwise “Free Running Token Rotation”

Advantage: deterministic traffic control ? upgrade times assured

TDMA (Time Domain Multiplexed Access)

A *cycle time* is defined within which all the nodes must have the possibility for communicating

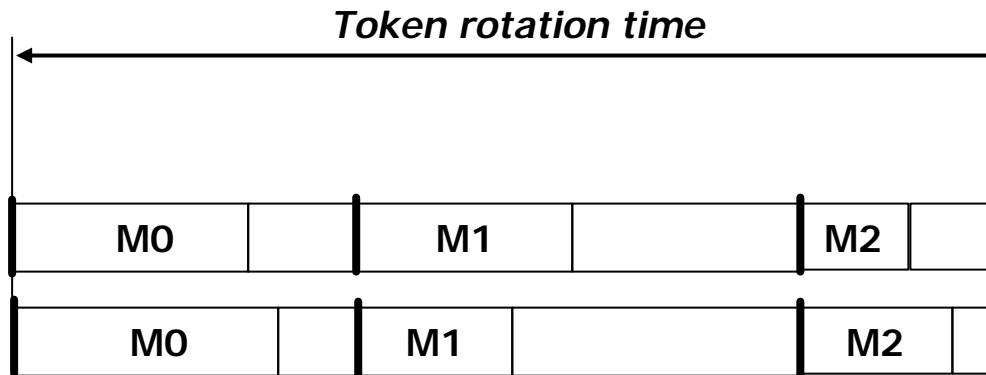
Within the cycle every node is assigned with a “time window” to be the bus master

Similar mechanism as Token passing Time Locked (without token)

Every station “knows” its proper time since it has a clock synchronous with the bus one

DIGITAL COMMUNICATION TECHNIQUES

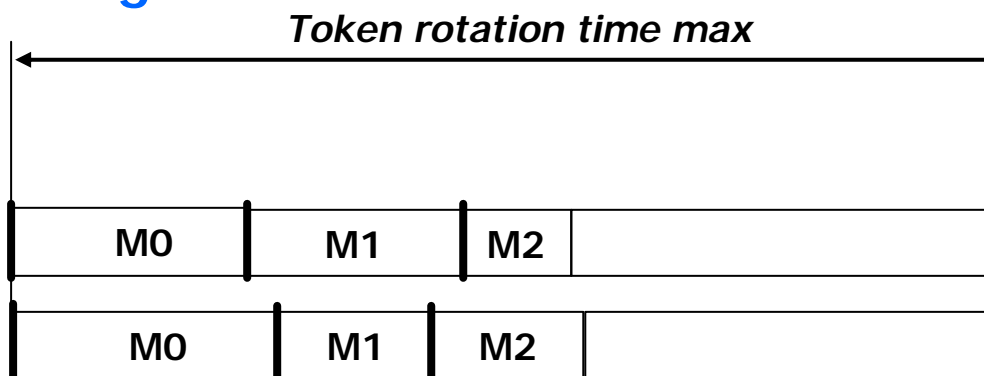
Time Locked



Bus idle

Trasmission

Free Running



Token passage

DIGITAL COMMUNICATION TECHNIQUES

Carrier Sense Multiple Access (CSMA)

All the network nodes can send data. Every node is listening by default

If the transmission medium is not occupied, the node sends the data: possibility of collisions. Collisions can be managed adopting different approaches:

1-Collision Detection

In presence of a collision the transmission is suspended

Collided messages are deleted from the medium and they will be retransmitted

The transmission resumes after a random time. The probability of new collisions is related to the network load and to the number of nodes connected to the network

2-Collision Avoidance

Similar to Collision Detection

The waiting time before retransmitting data is randomly generated by each node (it is not a unique values for all the nodes)

The probability of new collisions is related to network load and to the nodes connected to the network. The probability of collision is lower than in Collision Detection since each node waits for a different time before resuming the transmission

DIGITAL COMMUNICATION TECHNIQUES

3-Collision Resolution

A priority is assigned to each node during network configuration

Collided messages are deleted from the network for being retransmitted

Transmission resumes after an inactivity time related to the priority assigned to the node or to the message

It is similar to Collision Avoidance, but the waiting time is not randomly generated since it is computed based on priority (this guarantees that the collision cannot be repeated = resolution)

4-Bus Arbitration

A priority is assigned to each node during network configuration

Transmission resumes when the transmission medium is not occupied

If there is a collision, only the transmission with the bigger priority continues. The other nodes will resume their transmission when higher priority transmissions end

Priority frame and check of what a node is transmitting and what is receiving from the bus. A frame indicates that a higher priority node is transmitting. Ex. Can bus

This technique is not destructive since, when a collision happens, transmitted messages are not deleted but their transmission is blocked

DIGITAL COMMUNICATION TECHNIQUES



A letter is sent.

Communication services

Communication services of a field bus can be:

The sender doesn't know:

- If the receiver is present
- If the letter arrived
- If the receiver can read the letter

The receiver doesn't know:

- When it is coming

- **connectionless** based on complete and self-describing messages that can be understood by the receiver without additional information. Messages contain format, length, encoding, data and how to use them

Useful approach when communication requires occasional accesses to remote stations (commissioning data, identification, file and programs download or upload). The network configurator/manager is not responsible of establishing the connection

Approach not useful in case of repetitive communications (i. e. cyclical updating of information). The overhead relate to descriptive information becomes useless and heavy

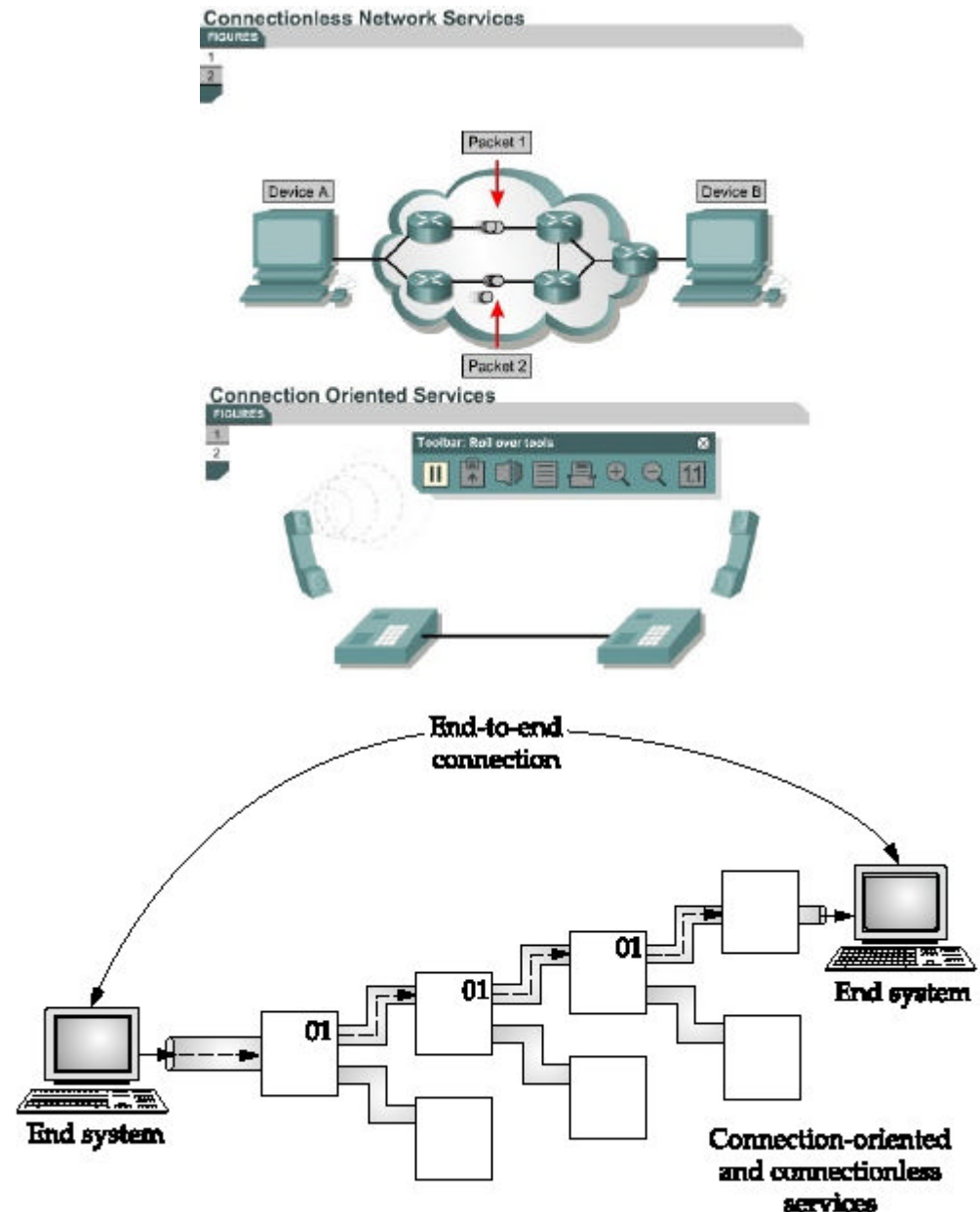
DIGITAL COMMUNICATION TECHNIQUES

Communication services

Communication services of a field bus can be:

connection oriented the coupling between sender and receiver is established by a connection ID. Exchanged messages contain only data and ID with a higher coding efficiency with respect to connectionless services. There is an implicit coupling between info source and destination

Useful approach for applications based on cyclical repetitions of great numbers of similar messages and for Data Link Layer services (at Application level it is possible to make implicit only information related to high level protocols)



DIGITAL COMMUNICATION TECHNIQUES

Addressing Techniques

- *point-to-point addressing*

Technique used by Client-Server communication services.

Source Address (SA) and Destination Address on Protocol Data Unit of the Data Link level

- *multicasting*

Messages are identified by a SA and a Group Address, known by nodes belonging to a group (i.e. nodes connected to the same functional unit).

Advantages are the lower transmission loads. There is the need of physical and temporal consistency of transmitted information

- *broadcasting*

Messages are concurrently transmitted to all the network nodes using a global ID

Essential for supporting network maintenance and management services. Each field bus has a set of global IDs

DIGITAL COMMUNICATION TECHNIQUES

Communication models

• *producer consumer*

It combines the concepts of global addressing and connection oriented service

Each station configured as producer couples data with its connection-ID (unambiguous in the network). Receiving stations will verify at Data Link Level if information must be forwarded to higher levels, checking if they are consumers coupled with the ID or not

It combines sending efficiency of cyclical information, without transmission repetition, with the possibility of guaranteeing physical and temporal consistency

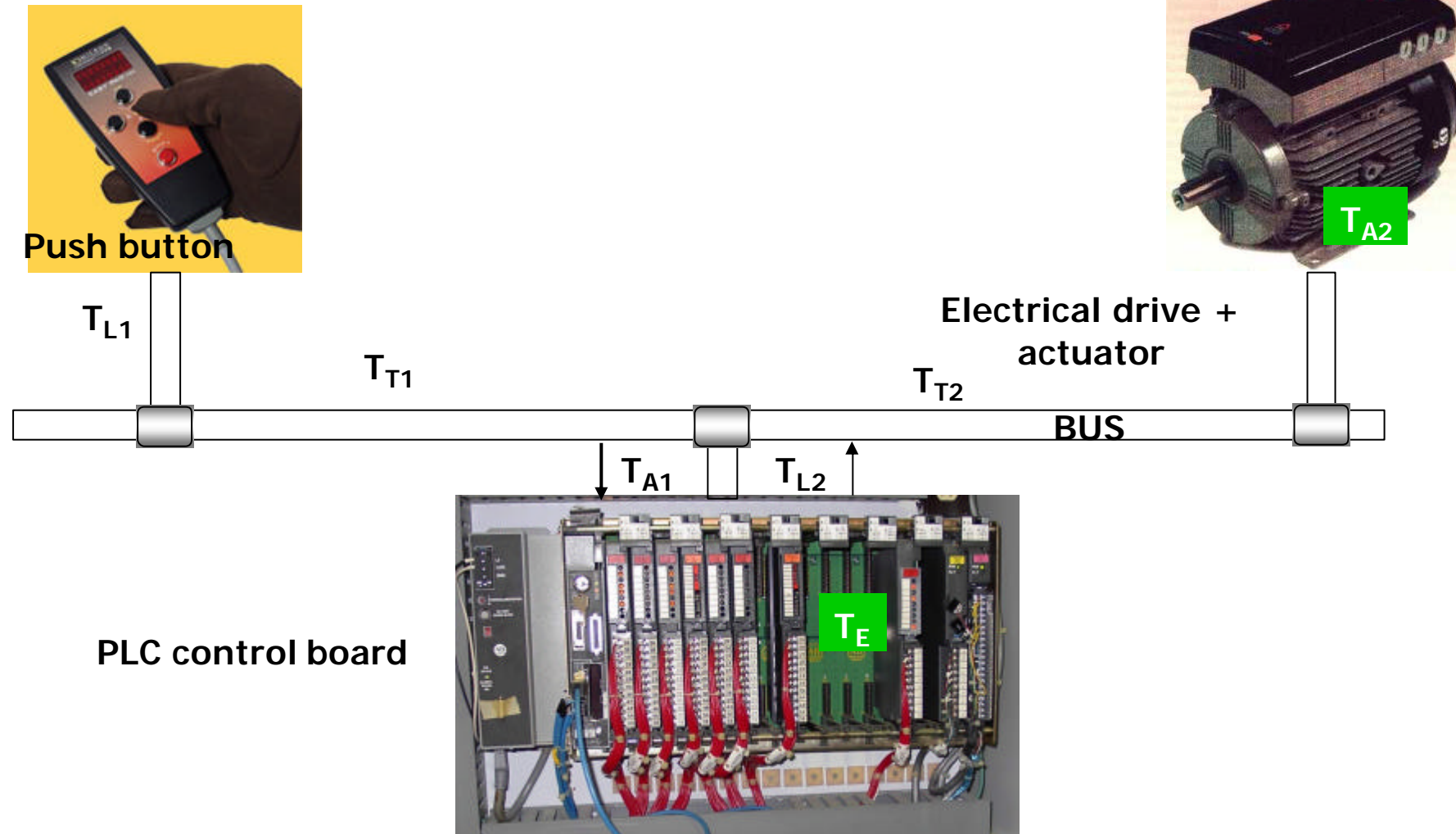
Unsuitable in case of acknowledged services

• *client server*

Point-to-point addressing of connectionless and/or with acknowledged services: a node becomes a client when it explicitly requests a services to a remote station (server)(command execution, event based transmission, I/O...)

DIGITAL COMMUNICATION TECHNIQUES

Response time and latency of a field bus: an example



DIGITAL COMMUNICATION TECHNIQUES

Response time and latency of a field bus: an example

$t=0$, the operator press the button

$t=T_{L1}$ The “pressed button” information is on the bus. Latency due to the bus DLL and related to the implemented MAC (centralized/distributed) and to the transmission technique (polling/interrupt). If the protocol is deterministic T_{L1} can be estimated (2 ms)

$t=T_{L1}+T_{T1}$ The information arrives to the PLC after a time that depends on the bit transmission speed and the bit number added to the propagation time of the medium (propagation speed and distance). i.e. 0.06 msec

$t=T_{L1}+T_{T1}+T_{A1}$ The information is in the PLC registers, ready for being elaborated by the CPU after a time related to internal interface register communication transmission speed

$t=T_{L1}+T_{T1}+T_{A1}+T_E+T_{L2}$ The information is sent to the drive after a time T_E equal to a PLC elaboration cycle (IN, EXEC, OUT) with a latency T_{L2} as discussed before. T_E is in the range 0,2-20 ms/Kinstructions

$t=T_{L1}+T_{T1}+T_{A1}+T_E+T_{L2}+T_{T2}+T_{A2}$ the command is implemented after a time T_{A2} needed by the drive software for acquiring information from the interface and to activate the drive (0.01 ms) and after a time T_{T2} due to transmission on the medium same as before)

DIGITAL COMMUNICATION TECHNIQUES

Response time and latency of a field bus: an example

$$T_{Rmax} = T_{L1max} + T_{T1} + T_{A1max} + T_E + T_{L2max} + T_{T2} + T_{A2max}$$

T_{A1max} at maximum equal to a PLC elaboration cycle so

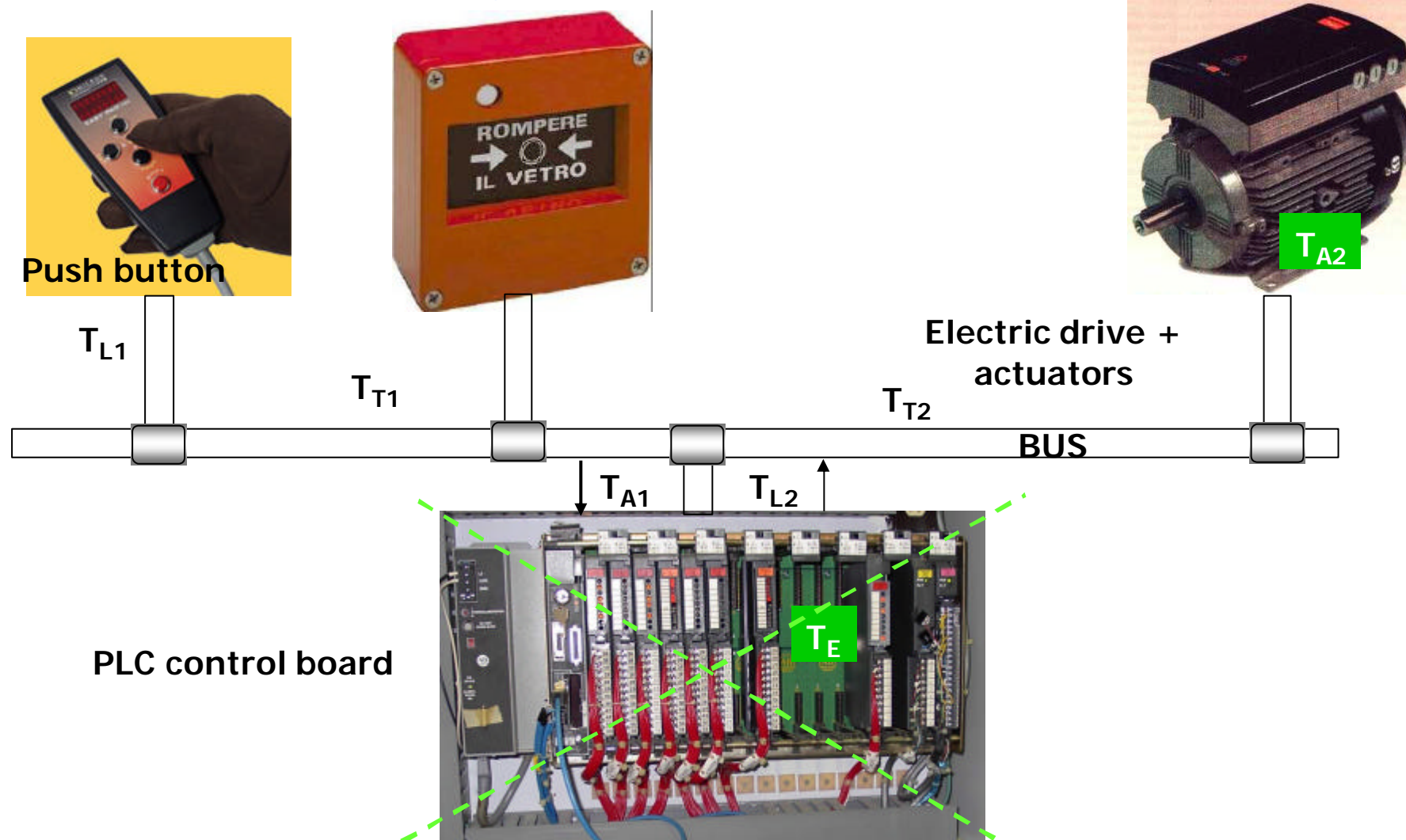
$$T_{Rmax} = (T_{L1max} + T_{L2max}) + (T_{T1} + T_{T2}) + 2T_E + T_{A2max} =$$
$$4 + 0.12 + (0.4-40) + 0.01 \approx 4.5-44.1 \text{ ms}$$

The field bus performance is only one of the elements that determine the response time of the entire system.

This is further highlighted if we add to this scheme an emergency button to stop the system as fast as possible eventually overtaking the elaboration of the PLC.

DIGITAL COMMUNICATION TECHNIQUES

Response time and latency of a field bus: an example



DIGITAL COMMUNICATION TECHNIQUES

Response time and latency of a field bus: an example

$t=0$, the operator press the button

$t=T_{L1}$ The “emergency stop” information is on the bus. Latency is due to the bus DLL, to the implemented MAC (centralized/distributed) and to the transmission technique (polling/interrupt). If the protocol is deterministic T_{L1} can be still estimated (2 ms)

$t=T_L+T_T$ the info arrives to the drive after a transmission time T_T ; again, this time depends on the modulation time of the frame and on the transmission on the medium. It can be estimated in $0.06+0.06$ ms (as before more or less)

$t=T_L+T_T+T_A$ the command is implemented after a time T_A needed by the drive sw to acquire the information from the interface and to stop the drive (0.01 ms)

In this case the response time will be

$$T_{Rmax}=T_{Lmax}+T_T+T_A \approx 2 \text{ ms}$$

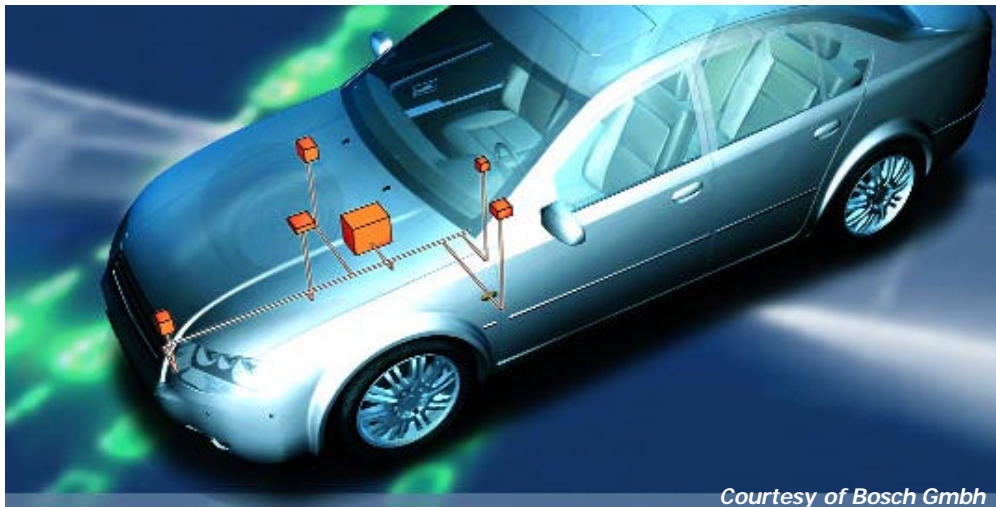
- The bus performance are responsible of the response times
- The transmission latency is a good indicator of the bus performance
- The control distribution (especially when simple) helps

DIGITAL COMMUNICATION TECHNIQUES

CAN (Controller Area Network) Bus

Serial communication protocol proposed by Bosch in the 80s in automotive applications to allow a distributed real time control, using intelligent sensors and actuators mounted on board of a vehicle assuring also a high safety level.

Successively it has been adopted in several industrial fields



www.semiconductors.bosch.de/pdf/can2spec.pdf

Serial transmission with velocity up to 1 Mb/sec (some releases until 10 Mb/s)

Excellent error detection and management with restriction of possible causes and confinement of the responsible, low costs, noise tolerance (i. e. low probability of a not detected error in normal working conditions)

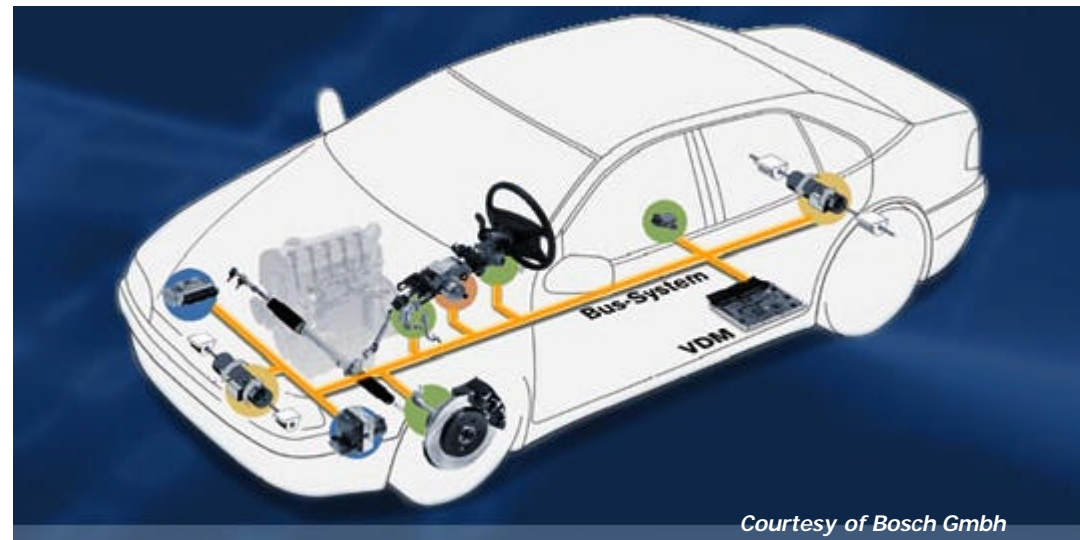
Real time applications (tools available to have reliable and fixed answer times)

DIGITAL COMMUNICATION TECHNIQUES

CAN (Controller Area Network) Bus

A simple and flexible wiring : twisted pairs are used (sometimes shielded)

Nodes can be added or cancelled without any re-configuration (no address nodes)



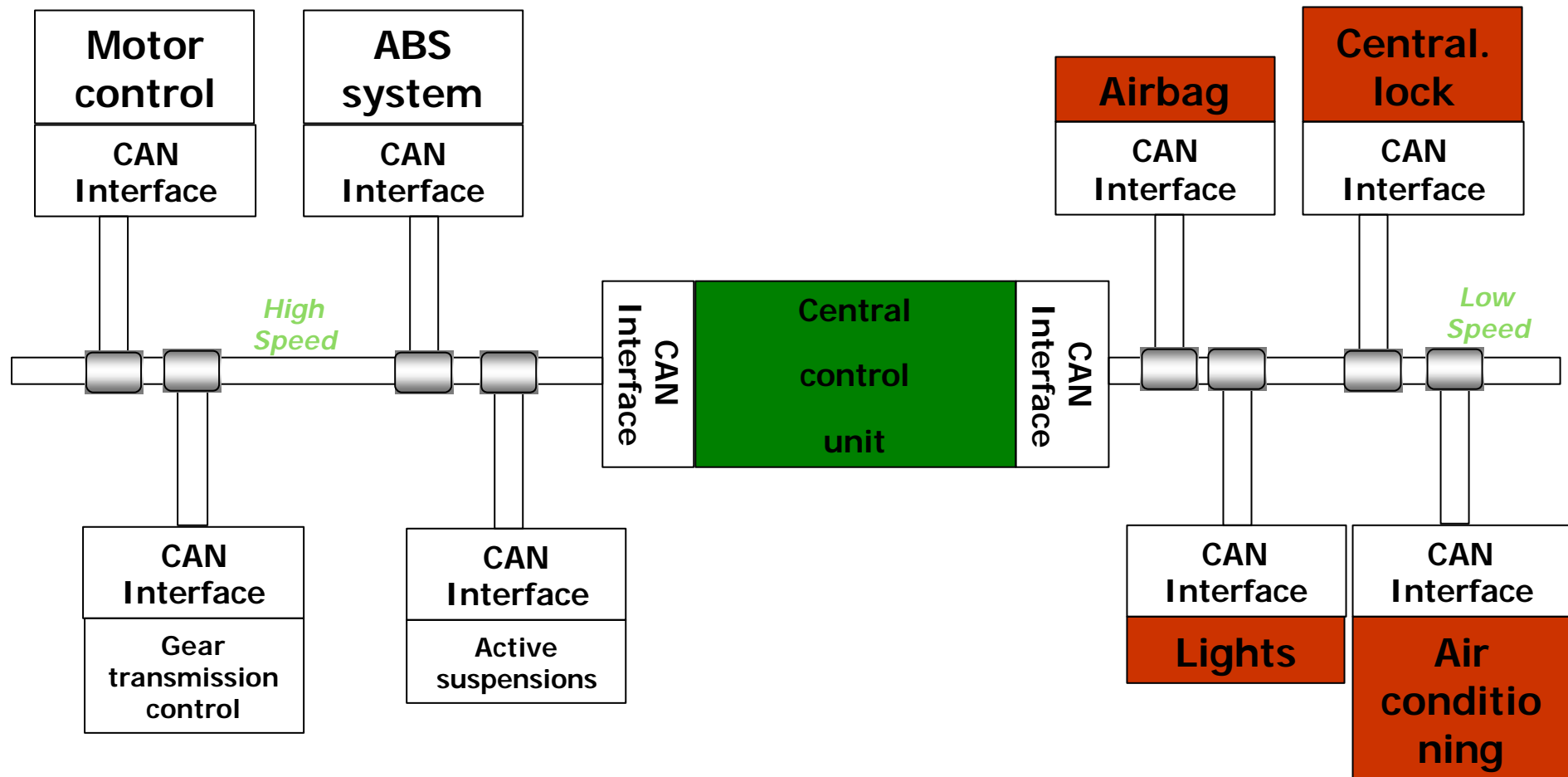
Noise immunity: shielding and differential transmission of the signal

Stabilized standard: wide diffusion and induced availability of hw devices and development tools for different application fields

Recognized as standard (ISO 11898)

DIGITAL COMMUNICATION TECHNIQUES

CAN (Controller Area Network): example



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: communication

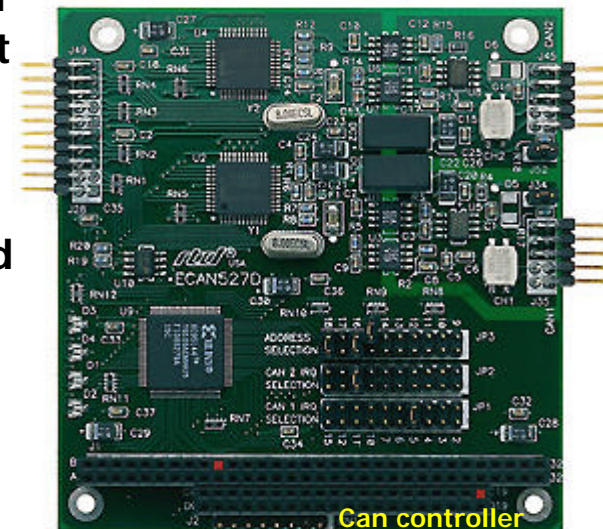
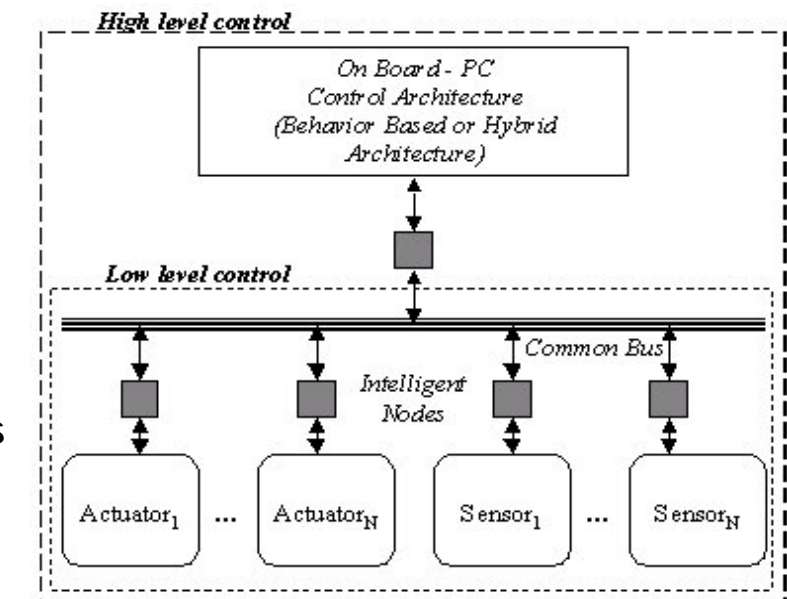
Intelligent sensors and actuators, (equipped with CPU, ADC, RAM e EPROM, timer integrated in the chip) able to pre-amplifying, convert the acquired value and partially pre-elaborate it before transmitting

MAC of the kind "Multi-Master" where all the nodes can transmit at the same time

Nodes are not addressed but a messages is labeled with a suitable *identifier* (maybe of the sender), on the basis of which the receivers elaborates it or not

A suitable priority in the bus access (arbitration) is associated to the identifier

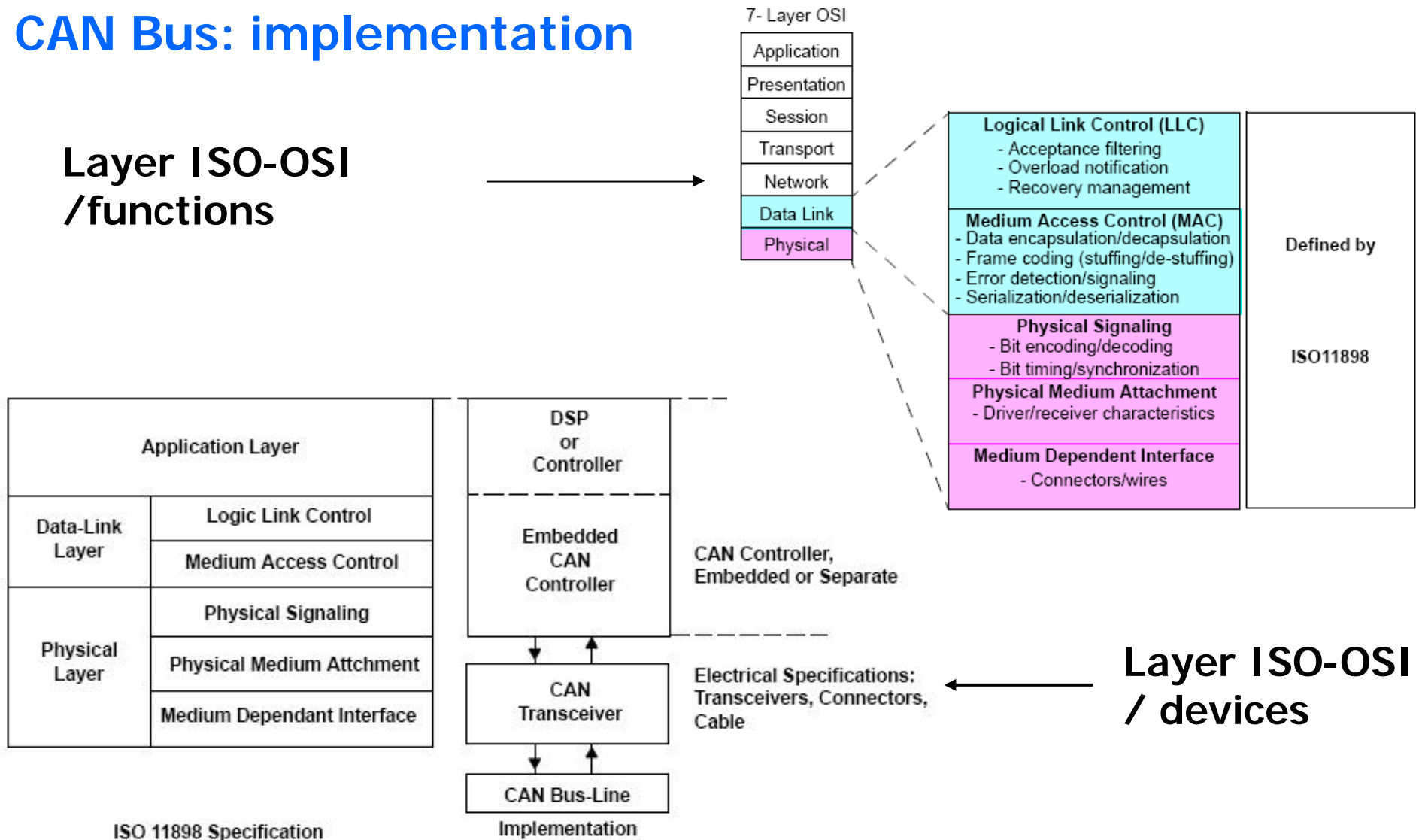
The CAN protocol implements only the Physical and the Data Link layers of the ISO/OSI model while a Application layer is foreseen however it is not implemented in the standard but left to the user



DIGITAL COMMUNICATION TECHNIQUES

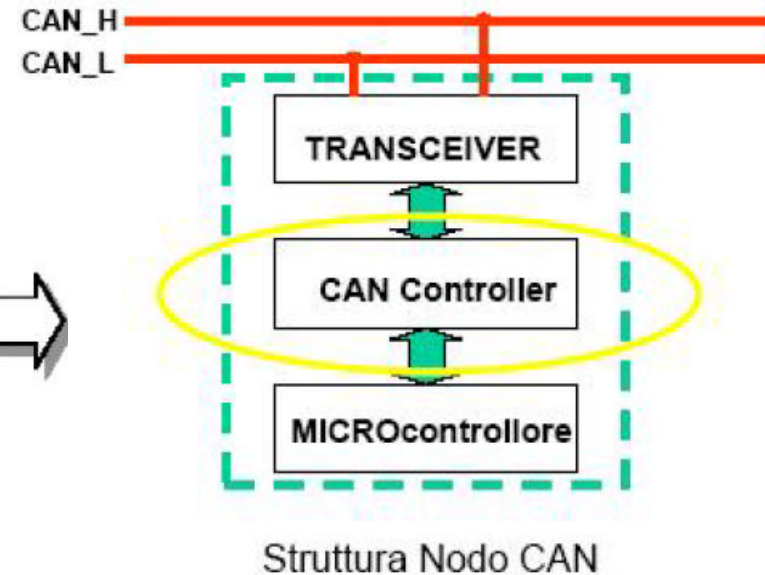
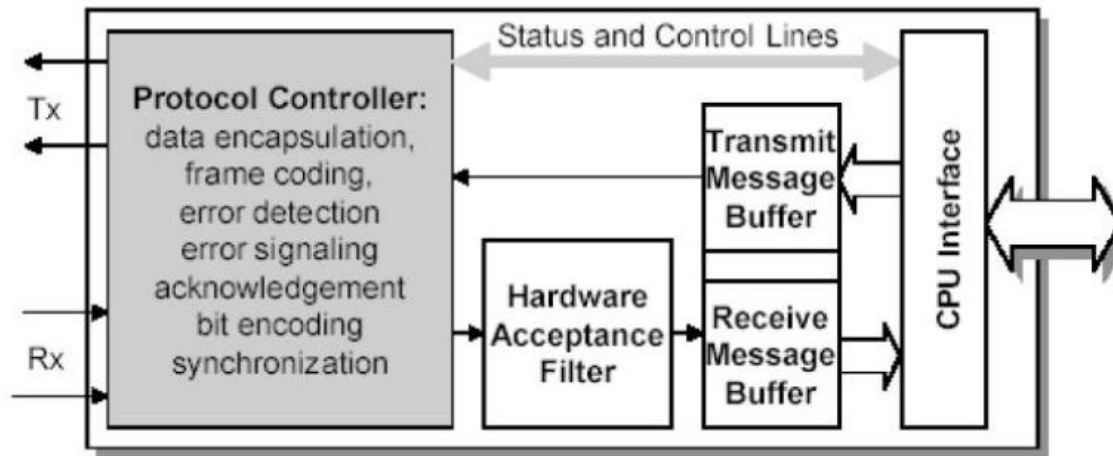
CAN Bus: implementation

Layer ISO-OSI /functions

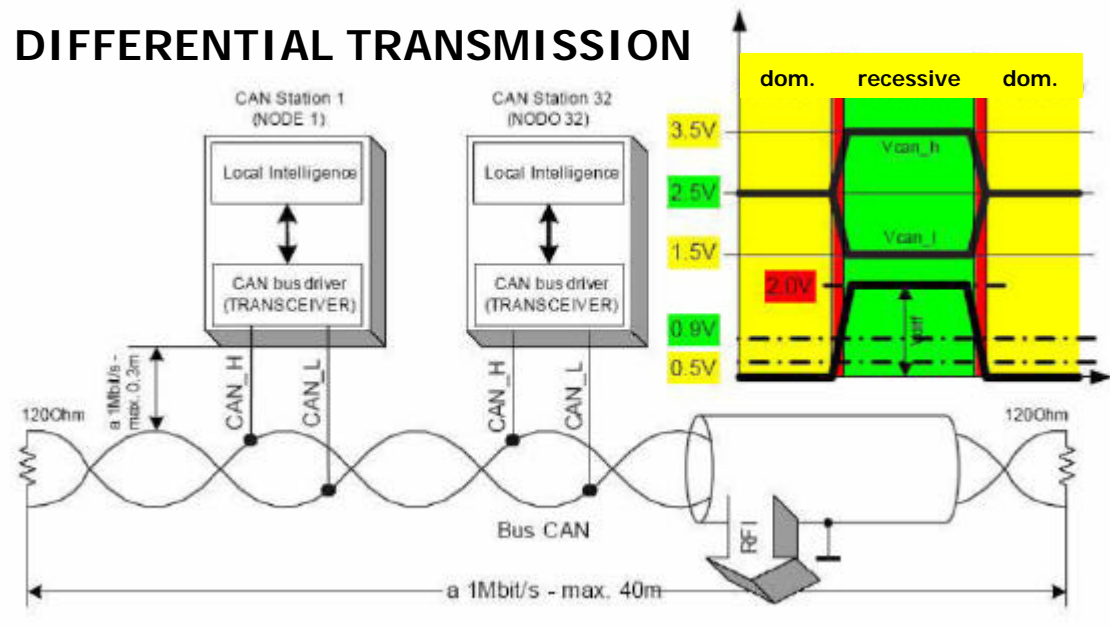


DIGITAL COMMUNICATION TECHNIQUES

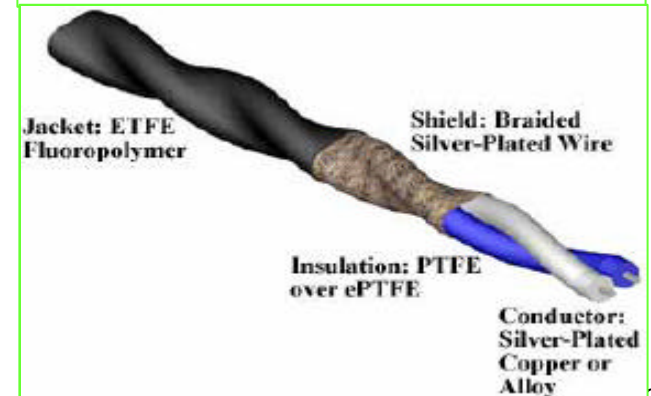
CAN Bus: node structure



DIFFERENTIAL TRANSMISSION



SJA 1000 Philips (twisted pair shielded)



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: physical level

Specifica le caratteristiche elettriche della comunicazione

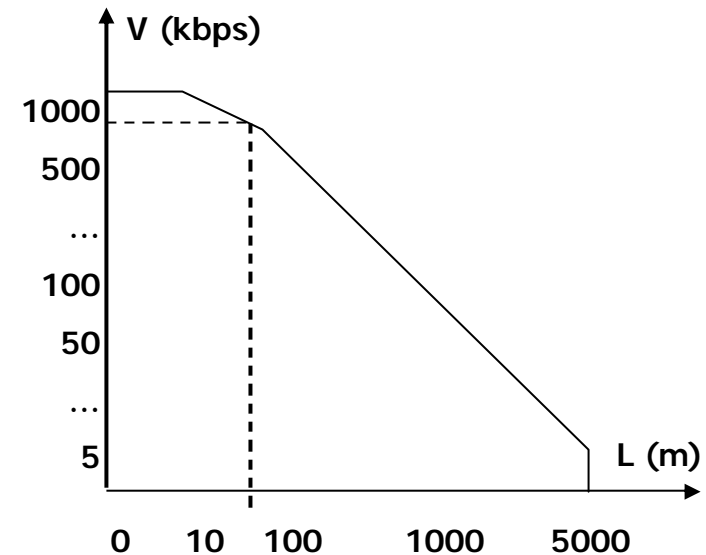
Il mezzo trasmissivo solitamente utilizzato è un doppino intrecciato, schermato (a seconda della rumorosità ambientale), in differenziale collegato ad un amplificatore differenziale (CAN_H e CAN_L, ok se rumore analogo sui 2 cavi)

Dipendenza circa lineare (log log) se mezzo trasmissivo elettrico (1 Mb/s entro i 40 m)

Codifica bit di tipo NRZ (not return to zero): più critica la sincronizzazione

I livelli logici trasmessi sono denominati "dominante" e "recessivo" e la corrispondenza con "1" o "0" cambia in base a come i nodi si affacciano sul bus (AND/OR)

In caso di trasmissione contemporanea di un bit "r" da parte di un nodo e di uno "d" da parte di un altro, "vince" il dominante e chi ha trasmesso "r" s'accorge del conflitto, restando in ascolto del bus e verificando la differenza tra ciò che ha trasmesso e quanto ha rilevato



DIGITAL COMMUNICATION TECHNIQUES

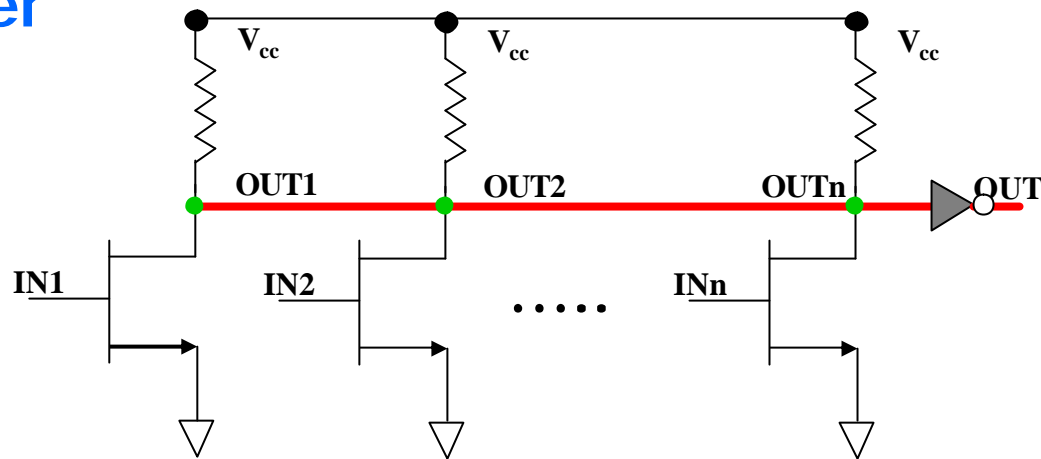
CAN Bus: physical layer

CONNECTION WIRED OR

$$OUT = IN1 \mid IN2 \mid \dots \mid INn$$

Dominant level "1"

Recessive level "0"

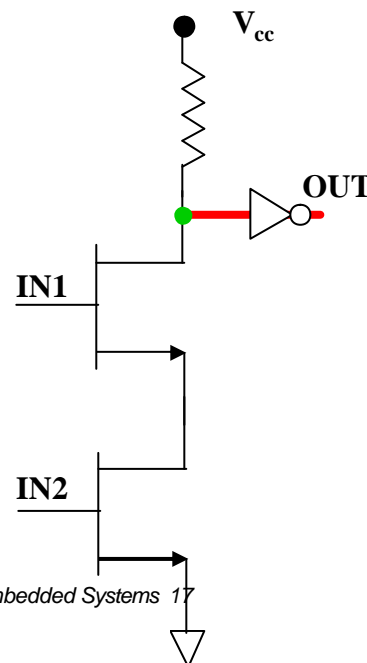


CONNECTION WIRED AND

$$OUT = IN1 \& IN2$$

Dominant level "0"

Recessive level "1"



	Node 1 Dominant	Node 1 Recessive
Node 2 Dominant	<i>dominant</i>	<i>dominant</i>
Node 2 Recessive	<i>dominant</i>	<i>recessive</i>

DIGITAL COMMUNICATION TECHNIQUES

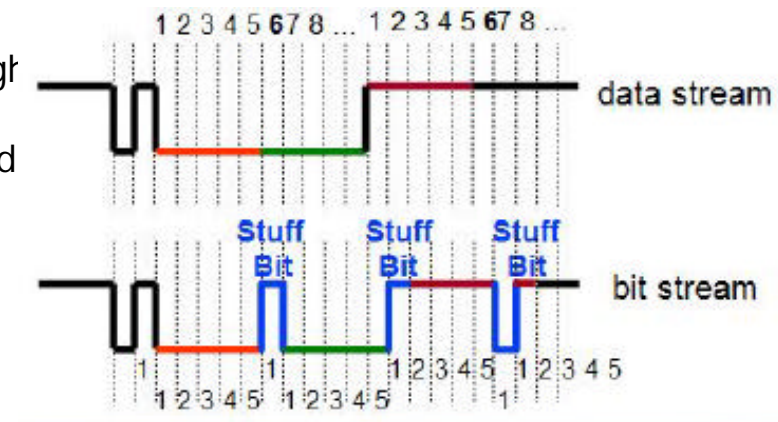
CAN Bus physical layer: bit timing configuration

<http://www.kvaser.com/wp-content/uploads/2014/08/boschcia99-can-bit-timing.pdf>

It consists of 3 sub-layers: PLS (Physical Signaling), PMA e MDI

PLS manages the signal temporization and synchronization through the so-called “bit stuffing” technique: every 5 consecutive equal bits, an inverse bit is inserted that will be discarded when received to improve the phase synchronisation

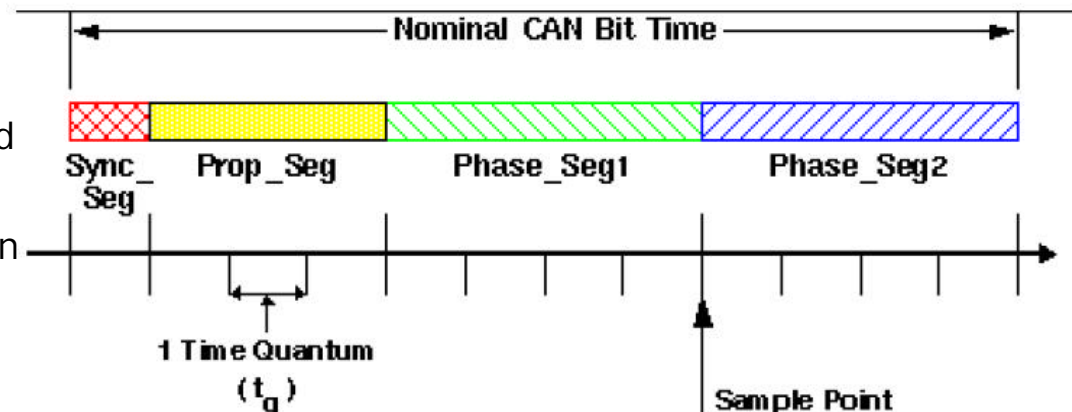
The time duration of a bit (Nominal Bit Time = $1/\text{bit_rate}$) is divided in 4 intervals each multiple of a suitable TIME QUANTUM (TQ) set through a local oscillatore:



Synchronization Seg. = 1 TQ where commutation edges must be inserted so as to synchronize the nodes

Propagation Time Seg. = from 1 to 8 TQ, used to compensate delays in tx/rx

Phase Segment 1 e 2 = with the same duration (1-8 TQ) are tuned so as transition from one to the other (it is the sampling point that is the reading of the level present on the bus) is compliant with the requirements of all the present devices (commutation edges)



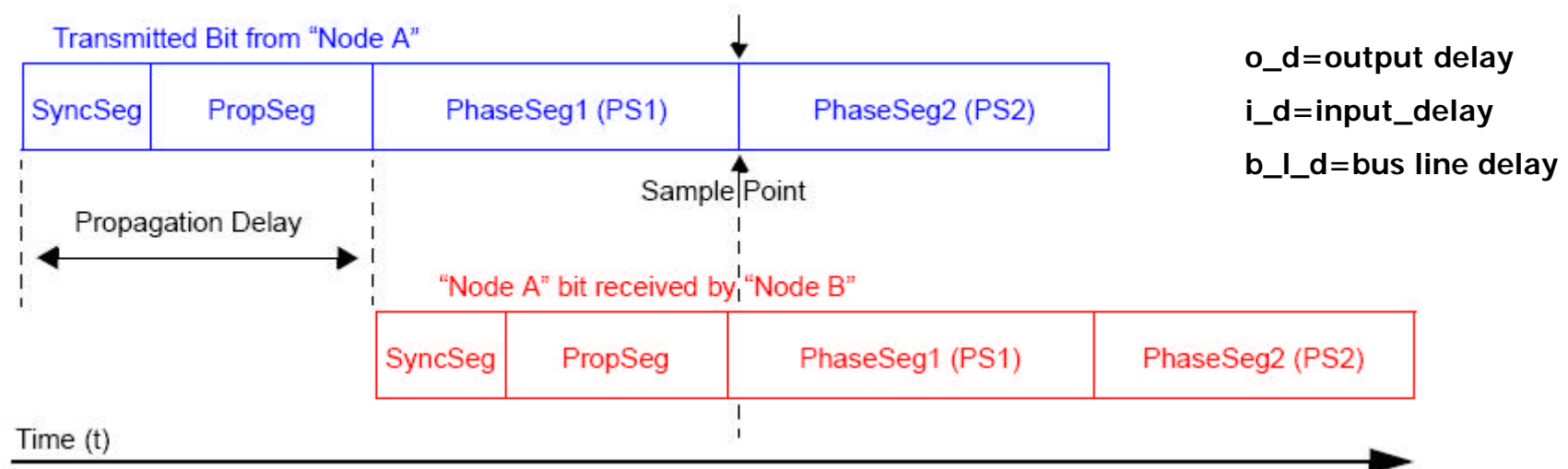
DIGITAL COMMUNICATION TECHNIQUES

CAN Bus physical layer: bit timing configuration

Every node must be able to sample the bit level within the same bit time: for instance in case of arbitration two nodes must “read” in the same time (bit time) dominant and recessive values put on the bus by the other nodes

In the example below if propagation delays that bring one of the nodes (ex. B) to sample after its own sample point could make the arbitration unsuccessful

If the bit rate is fixed the length of the connections must be limited: $t_{prop} = 2(o_d + b_l_d + i_d)$

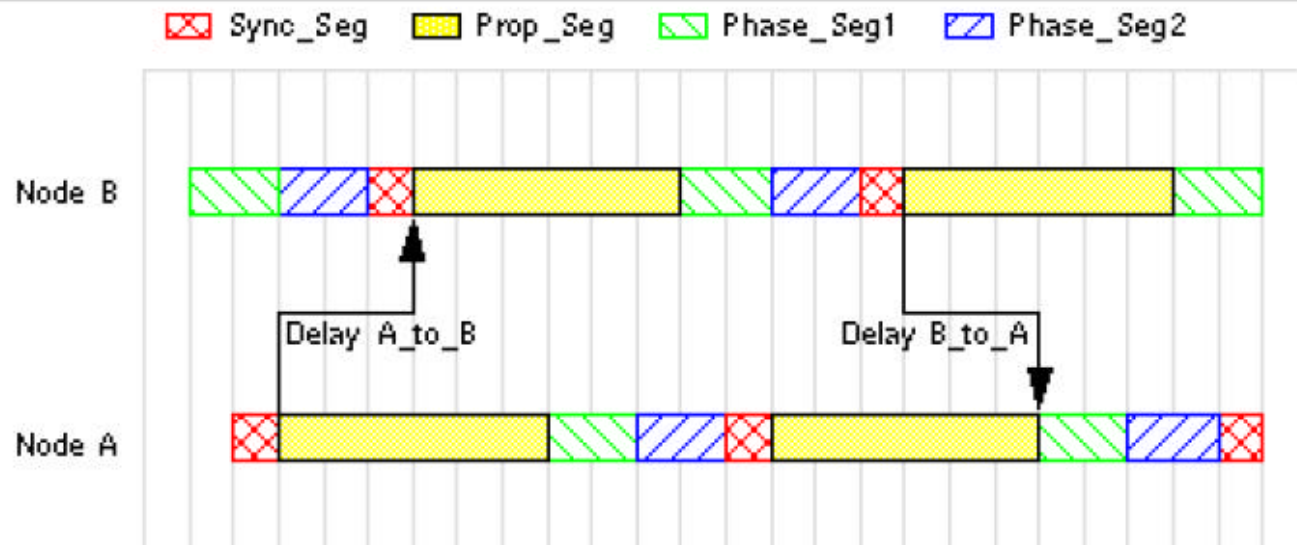


DIGITAL COMMUNICATION TECHNIQUES

PLS – compensation of phase shift to the distance

A e B are transmitting: initially A transmits a recessive level while B receives it after a delay due to the distance that separates them

If B transmits a dominant level on its own turn this level must be received before A continues in transmission since it thinks to be the arbitration winner. Since the Sample Point can happen in every point of the Phase_segment (see later), the bit sent by B must arrive to A before Phase_seg1 starts. This defines the length of the Prop_Seg (in the example 6 TQ)



$\text{Delay}_{A_to_B} \geq \text{node output delay}(A) + \text{bus line delay}(A \rightarrow B) + \text{node input delay}(B)$

$\text{Prop_Seg} \geq \text{Delay}_{A_to_B} + \text{Delay}_{B_to_A}$

$\text{Prop_Seg} \geq 2 \cdot [\max(\text{node output delay} + \text{bus line delay} + \text{node input delay})]$

DIGITAL COMMUNICATION TECHNIQUES

PLS: Phase Segment variation for synchronization

The length of PS is lengthened or shortened to compensate variations in local clocks (synchronization)

The synchronization among nodes is made on a recessive-dominant transition and keeps the distance between the edge and the sample so as to be always the same

The edges are detected on every TQ by comparing the present level of the line with that relative to the last sample point. If changes 2 possible actions:

Hard Synchronization: the bit time is reset to the moment when the edge arrives. It is done every time the start of frame arrives of the node that begins to transmit.

Bit Resynchronization: the PS are lengthened or shortened considering when the edge arrives; if after the Sync_Seg the phase shift is positive and PS1 is lengthened. If the edge arrives after the Sync_Seg, PS is made shorter.

The maximum shortening/elongation SJW (Synchronization Jump Width) goes from 1 to 4 TQ. Beyond this value the phase shift cannot be completely compensated but only partially and an error equal to (phase error – SJW) remains.

DIGITAL COMMUNICATION TECHNIQUES

PLS: Phase Segment variation for synchronization

The synchronization is performed not only to compensate clock variations, but also to compensate eventual noise spikes or due to the arbitration.

As already said, the synchronization among the nodes is done when the first node begins to transmit. However, since it is not said that this node will win the arbitration, the other nodes must synchronize with the winner and this operation is continuously performed.

Between two successive Sample Points only one synchronization can be done.

The same operation is done in the acknowledge field where the sender and the receiver synchronize themselves with the bit of «reception happened» (see later).

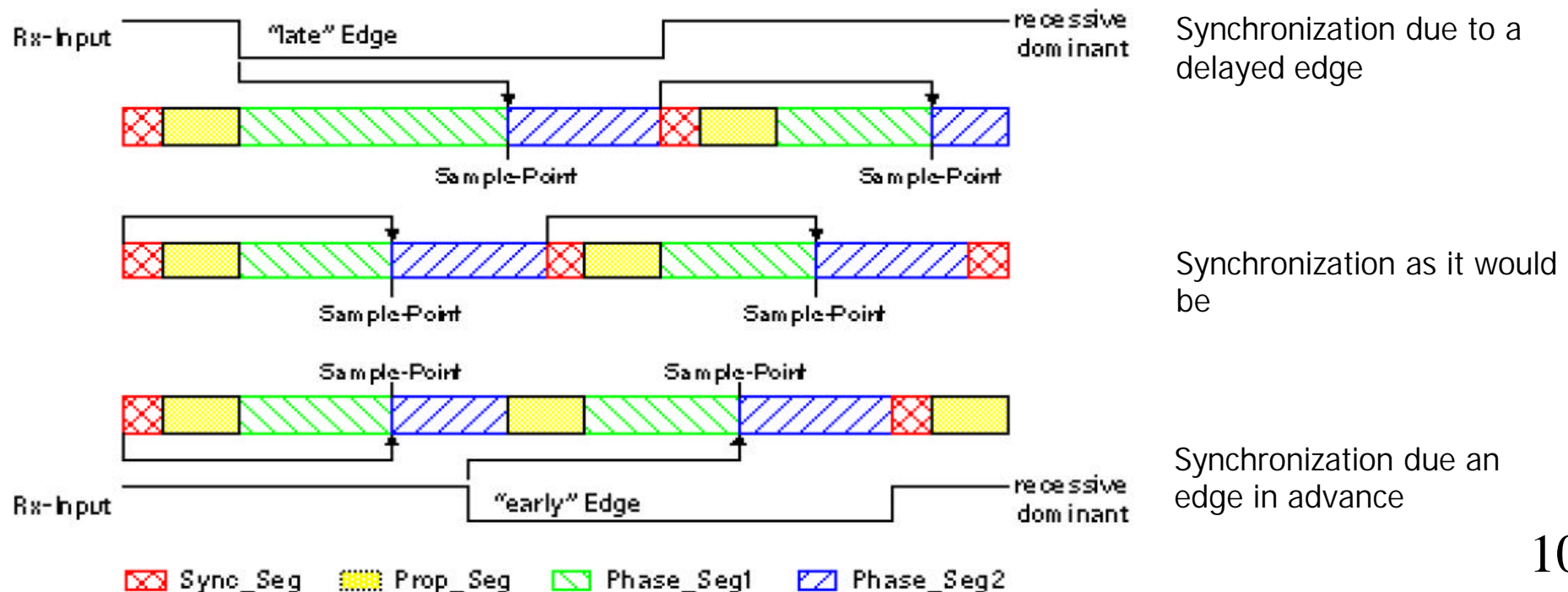
DIGITAL COMMUNICATION TECHNIQUES

Phase Segment variation in case of phase shift

In the first stream the edge is successive to the Sync Seg. PS1 is lengthened so as the distance from the Sample Point is the same compared when no delay is present.

In the second stream the edge arrives before the Sync Seg and is in PS2. Now the PS2 will be shortened and Sync cancelled so as to maintain the same distance between edge and SP as without anticipation.

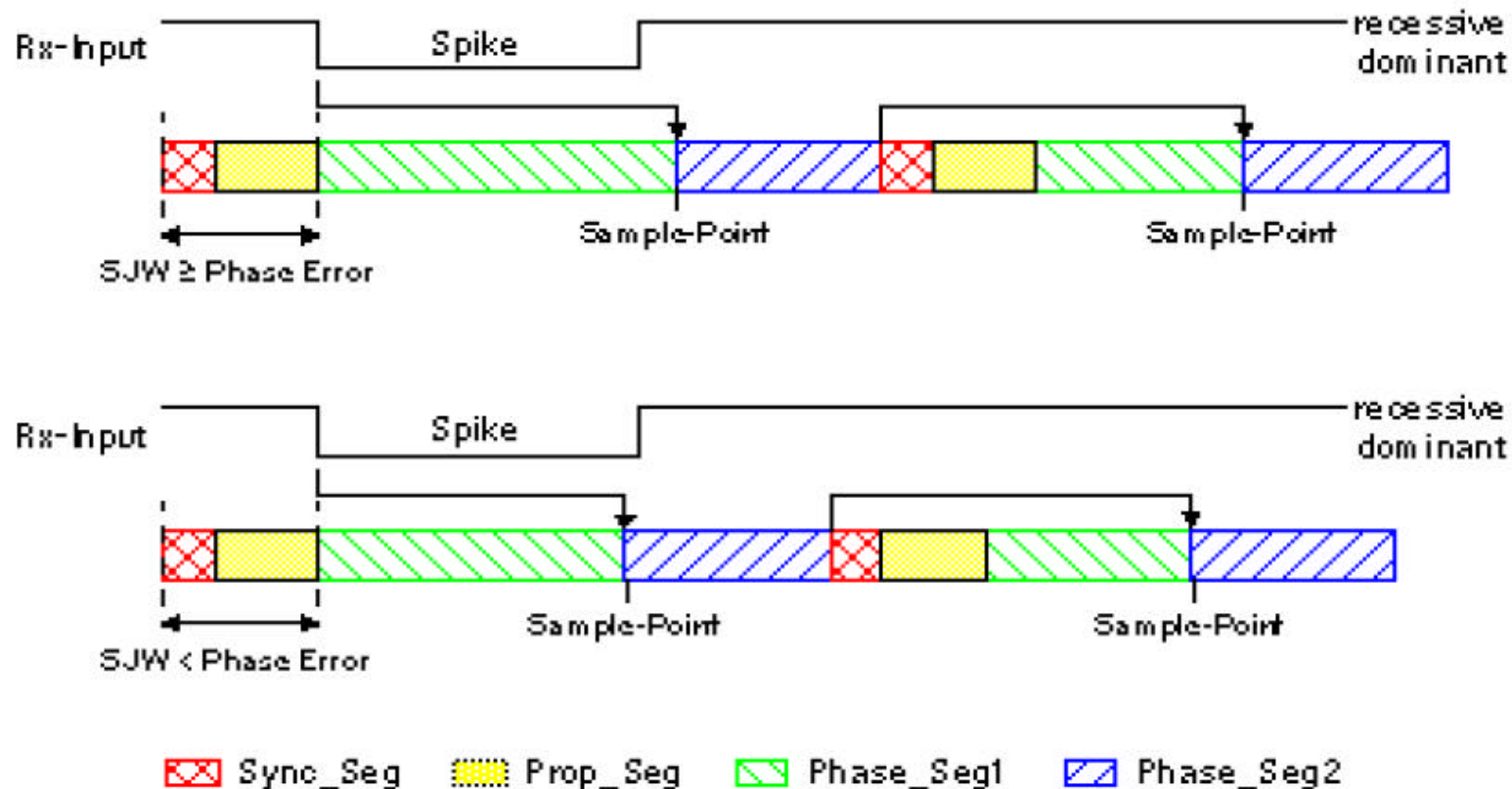
These variations are temporary but they imply a variation of the bit time.



DIGITAL COMMUNICATION TECHNIQUES

PLS: Phase Segment variation for synchronization

The picture below shows how a dominant spike will be filtered by the proposed synchronization technique.



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: physical level – PMA e MDI

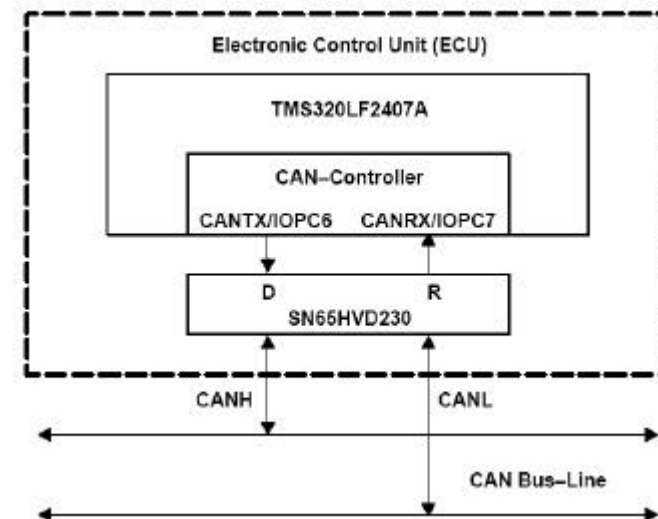
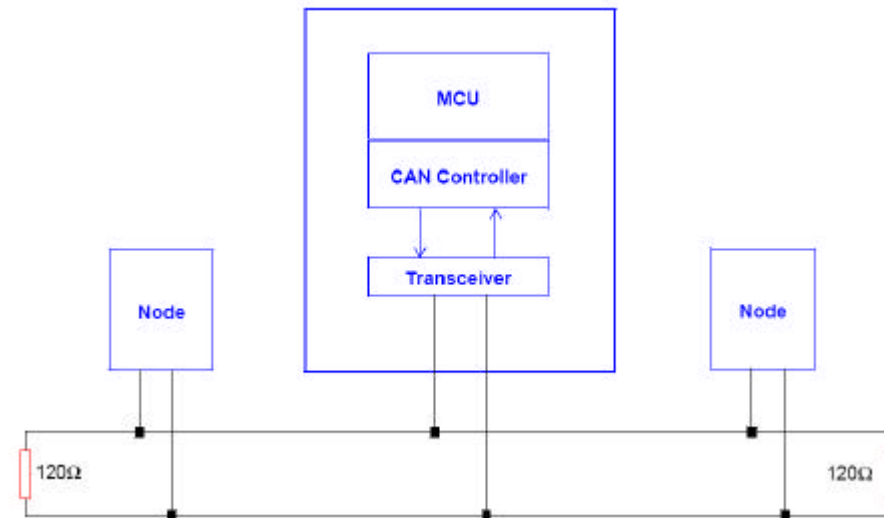
PMA (Physical Medium Attachment)

It defines the structure of the connection of every node to the bus through three components:

- Transceiver: bi-directional buffer to make access to the bus (detects the difference between the two lines CAN_H e CAN_L)
- Controller: manages tx/rx from and to the bus
- Microcontroller: “intelligent” part of the node controlling the other two

MDI - Medium Dependent Interface

It defines the physical and electrical characteristics in terms of connection type, adaptation impedance (120 Ω), line delay ...



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer

DLL is charged with:

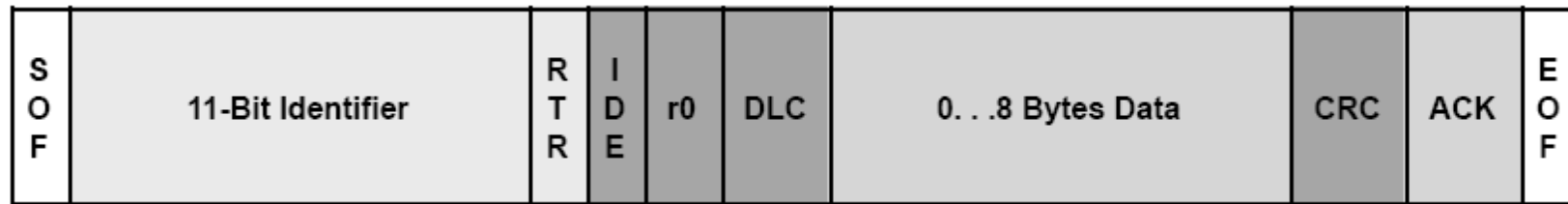
- **Message format definition**
- **Message validation**
- **Error management**
- **Error active nodes confinement**
- **Arbitration**

Message format:

- **Data Frame:** messages with data that must be kept by each node or discarded
- **Remote frame:** similar to DF but without data. It means “data request”
- **Error frame:** sent by a node that detects an error and causes the retransmission of a message. Each node calculates its own error rate: if high self-exclusion (fault confinement)
- **Overload:** sent by a “busy” node to ask to delay the next message dispatch
- **Interframe Space:** sent before each DF e RF, to separate them

DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Data frame

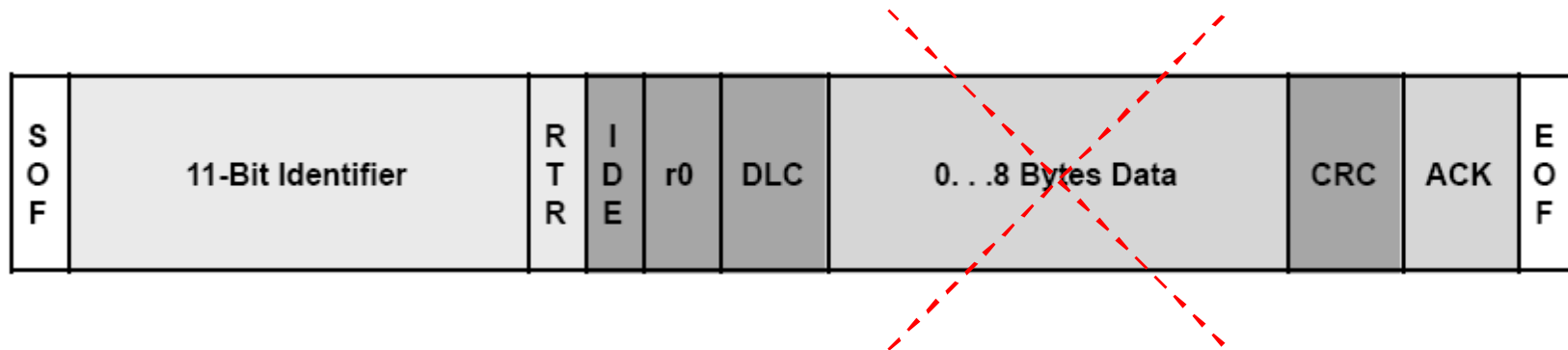


- **Start of frame:** 1 bit “dominant” = synchronism for beginning of DF (or RF)
- **Arbitration field (11/29 bit extended):** identifier of the message content plus bit RTR (Remote Transmission Request) to distinguish between DF and RF: in fact this bit is dominant for DF and recessive for RF. Thus if a request of datum simultaneous with their transmission, the request is aborted
- **Control field (6 bit):** 4 (DLC) specify the length of the Data Field (0-8 byte) while 2 (IDE and r0) are reserved

- **Data frame:** from 0 to 8 data bytes
- **CRC field:** 15 bit for CRC + 1 “recessive”. These bits are added to the frame so as the sequence is perfectly divisible by a prefixed polynomial term. If the validation is ok (null remainder), in ACK a dominant bit is written.
- **Ack Field:** 2 bits, ACK Slot e ACK Delimiter. By default they both are recessive: if at least one destination node received the message (although we do not who), it overwrites the ACK Slot with a dominant bit
- **End of frame:** 7 recessive bits

DIGITAL COMMUNICATION TECHNIQUES

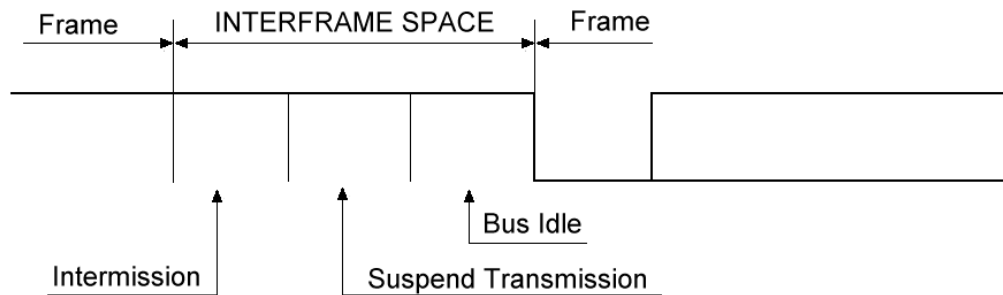
CAN Bus: Data Link Layer – Remote frame



- Very similar to the Data Frame with 2 important differences:
 - ✍ no data field
 - ✍ RTR is forced to be “recessive” so as to specify that this is a Remote Frame
- Used to ask for data whose type is specified in the “identifier” field
- The transmitter will answer with the information required in the data field

DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Interframe space

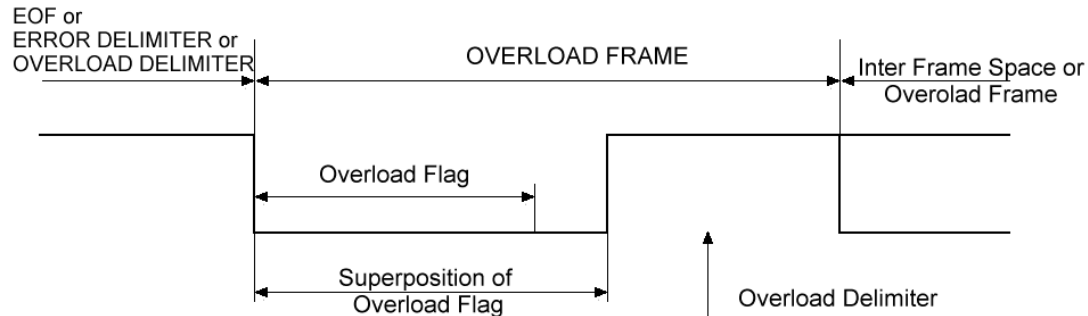


- Sequence of recessive bits to separate DF and RF
- Not used if Overload and Error are sent (they can also be consecutively dispatched)

- **Intermission:** the first 3 bits are "recessive" and it is not allowed to send other DF or RF; the Intermission bits can be interrupted only by an Overload Frame that prevents from sending the remaining ones. The Overload cannot break a DF
- **Suspend Transmission:** 8 recessive bits used by a node featuring transmission error rate not negligible. This node must wait 8 bits before sending again; if, in the meanwhile, another node will start a transmission, the node, if "error passive", can receive the message. In any case if the node is not "error passive" or it only receives, the suspend transmission field does not exist
- **Bus Idle:** sequence of "recessive" bits with not specified length that are interpreted as "free bus" and "ok to transmit" notification. This sequence ends when the SoF is encountered of a new Data or Remote Frame

DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Overload Frame



- Sequence of bits that delays the DF and RF emission if a node is “busy” and cannot correctly process other data. It is allowed to send it only correspondingly to the first 3 bits of an Intermission frame

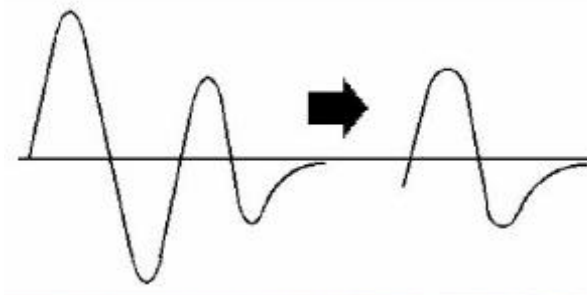
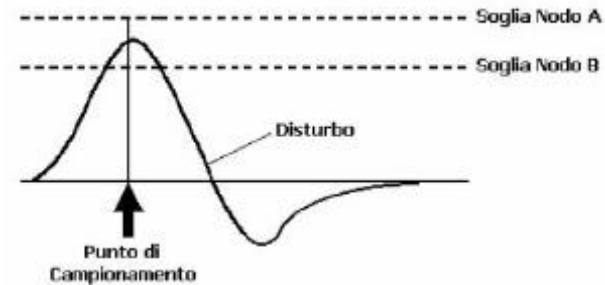
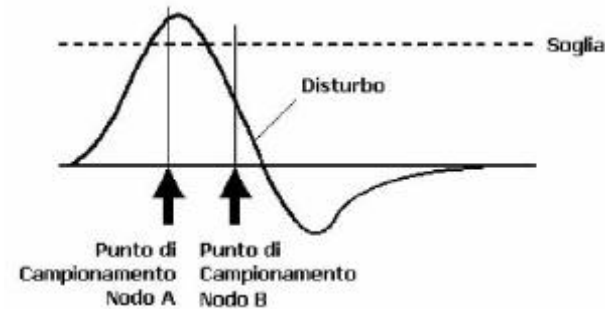
- In other words a “dominant” bit received during the I and II bit of an Interframe is interpreted as condition of “other busy node” that can be acknowledged by sending another Overload
- The Frame consists of two fields: Flag e Delimiter
- Flag: 6 “dominant” bits that can be sent within the Intermission field of an Interframe. In this case all the other stations of the bus detecting this condition answer on their turn with a Overload flag. If the 6 bits are sent after they will be interpreted as SoF of an DF or an RF generating an error condition since they violate the bit stuffing condition
- The Overload frames delay all nodes communication, therefore only 2 consecutive Overload are allowed by the same receiver. A mechanism is given to confine those nodes that due to their slowness are not able to process the received messages
- Delimiter: 8 recessive bits of ending frame

DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Error Frame

Possible error sources are both to mechanisms internal to the nodes and to noise

- phase shifts in sampling time,
- different discrimination thresholds
- signal attenuation on the line (reading error)



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Error management

- Error detection consists of the following steps:

- ✍ The node controller detects an error and transmits an Error Frame
- ✍ All the connected nodes ignore the not valid message
- ✍ The controller is upgraded
- ✍ The message is re-transmitted

- The detectable errors:

- ✍ *Bit monitoring*: if the bits on the bus are different from those that were sent (by the winner of the arbitration) a bit error occurs. This can happen everywhere but in the “identifier” field where it is considered as losing the bus control (see later: bus arbitration)
- ✍ *Bit stuffing*: if 6 consecutive equal bits are on the bus a Stuff Error is detected
- ✍ *Frame Check*: a format not corresponding to one of the 5 possible triggers a Form Error
- ✍ *CRC*: the received code does not correspond to the received sequence a CRC error occurs
- ✍ *Acknowledgement*: if the sent frame has been received, the ACK field must be overwritten with a “dominant” bit. If the transmitter does not detect this, it notifies an acknowledgement error

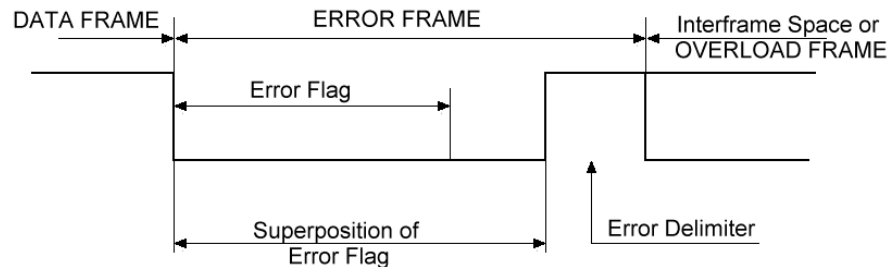
DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Fault confinement

- Each node has two fault counters for reception (REC) and transmission (TEC) that in case of faults add 1 for a receiving error and 8 for transmission error. If the transmission or the reception are ok they are diminished by 1.
- On the basis of the values of its own counters, each node can set itself as:
 - ✎ **Error active: both the counters are < 128 .** The error rate and their level do not disrupt the normal traffic on the bus so the node can play an active role and send error frames if necessary
 - ✎ **Error passive: if at least one counter is > 128 .** The node causes faults quite frequently and is allowed to send only passive error frames (no automatic re-transmission of the messages) but must wait a greater interval than active nodes before sending again (interframe with suspend)
 - ✎ **Bus off: if one counter is $= 256$.** The node is seriously affected by error so the access to the bus is forbidden.
 - ✎ Suitable bit sequences have been conceived in the standard that can be sent by the user code in the application layer, to re-enable a “bus off” node and to reset it as “error active” (with $REC=TEC=0$). For example a sequence with 128 frames with 11 consecutive recessive bits.
- This error management policy is very efficient so as a not detected fault can be considered a very rare event during the bus activity.

DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Error Frame



That node detecting an error in a DF or RF sends the Error Frame

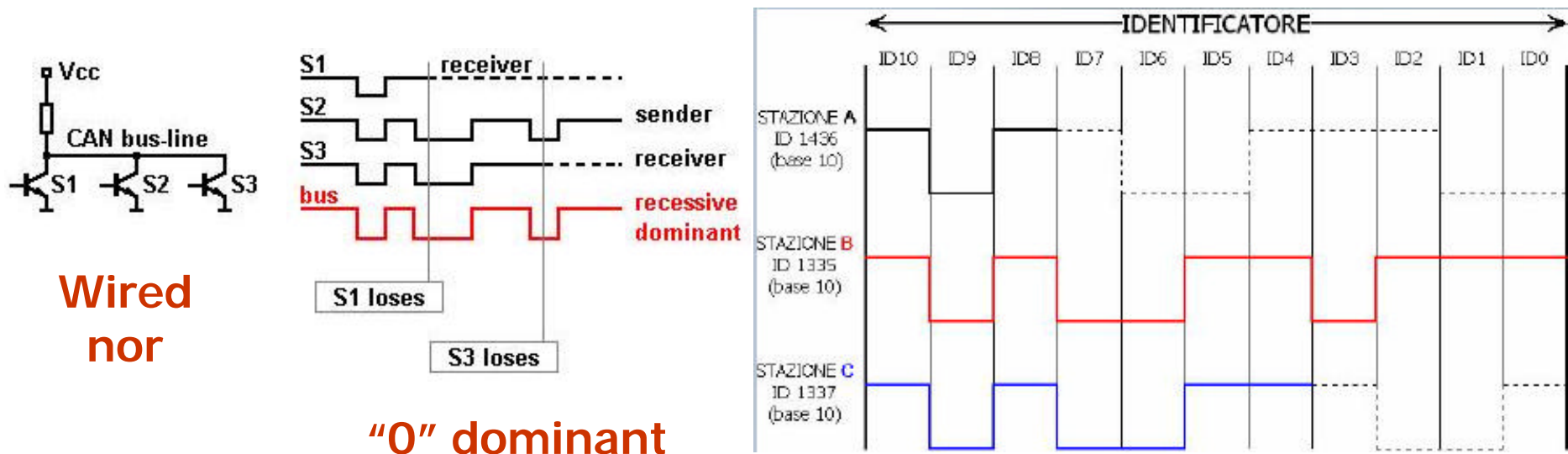
The Frame consists of two fields: Flag and Delimiter

- Flag: 6 bits dominant or recessive depending on the type of node
 - ✍ active error flag (dominant): if a node is qualified as “error active” by the fault confinement system detects an error, it sends on the bus 6 “dominant” bits that violate the stuffing rule causing the error detection by the other connected nodes (they will send in turn error flags). The traffic present at that moment is destroyed
 - ✍ passive error flag (recessive): they are sent by a node qualified as “passive”. They can be overwritten by an Active Error Flag but they can’t overwrite the ACK and EOF fields of a DF thus they can’t destroy the traffic
- Superposition of error flags: the other 6 bits sent by the other nodes (at maximum 12 dominants bits, all in all)
- Delimiter: 8 recessive bits identifying the End of the Frame

DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Arbitration

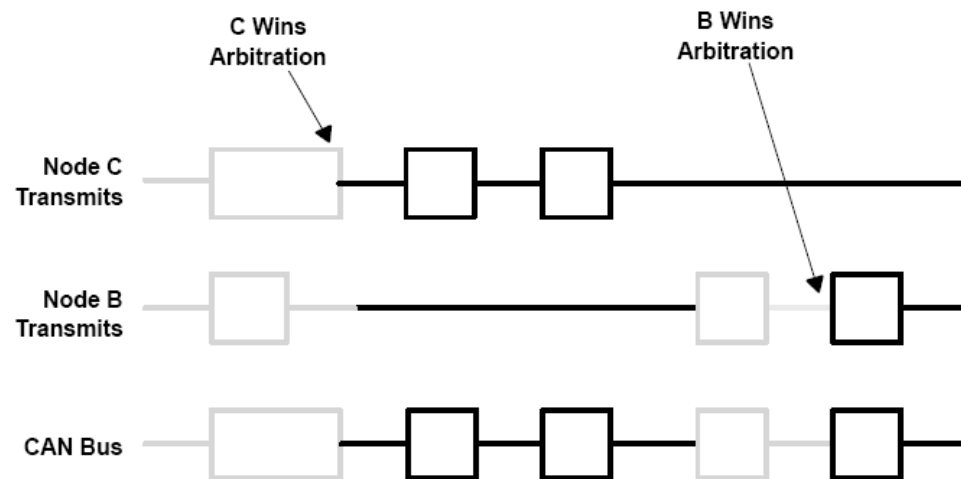
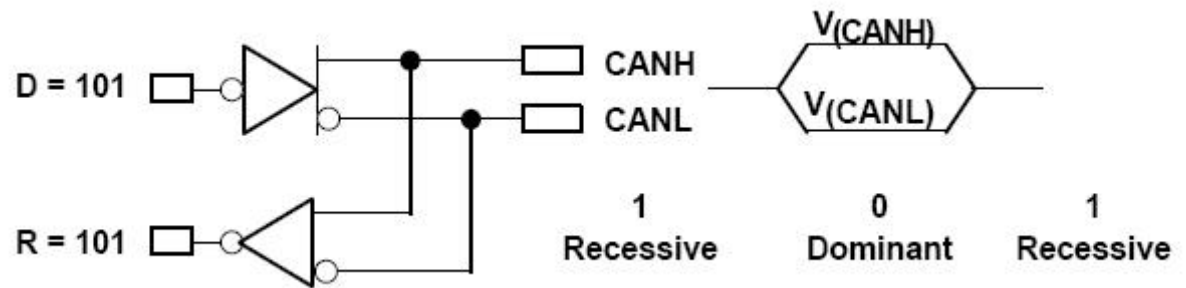
- When a node transmits, it writes the SOF first and then its own identifier: if many nodes try to contemporaneous transmission, the winner among them will be the minor value owner
- Pre-emptive priority mechanism
- During the arbitration, a node monitors the bus and compares the read level with the transmitted one: if the two bits do not correspond and in particular if the node sent a recessive bit, it stops its own transmission.



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer – Arbitration

- In a CAN node the logic level "1" is associated to the recessive level while the "0" to the dominant.
- This happens since the line must be set "recessive" by the transceiver when not driven so as to minimize the power consumption (if the level is pulled high no current passage).
- During the arbitration each node monitors the bus and compares the read level with that transmitted: if the two bits are not coincident and in particular if the node sent a recessive bit, it breaks its transmission.



DIGITAL COMMUNICATION TECHNIQUES

CAN Bus: Data Link Layer

- Object Layer: filtering (the packet received can belong to the node or not), interpretation of the received messages, interfacing with Application level
- Transfer Layer: error detection and management, faults confinement, arbitration, handshake and synchronization, message packing and packet distribution

